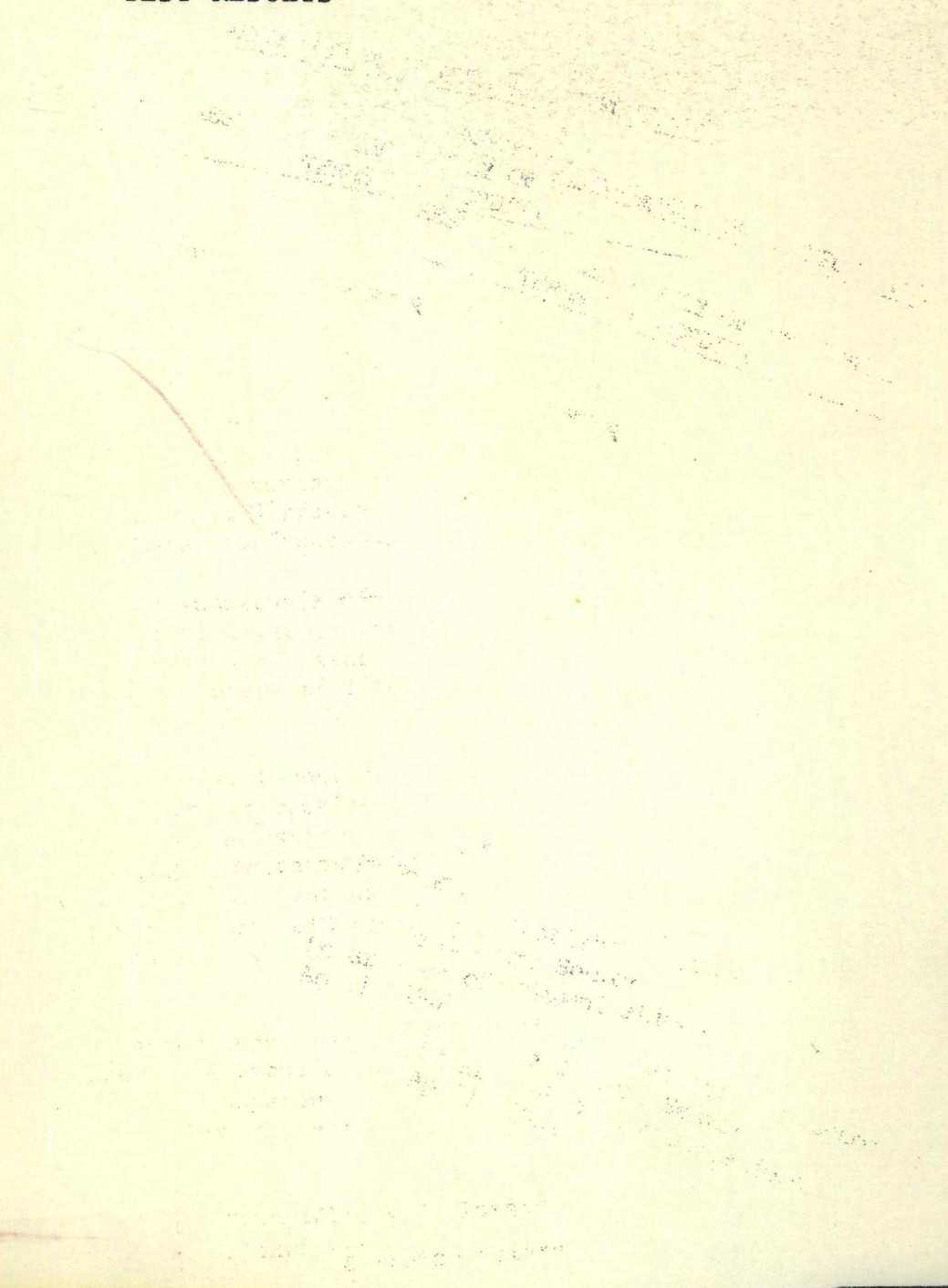


SCPC INTERFERENCE STUDY

TEST RESULTS



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SCPC INTERFERENCE STUDY
TEST RESULTS

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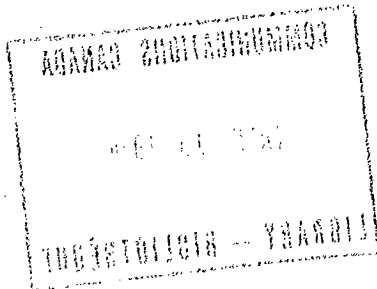
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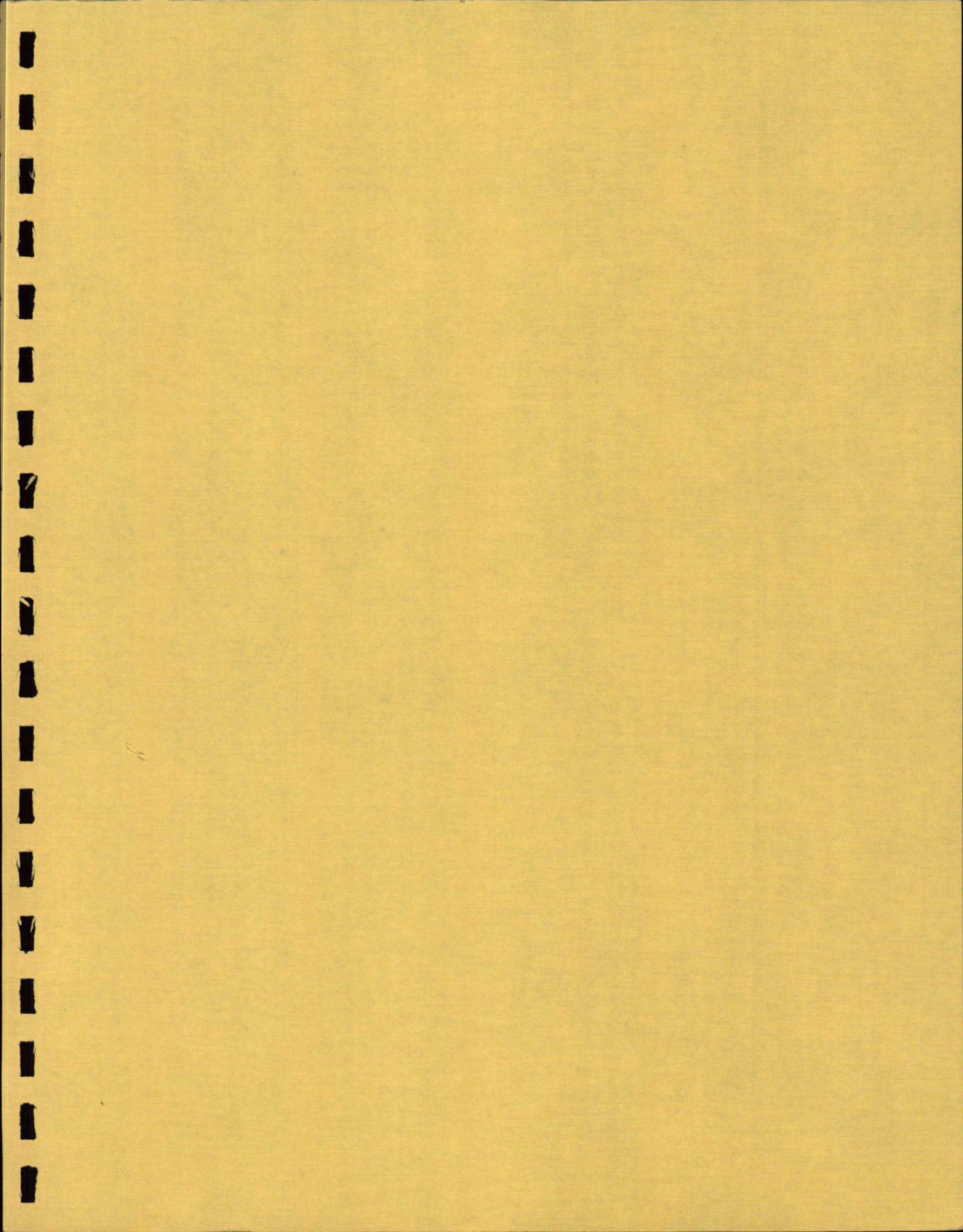
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SCPC INTERFERENCE STUDY

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1.

TEST SET-UP DESCRIPTION

A general representation of the Interference performance of the SCPC link test set-up is shown in Figure 1.

The S/N / BER Measuring equipment injects a signal, a 1 kHz tone or a bit stream, whichever is the case, into the SCPC transmit equipment, compares it to the received signal from the SCPC receive equipment after being interfered with through the series of combiners, and detects the noise captured in the channel link or the errors that occurred during transmission in order to give a read-out in the form of a signal to noise ratio (S/N), an impulse noise count or a bit error rate (BER).

The SCPC Transmit and SCPC Receive equipments can be of the analog FM type or the digital QPSK type. The received IF carrier level is set with the attenuator next to the output of the XMT equipment, to the nominal value of -37 dBm for FM-SCPC and -43 dBm for QPSK-SCPC.

Separate attenuators are used for the white (Thermal) noise generator and the interfering signal(s) to allow the variation of one parameter keeping the other constant.

In some cases either one of the combiners in Figure 1 is bypassed or terminated to allow measurements with a single type of interference (i.e. white noise only or signal only).

Figure 2 describes the set-up for FM-SCPC type channel link.

For one interfering carrier, the 4-way combiner is bypassed at points P1 and P2. A video waveform generator is used to provide a full-field colour bar signal of 1 V p-p (nominal) of amplitude. This signal is passed through an attenuator-amplifier-preemphasis chain before it is fed to a two-way splitter which combines it with the synchronized energy dispersal waveform. This triangular waveform is synchronized with the field rate (or line rate) using the vertical drive (or horizontal drive) output of the video waveform generator.

For four interfering carriers, the 4-way coupler is inserted between P1 and P2 such as the colour bar modulated signal constitutes one of its four inputs. The second input is a dispersed unmodulated carrier, while the third and fourth carriers are modulated by two picked up live TV signals from two different local channels; each of these two signals is fed to a BPU (Baseband Processing Unit) where it is subjected to amplification, preemphasis and variable attenuation as well as 30 Hz energy dispersal, if so is the case; while in the case of line rate (or half-line rate) dispersal the 30 Hz EDW is switched off and the synchronized dispersal frequency is mixed with the video signal through an external combiner (see special circuits).

This part of the set-up, down to the 4-way combiner, holds for digital SCPC measurements as well.

For Radar like interference measurements a pulsed sine wave signal of preset amplitude and period is injected at point P2, replacing the video signals (see Figure 4).

Figure 3 shows the set-up for digital QPSK-SCPC interference measurements. In this case the interfering signal(s) and noise are added through the 30 dB couplers existing on the Channel Unit Test Panel, which acts as a transponder.

In analog FM-SCPC measurements, a 1004 Hz tone at -5.5 dBm, in 600 ohms load, is fed to the SCPC transmit channel; the received tone is set at -2.2 dBm/600 ohms. When delivered to the TMS and the TNMS the tone is notched and the interference is C-MSG weighted, then an internal power meter detects the energy content of the interference relative to that of the tone.

In digital, QPSK-SCPC measurements, a bit stream of length 2048 bits is repeatedly injected into the SCPC transmit channel at the rate of 64 Kbits/sec. The received bit stream is compared to that transmitted and the occurring errors are counted for a presettable period of time corresponding to 10^n bits, where n is set equal to 5, 6, 7, 8 or 9 for the most convenient reading.

2. INTERFERENCE LEVEL CALIBRATION

2.1 VIDEO SIGNALS AND EDW

The modulators have a carrier frequency of 70000 ± 0.7 kHz and a nominal output of +5.2 dBm. Their sensitivity is calibrated by Bessel zero method, to -21 dBm for 1 MHz RMS deviation; hence the triangular waveform at the modulator input is set to 24.5 mV p-p per 1 MHz deviation. The video level at the modulator input is set so that a 15 kHz sine wave of the same amplitude as the video signal (1 V p-p nominal) delivers at this point 165.6 mV p-p (for 6.8 MHz peak-to-peak deviation).

2.2 VIDEO CARRIERS

The interfering carrier(s) level is set by comparison. The procedure consists of (refer to Figures 2 and 3):

- Terminating the noise input.
- Measuring the SCPC signal level at point P4 when P3 is terminated.
- Set the attenuator located between P2 and P3 as to provide at point P4 the same reading as for the previous step when the SCPC carrier is switched "off".
- Take note of this attenuator setting at which $C_0/I = 0$ dB.
- Add to the attenuator setting the desired C_0/I level (dB).

2.3 WHITE NOISE IN FM-SCPC

The noise level is set, for analog FM-SCPC interference, by the comparison method using a calibrated filter. The procedure is as follows (refer to Figure 2):

- Insert the calibrated filter between point P5 and the m watt-meter.
- Measure the SCPC carrier level when the noise input is terminated.

- Switch the carrier off and reconnect the noise generator.
- Set the noise generator attenuator as to provide the same reading as the previous step.
- Reduce the attenuation by the value: $10 \log (B/b) - K$
 where: B = noise bandwidth of the calibrated filter
 (= 22 863 kHz)
 b = SCPC channel bandwidth (= 26 kHz)
 K = desired C_0/N (= 15 dB).
 (for the given values, the attenuation decrease is:
 $10 \log (22863/26) - 25 = 14.4$ dB)
- Take note of this attenuator setting and reconnect the set-up as in Figure 2.

2.4

WHITE NOISE IN QPSK-SCPC

The noise level for digital QPSK-SCPC interference, is set as follows (refer to Figure 3):

- Set the noise generator attenuator to its maximum attenuation.
- Terminate P3.
- Set the SCPC carrier level as to provide -43 dBm at point P4.
- Set the slide switch inside the downconverter drawer of the RCVR equipment to MGC (manual gain control).
- Switch "IN" the 14.2 dB attenuator on the CU Test Panel.
- Connect an RMS voltmeter to the QPSK MON. on the SCPC demodulator and read the signal level.
- Switch the SCPC carrier off.
- Set the noise attenuator as to provide the same reading as the previous step.
- Take note of this attenuator setting where $C_0/N = 14.2$ dB.
- Set the slide switch back to AGC.

- Switch "OUT" the 14.2 dB attenuator.
- Modify the noise attenuator setting as to obtain the desired C_0/N instead of 14.2 dB (i.e. for $C_0/N = 15$ dB increase the attenuation by 0.8 dB).

2.5

PULSED CARRIER

The interfering pulsed carrier level, for radar like interference test, is set in a similar way to that of the video carrier (see section 2.2). The procedure consists of (refer to Figures 2 and 4):

- Measuring the SCPC carrier level at P4 as in section 2.2.
- Setting the pulse generator for a 10% duty cycle pulse, 1 V p-p nominal, using an oscilloscope and/or a frequency/period counter.
- Setting the attenuator between points P2 and P3 to maximum attenuation.
- Setting the 70 MHz modulator to pulsed mode and to maximum output level.
- Setting the attenuator for equal reading, as in section 2.2, which corresponds to $C_0/I_p = -10$ dB.
- Increasing the attenuation by 10 dB where $C_0/I_p = 0$ dB and keeping note of this setting to which is added the desired C_0/I_p .

3. TEST CONDITIONS

The test conditions are set to meet, with few exceptions, the contract requirements. These exceptions are imposed by the statistical nature of the interference process, the limited equipment facility, the spectral energy distribution of the interfering signals, etc.

The changes that had to be done are listed below:

3.1 MEASUREMENT SET A: Q-PSK-SCPC

3.1.1 *Set of Measurements A1 - No changes.

3.1.2 *Set of Measurements A2 - The range of Co/N is reduced to vary from +10 dB to +17 dB instead of +6 dB to 35 dB; due to equipment limits and time of measurement limits.

3.1.3 *Set of Measurements A3 - Due to the spectrum of the full-field color-bar video signal, the test was performed at the worst case frequency separation ($f_c = 1.64$ MHz) for no error was observed at $f_c \leq 500$ kHz and $C_o/I_u \geq -11$ dB. The range of Co/Iu in this case is from 0 dB to +8 dB instead of +10 dB to +35 dB. (See spectrum photographs No. 1 and No. 2).

3.1.4 *Set of Measurements A4 - Since the energy dispersal waveform (EDW) have to be synchronized with the video signal because of very practical reasons (see remark ***) and since this signal does not contain a steady 100 Hz component, the measurement set with $f_L = 100$ Hz was omitted and replaced by $f_L = 7.875$ kHz. Measurements with $f_c = 10$ kHz were replaced by $f_c = 22.5$ kHz for all the carriers are AFC operated and the 70 MHz occurs halfway between two digital SCPC channels separated by 45 kHz. The range of C_o/I_D is shifted down to -12 dB to +9 dB instead of +10 dB to +35 dB for the worst BER at $C_o/I_D = +9$ dB was 5×10^{-7} . The frequency separation $f_c = 3.6$ MHz was added to the f_c range in order to show the effect of the color carrier. Another value of f_c , that of the worst case, was included in the results.

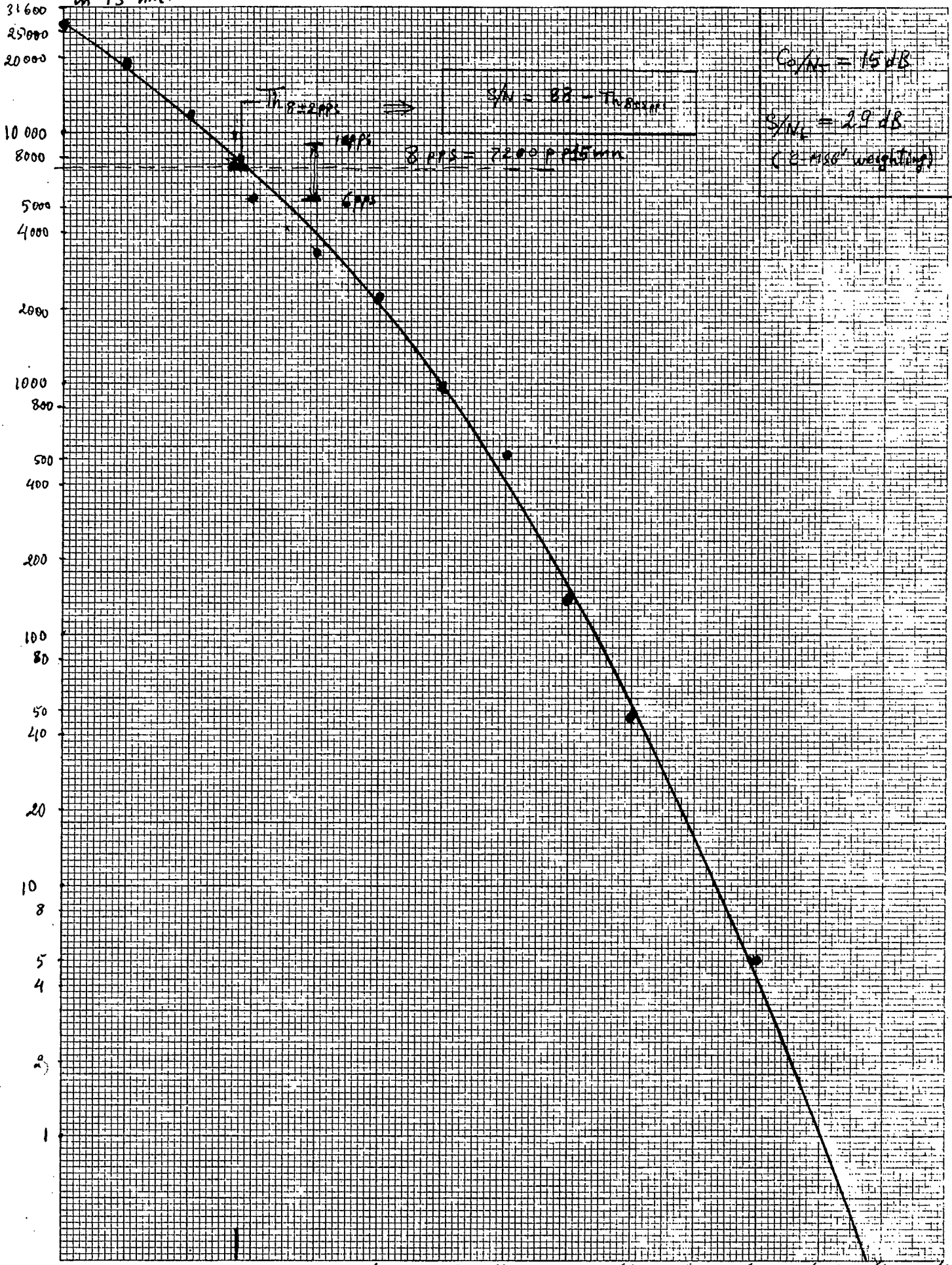
3.1.5 *Set of Measurements A5 - The same changes as for A4 are applied.

3.1.6 *Set of Measurements A6 - These tests were made at $f_c = 10$ kHz only for the following reasons:

**The spectrum envelope of the radar signal is proportional to

$$S = \left[\text{PRF} * \text{PW} * \frac{\sin(\pi * f_c * \text{PW})}{(\pi * f_c * \text{PW})} \right]^2$$

Impulse Count
in 15 min.



$C_0/N_0 = 15 \text{ dB}$
 $S/N_c = 29 \text{ dB}$
 (2-1000 weighting)

$S/N = 28 - \text{Threshold}$
 $S/N = 2200 \text{ p.p. } 15 \text{ min}$

Th = 2200
 1000
 600

Threshold setting (dB)

because of the sine function, the spectrum presents gaps at the frequencies where (see spectrum photographs)

$$\pi * f_c * PW = K * \pi \quad K = \pm 1, \pm 2, \dots$$

i.e. $f_c = K/PW$; and

— for $PW = 10\mu s$ and $30\mu s$, gaps occur at $f_c = 100$ kHz, 500 kHz, 1 MHz and 2 MHz.

— for $PW = 1\mu s$ and $3\mu s$, gaps occur at $f_c = 1$ MHz and 2 MHz.

Therefore measurements at 1 and 2 MHz frequency separations were omitted. **The main envelope peak level, at $f_c = 0$, is $20 \log (PRF * PW)$ dB below the CW power I_p . The first side peak is 13.46 dB below the main peak. Thus, and as seen in the results, the interference is very weak for relatively high frequency separation i.e. i.e. $f_c \geq 100$ kHz.

3.1.7 *Set of Measurements A7 - This same changes as for A6 are applied.

*** Remark: The energy dispersal waveform (EDW) of frequency f_L should be synchronized with the video signal in order to eliminate any possible very low frequency beating, in which case the interfering signal fluctuates noticeably.

On the other hand, it is useless to think of a video signal dispersed with EDW of frequency other than a harmonic, or a sub-harmonic of either the line rate or the frame rate, for it is impossible, cost-effectively, to extract the image from the dispersed signal.

3.2 MEASUREMENT OF SET B: COMPANDED FM-SCPC

In this series of measurement, as well as for four interfering carriers, a major modification concerning the impulse noise count was made.

Because of the very steep slope of the impulse count versus the threshold setting and the very time consuming nature of this type of measurements, it was thought necessary to measure the curve of impulse noise count versus threshold, once and for all, and then proceed to measure a normalized threshold setting (that of 8 ± 2 pulses per second).

A threshold detector circuit was designed to meet these requirements; for details see "special circuits" attached hereto.

The other changes are:

- 3.2.1 *Set of Measurements B1 - No change.
- 3.2.2 *Set of Measurements B2 - No change
- 3.2.3 *Set of Measurements B3 - A multiburst video signal was used. The C_0/I_u range is shifted to cover -30 dB to +10 dB, for the results were steady above $C_0/I_u = +5$ dB.
- 3.2.4 *Set of Measurements B4 - Measurements were made at only $f_b = 30$ Hz and $f_L = 15.75$ kHz (see 3.1.4). The range of C_0/I_D is -30 dB to -2 dB for S/N measurements, and -30 dB to +10 dB for the equivalent S/N of the impulse noise measurements.
- 3.2.5 *Set of Measurements B5 - The same changes as in B4 are applied.
- 3.2.6 *Set of Measurements B6 - The same changes as in A6 (3.1.6) are applied, except for the C_0/I_p range: -30 dB to +30 dB.
- 3.2.7 *Set of Measurements B7 - The same changes as in B6 are applied.
- 3.3 MEASUREMENT SET C - Comp. FM-SCPC with four interfacing carriers.
 - 3.3.1 *Set of Measurements C1 - The S/N ratio in this set was measured according to the condition stated in the Contract. (detailed in section 2.2.1 of the SOW) with the exception that
 - **measurements with $f_L = 100$ Hz were omitted for the reasons stated earlier in this report (section 3.1.4);
 - **the frequency separation $f_c = 3.6$ MHz was added to the f_c range.
 - **the range of C_0/I_D was shifted to cover -20 dB to +20 dB.
 - 3.3.2 *Set of Measurement C2 - The equivalent S/N ratio of the impulse noise count is measured in the same conditions as in C1.

3.4 MEASUREMENT SET D - Q-PSK-SCPC with four interfering carriers

One major modification to this set of measurements was made due to very important fluctuations in the BER count; the BER curves were substituted by one point, each, where the interference effect is undoubtedly noticeable over the effect of thermal noise. This point was called the onset of degradation, and was defined as the C_0/I_D ratio at which the BER count is three times that of thermal noise alone. In the present case, the onset of degradation is the C_0/I_D where the BER count equals 10×10^{-6} , in contrast with 3.3×10^{-6} for thermal noise only ($C_0/N = 15$ dB in the channel bandwidth). The other modifications are the same as in C1. The results of this set of measurements are listed in tabular form.

4. SUPPLEMENTARY RESULTS

Some additional measurements were taken and are included in this report. The results of these measurements are thought to be helpful in the interpretation of the results obtained for the measurements stated in the contract.

4.1 *Set of Measurements S1 - In this set, the interference effect on Q-PSK-SCPC of a single carrier modulated with live TV signal, dispersed with $f_L = 30$ Hz @ $\Delta f_{pp} = 1$ MHz; the carrier to interference ratio was set to $C_0/I_D = 0$ dB and $C_0/N \rightarrow \infty$. The frequency separations are chosen for the worst possible cases when the SCPC carrier coincides with a spectrum peak of the TV signal (most of the time). The results are given in the form of a chart showing the BER versus real time.

4.2 *Set of Measurements S2 - In this set, the interference effect on Q-PSK-SCPC of an undispersed, unmodulated carrier with frequency separation of $f_c = 10$ kHz and 22.5 kHz, is pointed out.

4.3 *Set of Measurements S3 - In this set of interference effect on comp. FM-SCPC of a radar-like signal of 30μ s pulse duration and 3 kHz pulse repetition frequency at different frequency separations is pointed out.

TEST

SET-UP

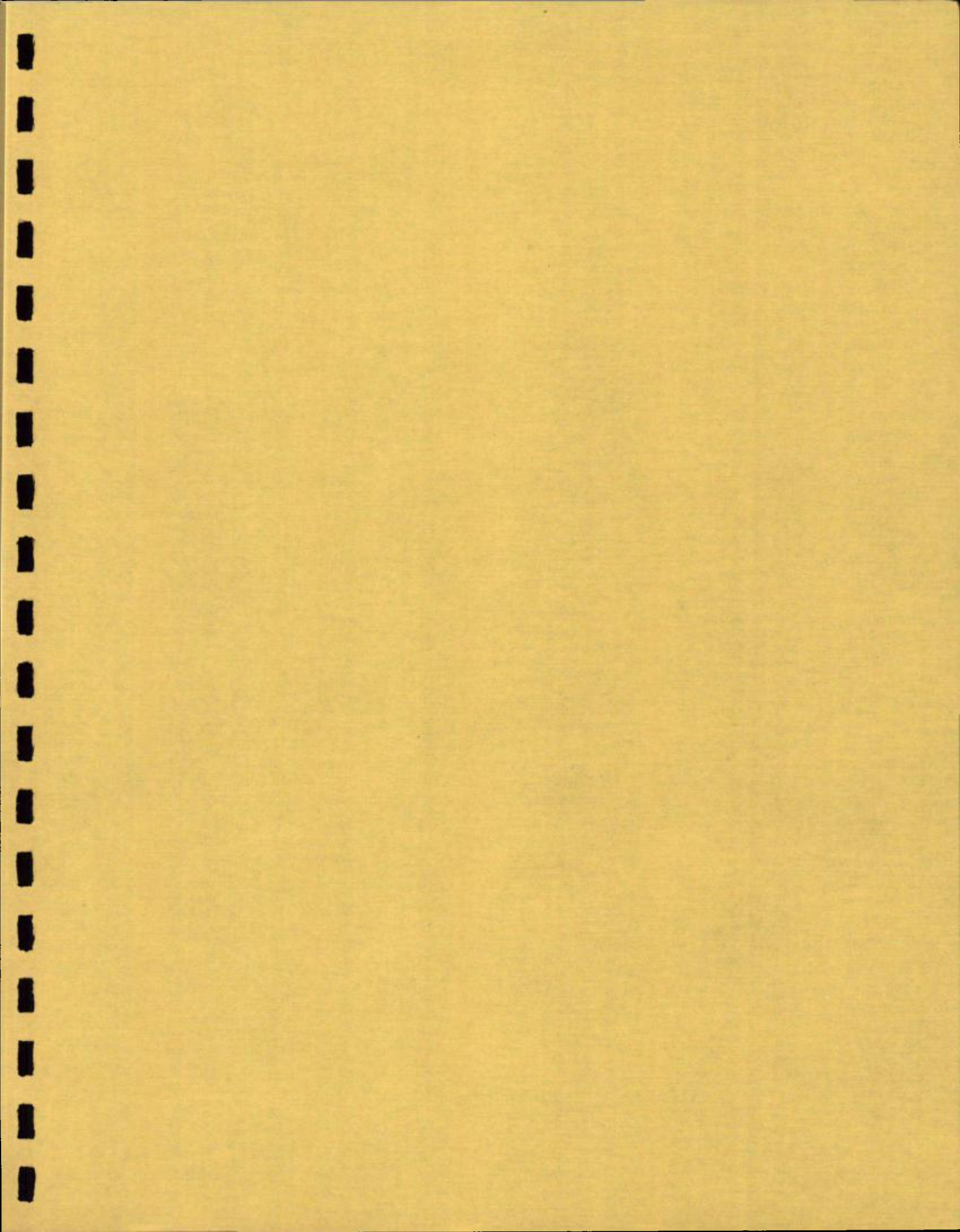
SCPC/FM & SCPC/PSK

- 4.4 *Set of Measurements S4 - In this set the interference effect on Comp. FM-SCPC of an unmodulated carrier, dispersed with $f_L = 15.75$ kHz, 30 Hz and 0 Hz (undispersed) is pointed out, for different f_C , covering the range 10 kHz to 750 kHz by steps of 30 kHz (up to 340 kHz).

5. CONCLUSION

From the obtained results we can conclude the following:

- 5.1 *The interference effect on a SCPC link is highly dependant of the interfering signal spectrum as well as its power level with respect to the SCPC carrier.
- 5.2 *The EDW should be synchronized with the video waveform in order to prevent extra bursts of high interference.
- 5.3 *Higher frequency EDW gives the SCPC more tolerance to interference. The same holds for higher peak-to-peak deviation.
- 5.4 *Line rate dispersal with triangular waveform operation seems to be not very cost effective and one might suggest operation with saw-tooth waveform instead. It is thought that such waveform gives similar results as the half-line rate triangular waveform (7.875 kHz).
- 5.5 *The SCPC link shows high immunity to radar-like interference of short pulse duration, but this type of interference affects more adjacent channels than longer pulse duration type does.



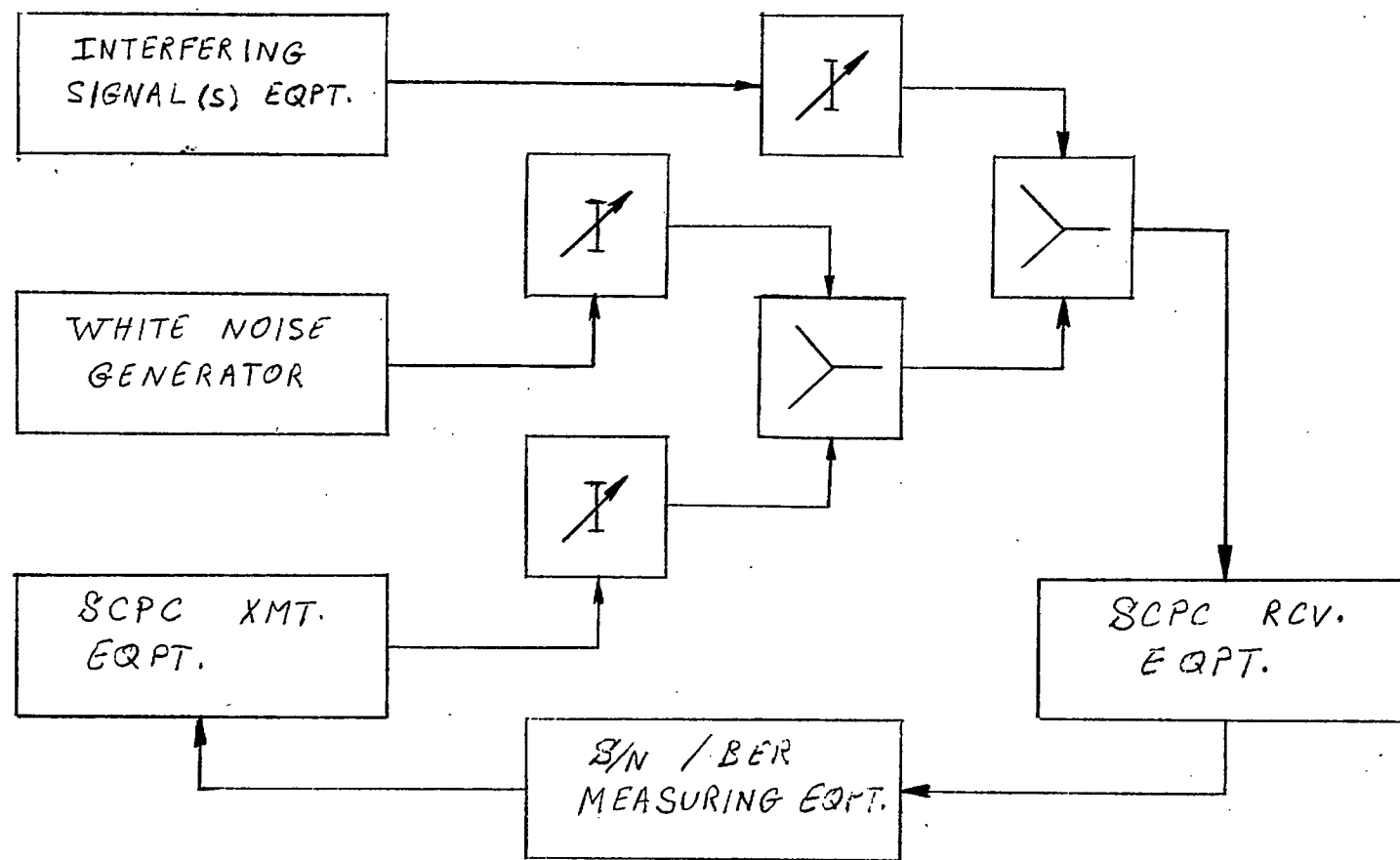
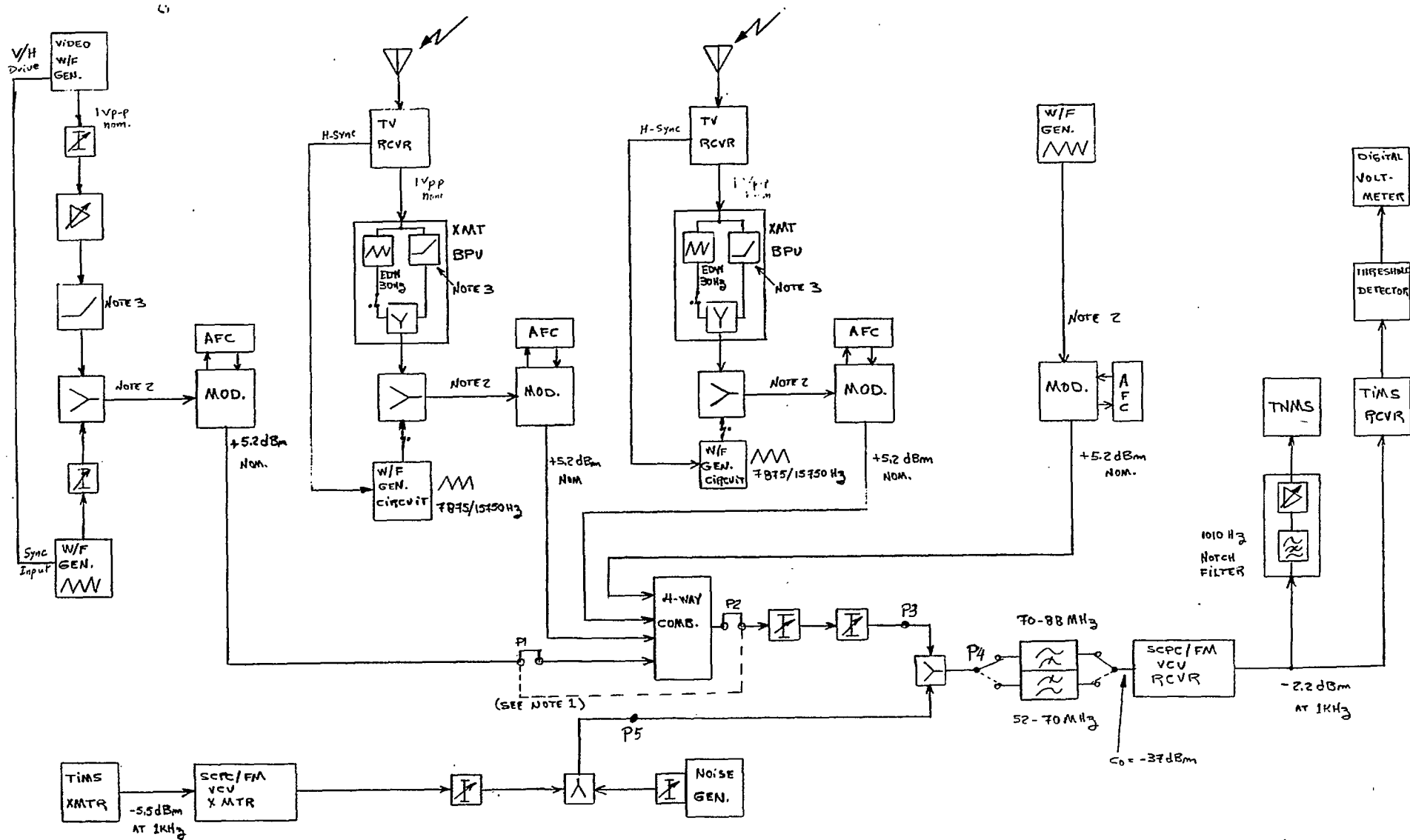


fig 1 - General representation
of the interference test set-up



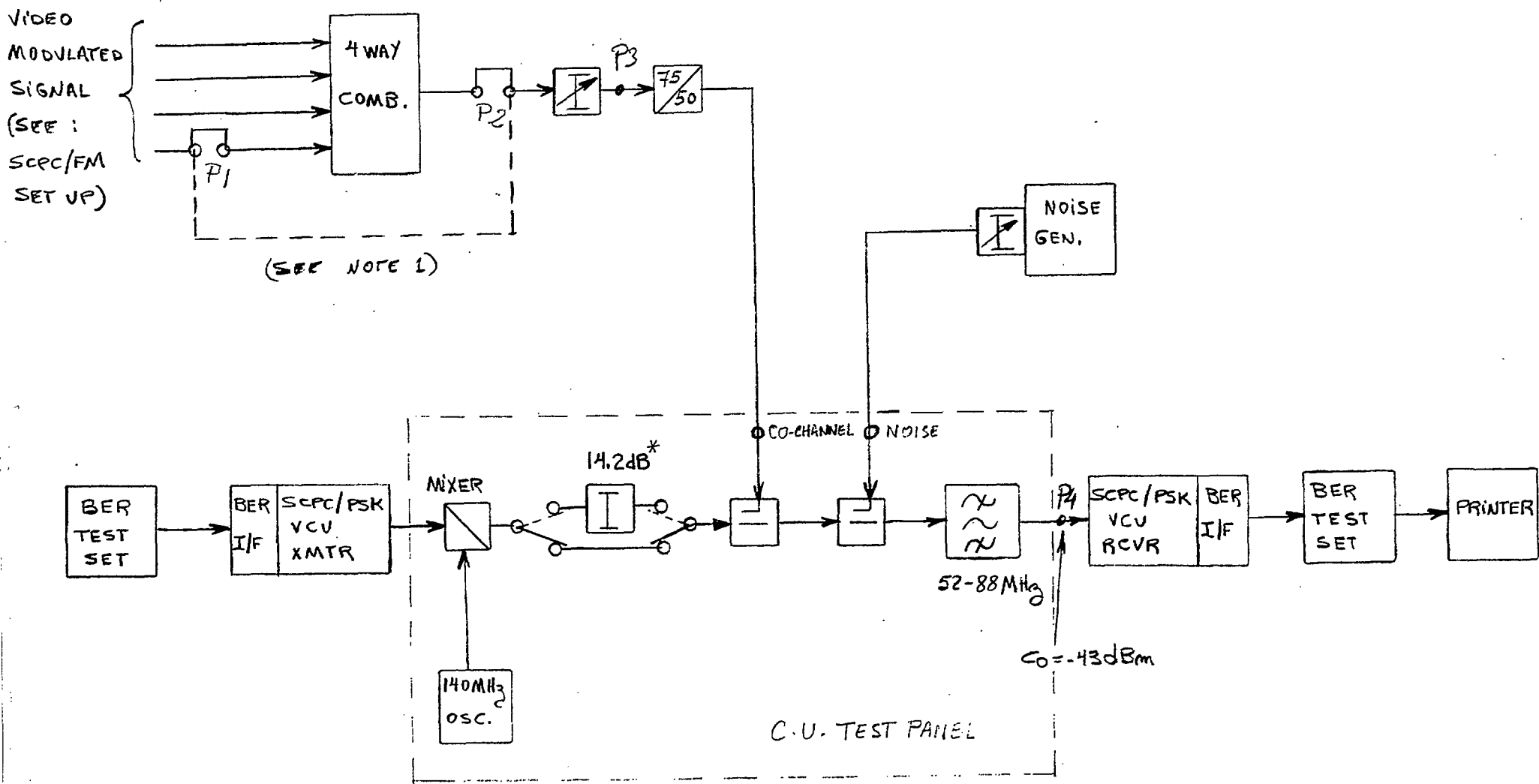
ABBREVIATIONS

- BPU : BASEBAND PROCESSING UNIT
- EDW : ENERGY DISPERSAL WAVEFORM
- TMS : TRANSMISSION IMPAIRMENT MEASURING SET
- TNMS : TEST NOISE MEASUREMENT SET

- VCU : VOICE CHANNEL UNIT
- W/F : WAVEFORM

Fig 2- SCPC/FM
INTERFERENCE STUDY

TEST SET-UP



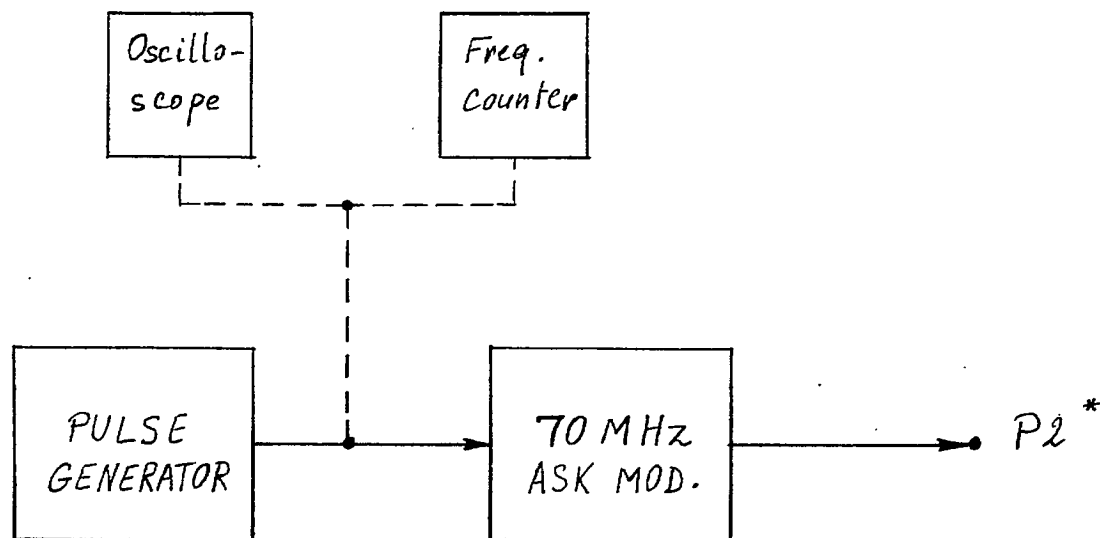
ABBREVIATIONS

- BER : BIT ERROR RATE
- I/F : INTERFACE
- PSK : PHASE SHIFT KEYING
- C.U. : CHANNEL UNIT

* FOR CALIBRATION PURPOSE

fig 3 - SCPC/PSK
INTERFERENCE STUDY

TEST SET - UP



* To substitute the 4-way splitter in figs 2 and 3 in the case of radar like interference

fig 4 - SCPC/RADAR
INTERFERENCE STUDY
TEST SET-UP

NOTES ON TEST SET-UP

1- FOR ONE INTERFERING CARRIER TESTS, REMOVE P1 AND P2 AND
CONNECT AS SHOWN BY THE DOTTED LINE

2- LEVELS AT MODULATOR BASEBAND INPUT

VIDEO SIGNAL : 165.5 mV P-to-P AT $f = 15 \text{ MHz}$ FOR A
DEVIATION OF 6.8 MHz P-to-P

DISPERSAL SIGNAL : 24.5 mV P-to-P FOR A DEVIATION OF
1.0 MHz P-to-P

MODULATOR SENSITIVITY : -21 dBm FOR 1.0 MHz RMS

3- PRE-EMPHASIS

525 LINES CONFORMING TO CCIR RECOMMENDATION 405-1

THRESHOLD DETECTOR
AND
WAVEFORM GENERATOR
CIRCUIT

SCHEMATIC DIAGRAMS

THRESHOLD DETECTOR DESCRIPTION

The purpose of this circuit is to provide a quick measurement of the threshold level at which the impulse noise count is 8 ± 2 p.p.s. (pulses per second). Since the slope of the Impulse Count versus the threshold setting is extremely sharp (squared exponential) at this point and since this curve is almost independent of the type of interference, the threshold at 8 ± 2 p.p.s. seems to be a very accurate measurement of the signal to noise ratio.

The input to the threshold detector is connected to the output of the notched C-MSG filter board inside the TIMS (HP2943A).

This input shows a relatively high impedance to the output Op-Amp on the filter board.

The first Op-Amp in the detector circuit provides an adjustable gain to the received noise. This gain is presettable, using P_1 , to a value between unity and $\times 100$.

The next stage provides a log conversion (linear to dB conversion) through the diodes D_2 and D_3 .

The third stage is an amplifier that permits scale adjustment through potentiometer P_2 .

The last Op-Amp stage compares the incoming impulse noise level to the transistorized peak detector level (consisting basically of Q_1 , C_5 and R_{11}) filtered through R_{10} and C_4 .

Component values and time constants are experimentally selected to provide 8 ± 2 current spikes through Q_1 per second.

R_{12} and C_6 deliver a smoothly filtered dc output level that is to be measured with a DVM.

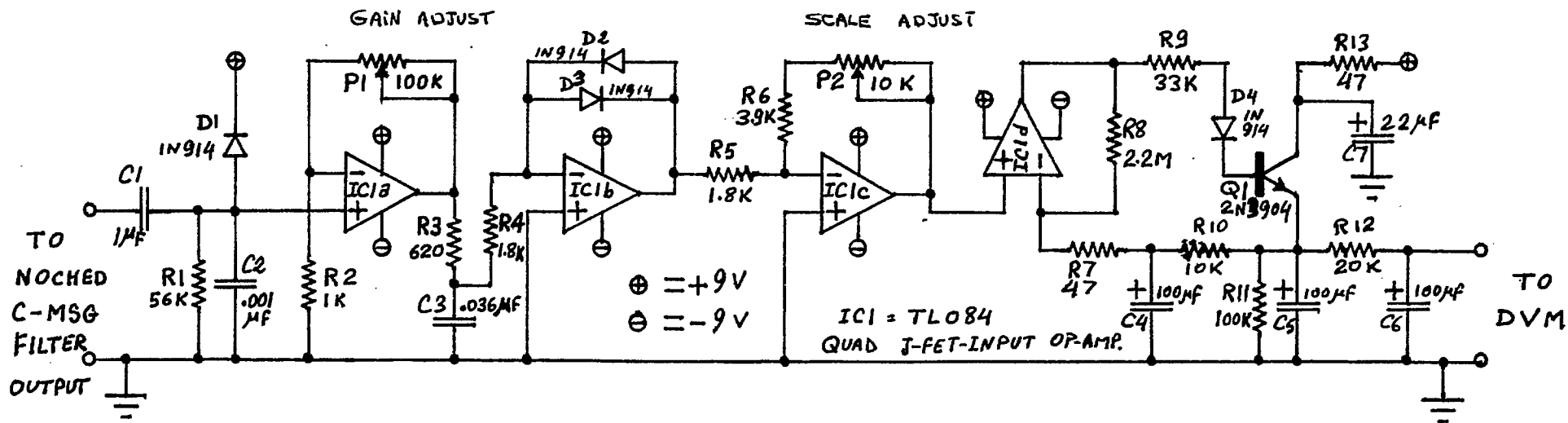
Calibration: it was observed that when white noise only was interfering with FM-SCPC, the threshold setting for the same impulse count was to be varied dB/dB in step with C_0/N . Hence this procedure was used to calibrate the scaling of the threshold detector (i.e. by varying C_0/N in the range of interest). Potentiometer P_2 is set to provide 20 mV/dB, and P_1 is set when $C_0 \gg I_D$, to output 2.00 Vdc.

When in operation, to determine the Th_{8pps} (dBRN), the following equation is used:

$$Th_{8pps} \Big|_{C_0/I_D} \text{ (dBRN)} = \frac{\text{DVM Reading (V)} - 2.00 \text{ (V)}}{0.02 \text{ (V/dB)}} + 59 \text{ (dBRN)}$$

And the signal to impulse noise is (for C-MSG Filtering)

$$S/N \text{ (dB)} = 88 \text{ (dB)} - Th_{8pps} \text{ (dBRN)}$$



CIR. 1 -

TRESHOLD DETECTOR

TRIANGULAR W/F GENERATOR

The purpose of this circuit is to provide a 15.75 kHz/7.875 kHz triangular waveform synchronized with the TV signal line with adjustable phase and amplitude to serve as energy dispersive waveform.

The input to this generator is connected to the horizontal sync test point inside the TV monitor. It takes an input voltage swing of 8 to 50 Vpp.

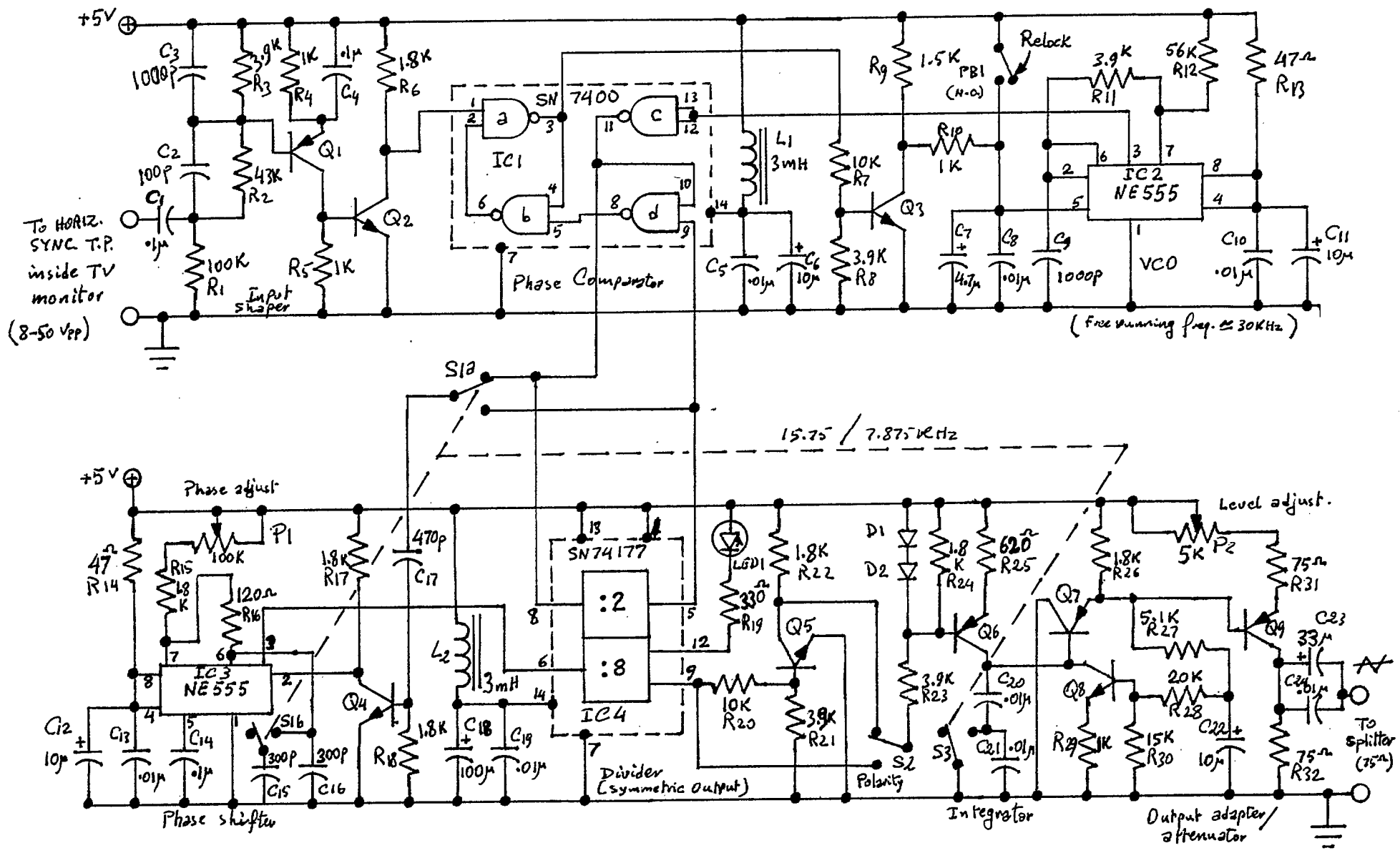
The input waveform is shaped in Q1 and transformed to TTL level in Q2. Its phase is compared to that of the VCO in the NAND Gates of IC1, and the phase difference is amplified in Q3, filtered through R10 and C7 and fed-back to the VCO (IC2) as frequency control voltage.

The free running frequency of the VCO is approximately 30 kHz; and its locking frequency is divided by 2 (or 4) in IC4 to provide 15.75 kHz (or 7.875 kHz) symmetrical squarewave after being delayed in IC3 by an adjustable phase shift presettable through P1.

The inverter Q5 and the switch S2 provide a selectable polarity.

The square-wave is then integrated in C20 and C21 (or C20 only) to give a triangular waveform that is attenuated to an adjustable level (through P2) and converted into a current source with source impedance of 75 Ω through Q9 and R32; C23 provides ac coupling to the 75 ohm load (the splitter). An automatic dc level control to the triangular waveform is provided, through Q8, R28, R29, R30 and C22, in order to maintain maximum linearity.

The swing at the collector of Q8 is 1.6 V p-p, which provides an output swing of 12 to 800 mV p-p when loaded, i.e. -27.2 to +9.3 dBm into 75 ohm load.

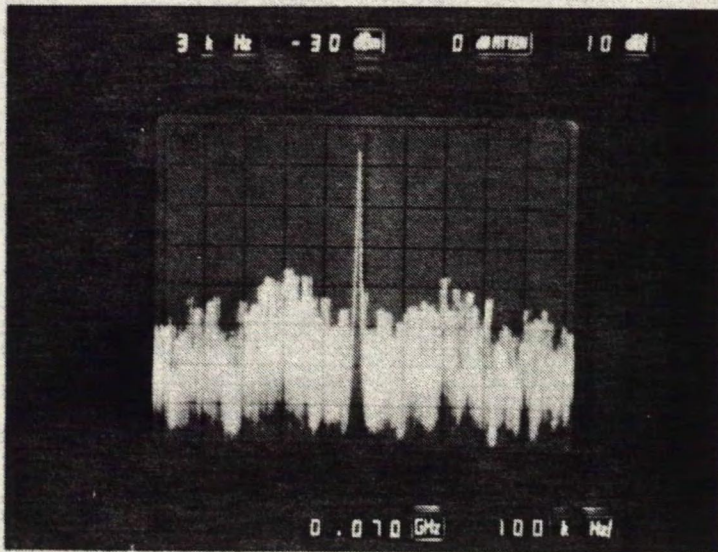


CIR.2- CIRCUIT DIAGRAM OF SYNCHRONIZED 15.75/7.875 KHz TRIANGULAR W/F GENERATOR

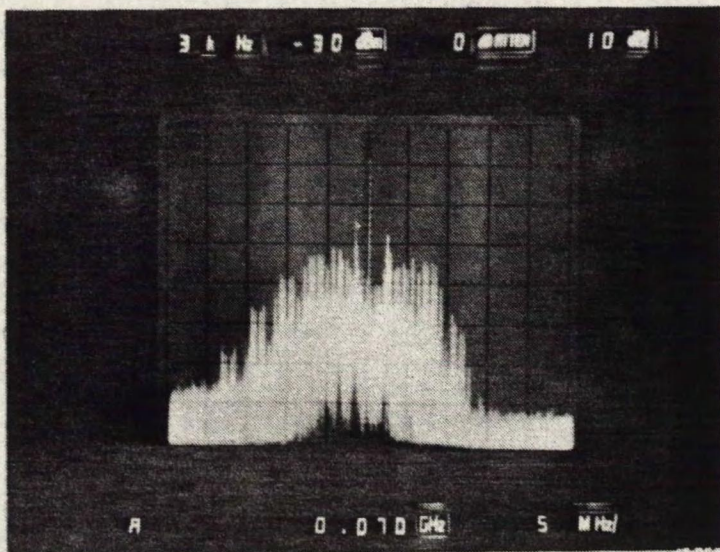
SPECTRUM

PHOTOGRAPHS.

FM MODULATED SIGNAL WITH COLOR BAR



SP-1

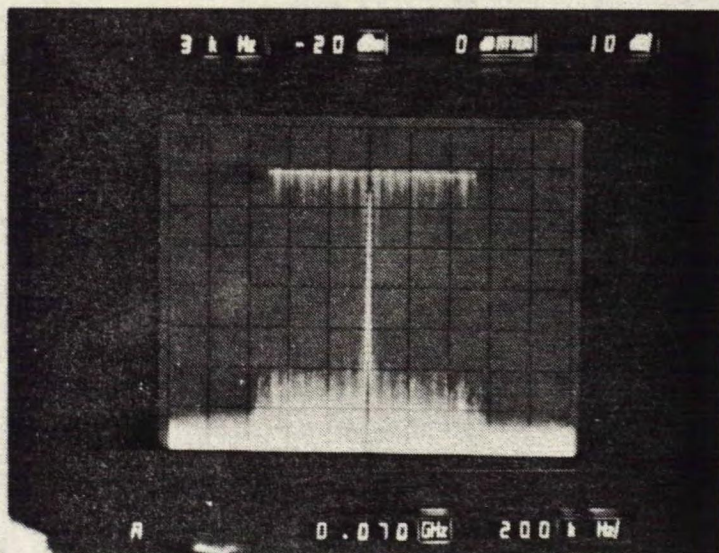


SP-2

$$f_c = 10 \text{ kHz}$$

$$C_o/I_U = 0 \text{ dB}$$

I_U : MODULATED WITH
COLOR BAR
AT 6.8 MHz
P-to-P DEVIATION



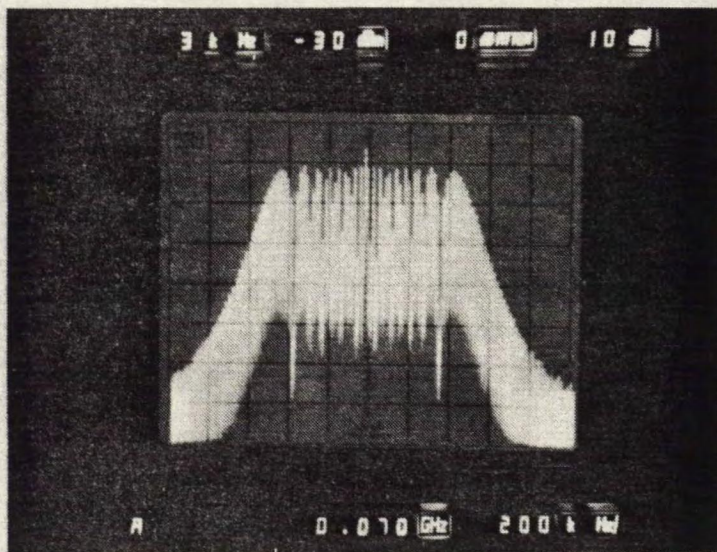
$$C_0/I_0 = -10 \text{ dB}$$

$$f_L = 30 \text{ Hz}$$

$$\Delta f_{PP} = 1 \text{ MHz P-to-P}$$

$$f_c = 10 \text{ kHz}$$

SP-3



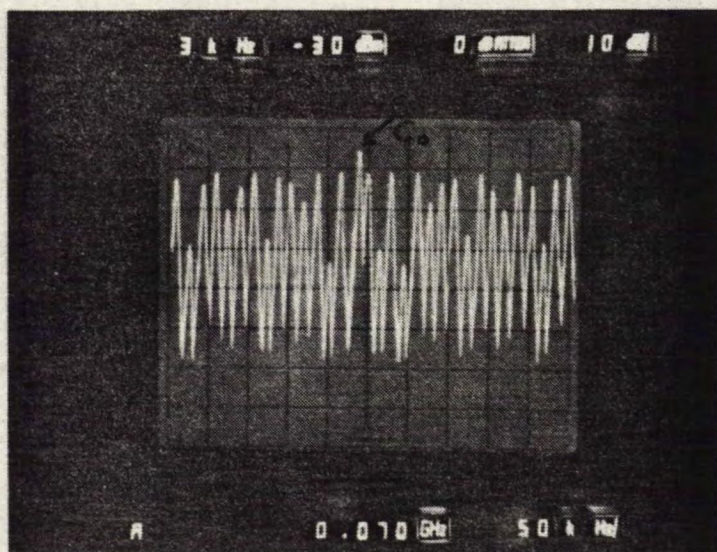
$$C_0/I_0 = -10 \text{ dB}$$

$$f_L = 15.75 \text{ kHz}$$

$$\Delta f_{PP} = 1 \text{ MHz P-to-P}$$

$$f_c = 10 \text{ kHz}$$

SP-4



$$C_0/I_0 = -10 \text{ dB}$$

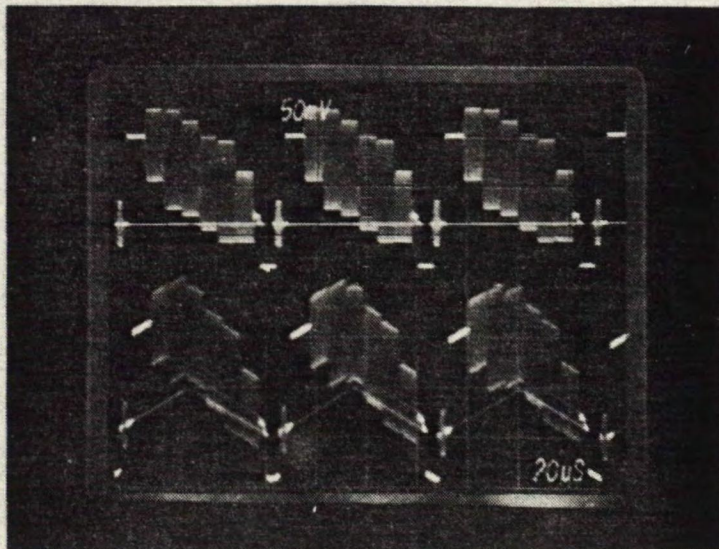
$$f_L = 15.75 \text{ kHz}$$

$$\Delta f_{PP} = 1 \text{ MHz P-to-P}$$

$$f_c = 10 \text{ kHz}$$

SP-5

DISPERSED COLOR BAR VIDEO SIGNALS

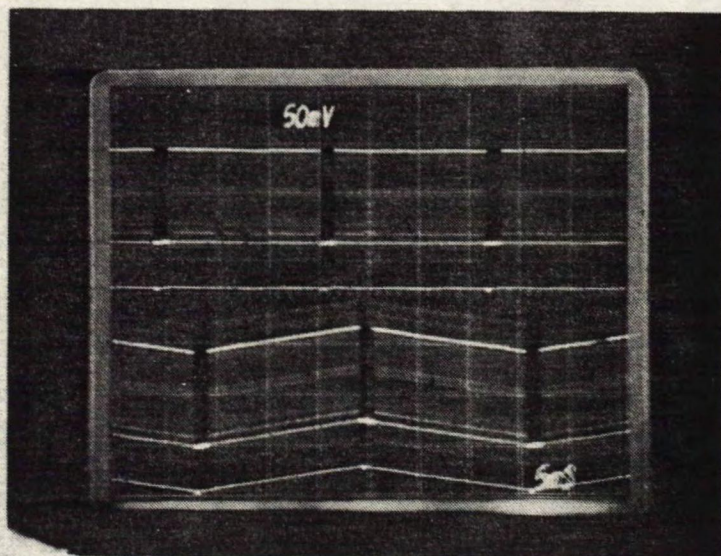


SP-6

FULL FIELD COLOR BAR
SIGNAL DISPLAYED AT
LINE RATE
LEVEL = 165.6 mV P-to-P
NO PRE-EMPHASIS

FULL FIELD COLOR BAR
SIGNAL DISPLAYED AT
FIELD RATE AND
DISPERSED WITH TRIANGULAR
W/F. $f_L = 15.75 \text{ kHz}$
 $\Delta f_{PP} = 2 \text{ MHz}$ P-to-P

NO PRE-EMPHASIS



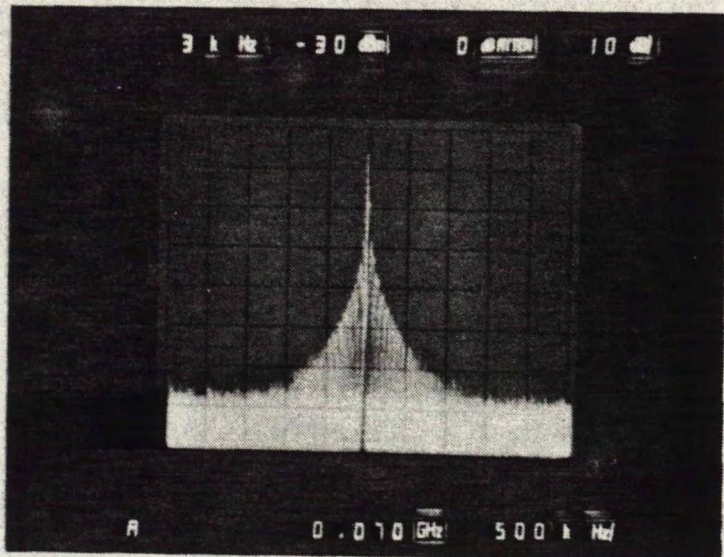
SP-7

FULL FIELD COLOR BAR
SIGNAL DISPLAYED AT
FRAME RATE
LEVEL = 165.6 mV P-to-P
NO PRE-EMPHASIS

FULL FIELD COLOR BAR
SIGNAL DISPLAYED AT
FRAME RATE AND
DISPERSED WITH TRIANGULAR
W/F. $f_L = 30 \text{ Hz}$
 $\Delta f_{PP} = 2 \text{ MHz}$ P-to-P

NO PRE-EMPHASIS

RADAR-LIKE SIGNAL



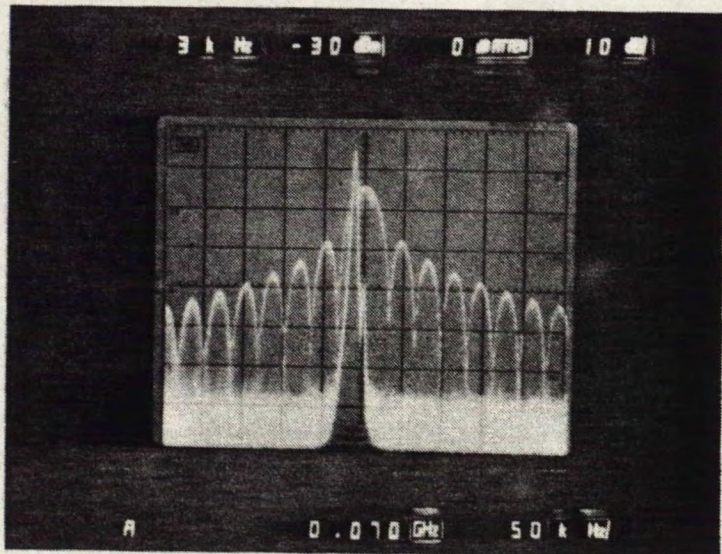
SP-8

$$C_0/I_P = -10 \text{ dB}$$

$$f_c = 10 \text{ kHz}$$

$$PW = 30 \mu\text{s}$$

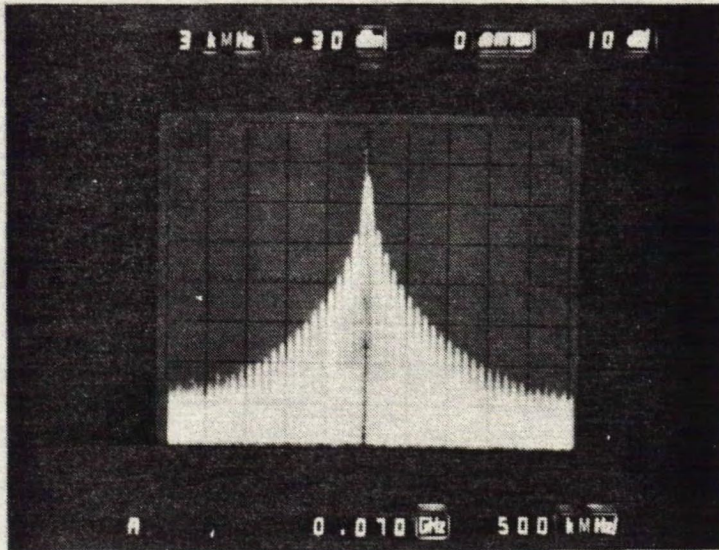
$$PRF = 0.3 \text{ kHz}$$



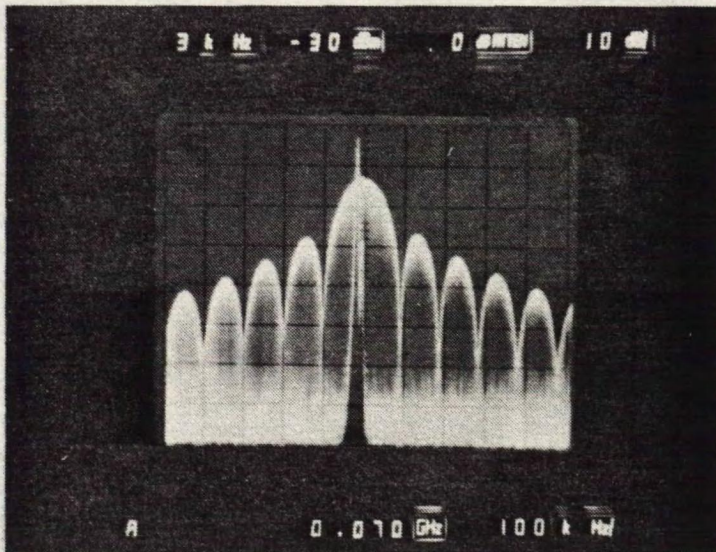
SP-9

Note: 1st side-band peak is 14 dB below main peak

RADAR - LINE SIGNAL



SP-10



SP-11

$$C_0/I_p = -20\text{dB}$$

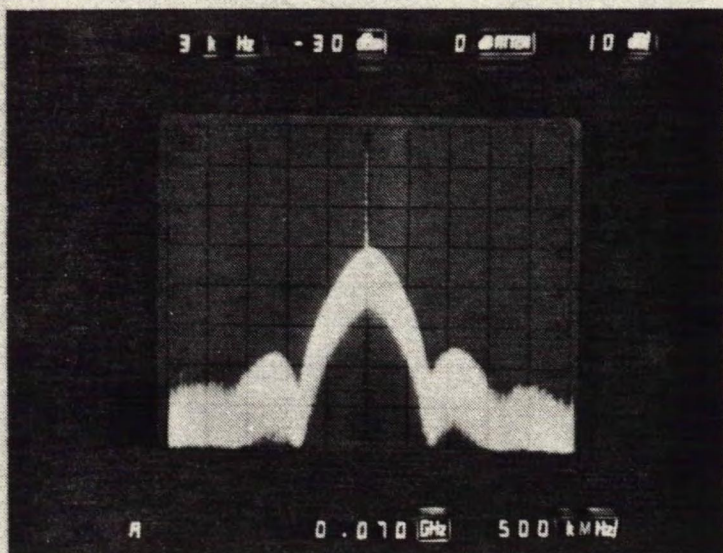
$$f_c = 10\text{ kHz}$$

$$PW = 10\text{ }\mu\text{s}$$

$$PRF = 1.0\text{ kHz}$$

RADAR-LIKE SIGNAL

29



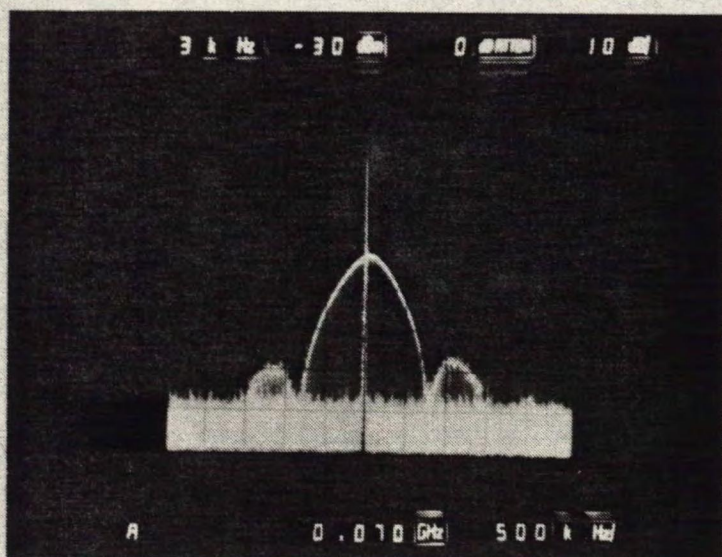
$$C_0/I_p = -20 \text{ dB}$$

$$f_c = 10 \text{ kHz}$$

$$PW = 1 \text{ } \mu\text{sec}$$

$$PRF = 3 \text{ kHz}$$

SP-12



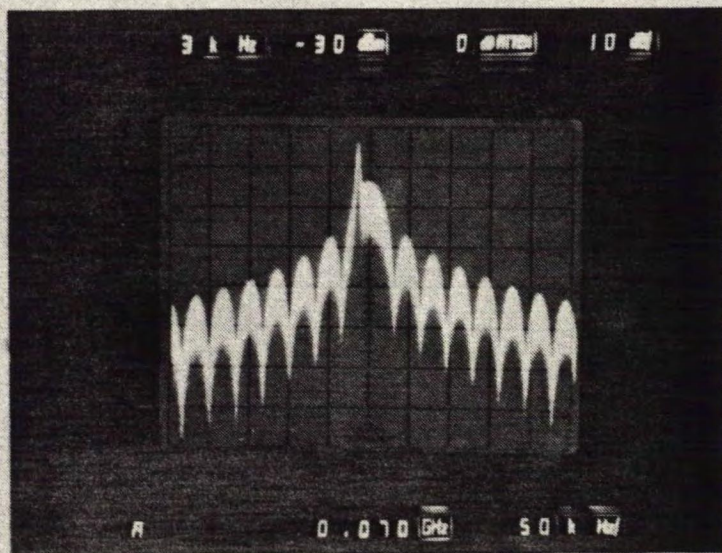
$$C_0/I_p = -20 \text{ dB}$$

$$f_c = 10 \text{ kHz}$$

$$PW = 1 \text{ } \mu\text{sec}$$

$$PRF = 0.3 \text{ kHz}$$

SP-13



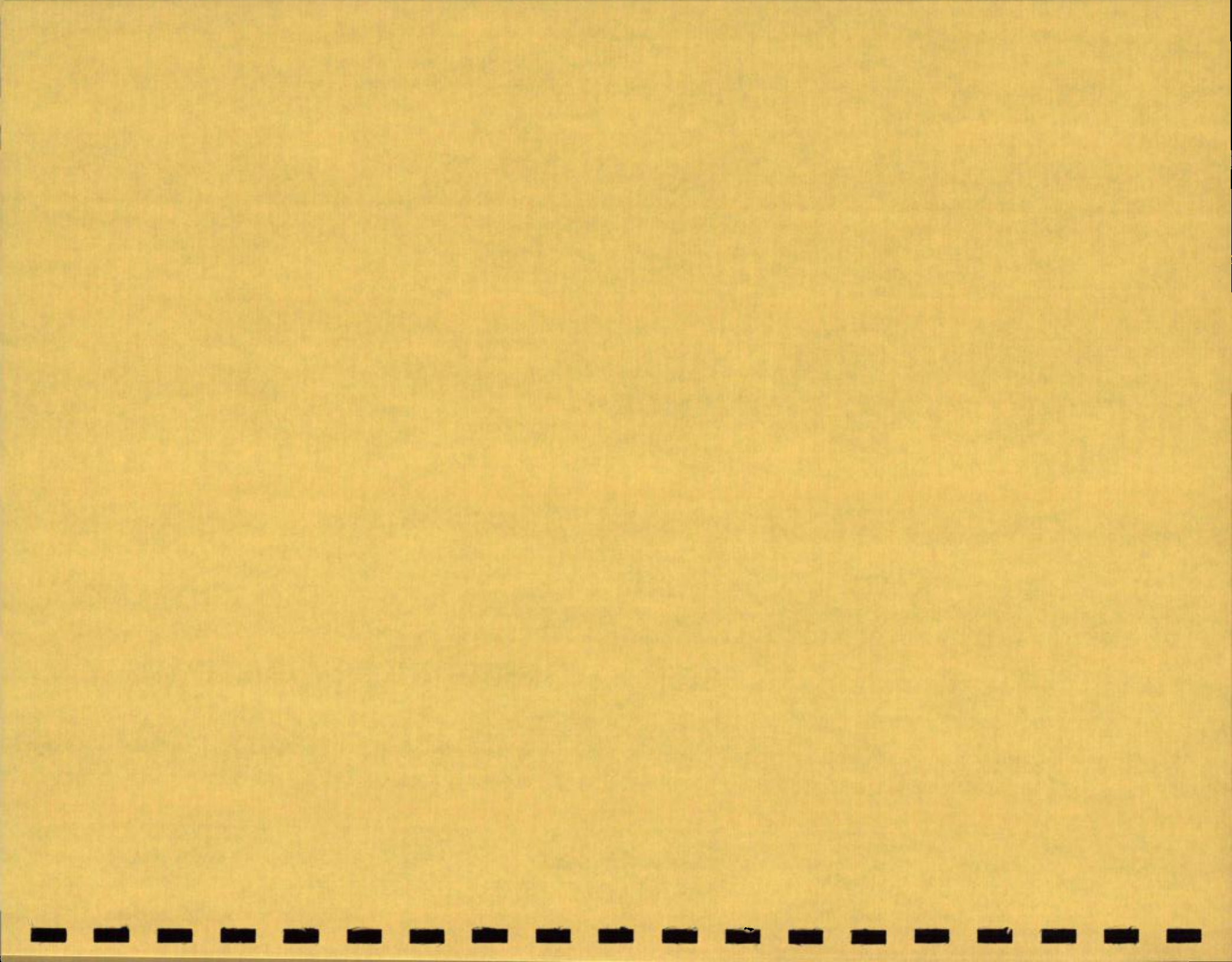
$$C_0/I_p = -10 \text{ dB}$$

$$f_c = 10 \text{ kHz}$$

$$PW = 30 \text{ } \mu\text{sec}$$

$$PRF = 3 \text{ kHz}$$


SP-14



A1

BIT ERROR RATE (BER) VS CARRIER LEVEL (C)

NO THERMAL NOISE AND NO INTERFERING CARRIER

CARRIER LEVEL (C) (dBm)	BER	
-35	1.23×10^{-7}	
-37	NO ERROR OBSERVED	
-39		
-41		
-43 (NOTE 1)		
-45		
-47		
-49		
-51		
-53		
-54		NO ERROR OBSERVED
-55		LOSS OF DATA

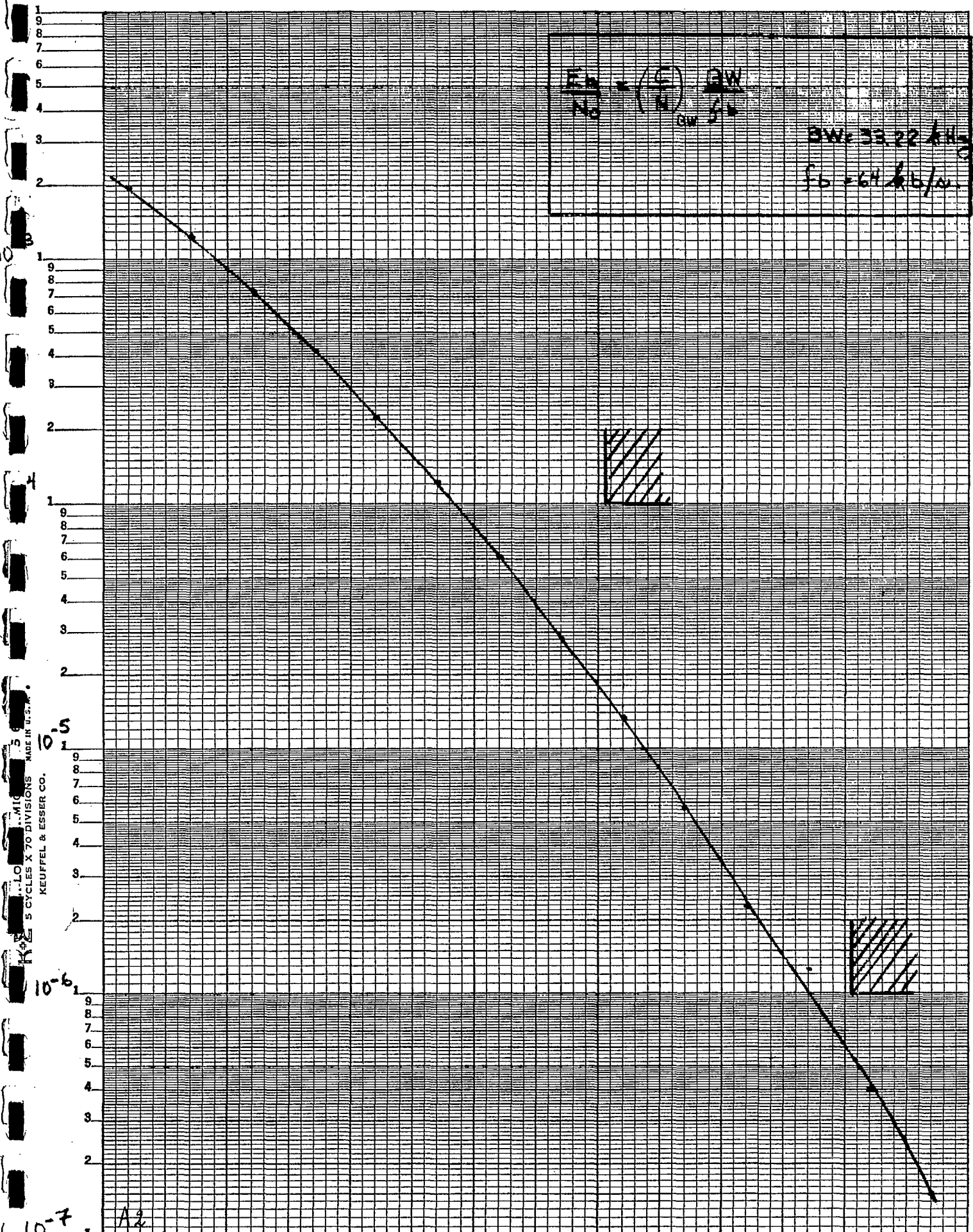
NOTE 1 : NOMINAL CARRIER LEVEL (C₀)

A2

DET

$$\frac{F_b}{N_b} = \left(\frac{F}{N} \right) \frac{BW}{f_b}$$

BW = 33.22 kHz
 $f_b = 64 \text{ kb/s}$

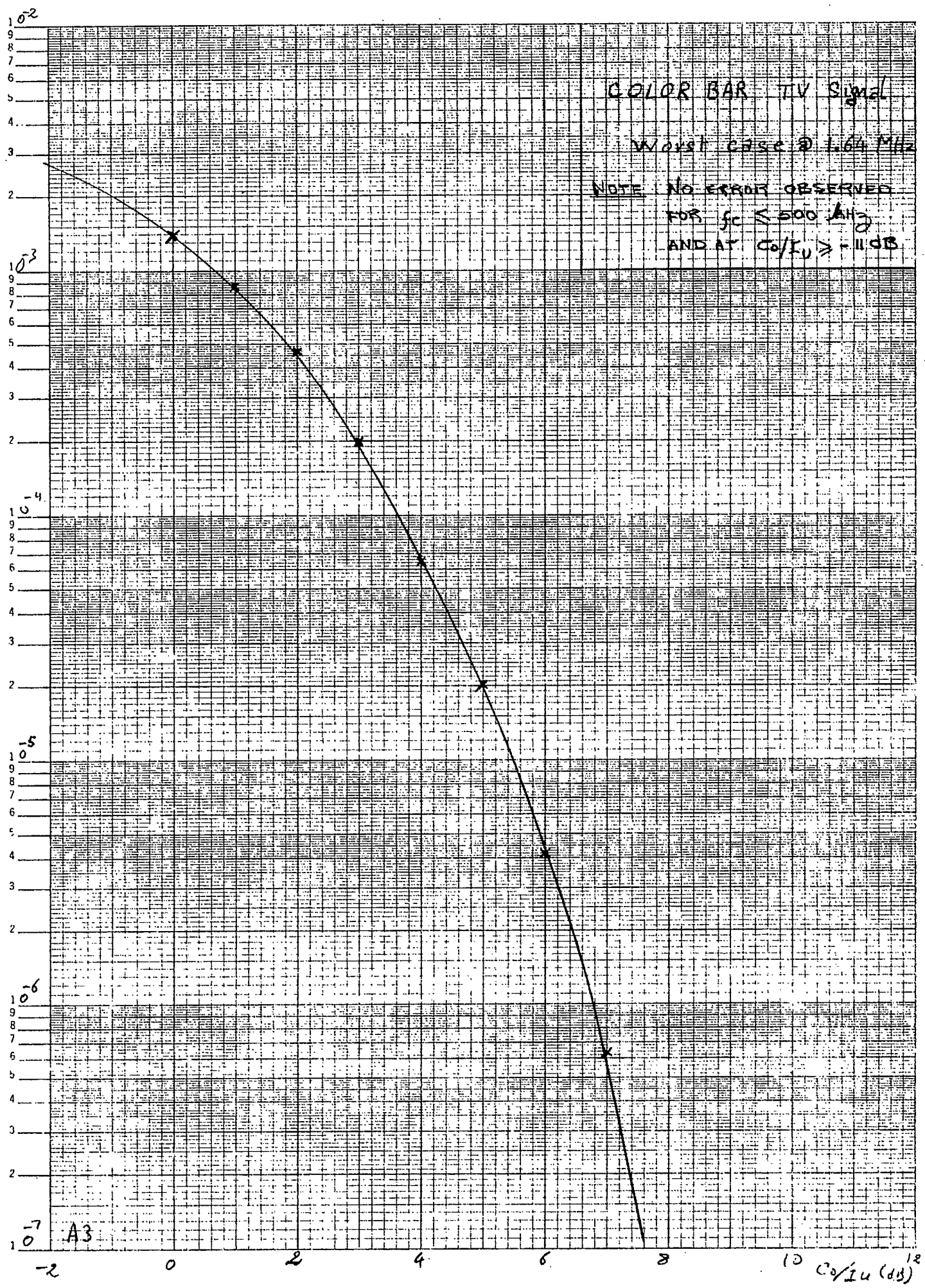


5
 5 CYCLES X 70 DIVISIONS MADE IN U.S.A.
 KEUFFEL & ESSER CO.

10 7.15 11 8.15 12 9.15 13 10.15 14 11.15 15 12.15 16 13.15 C/N Eb/No

A3

DCR



A3

A4

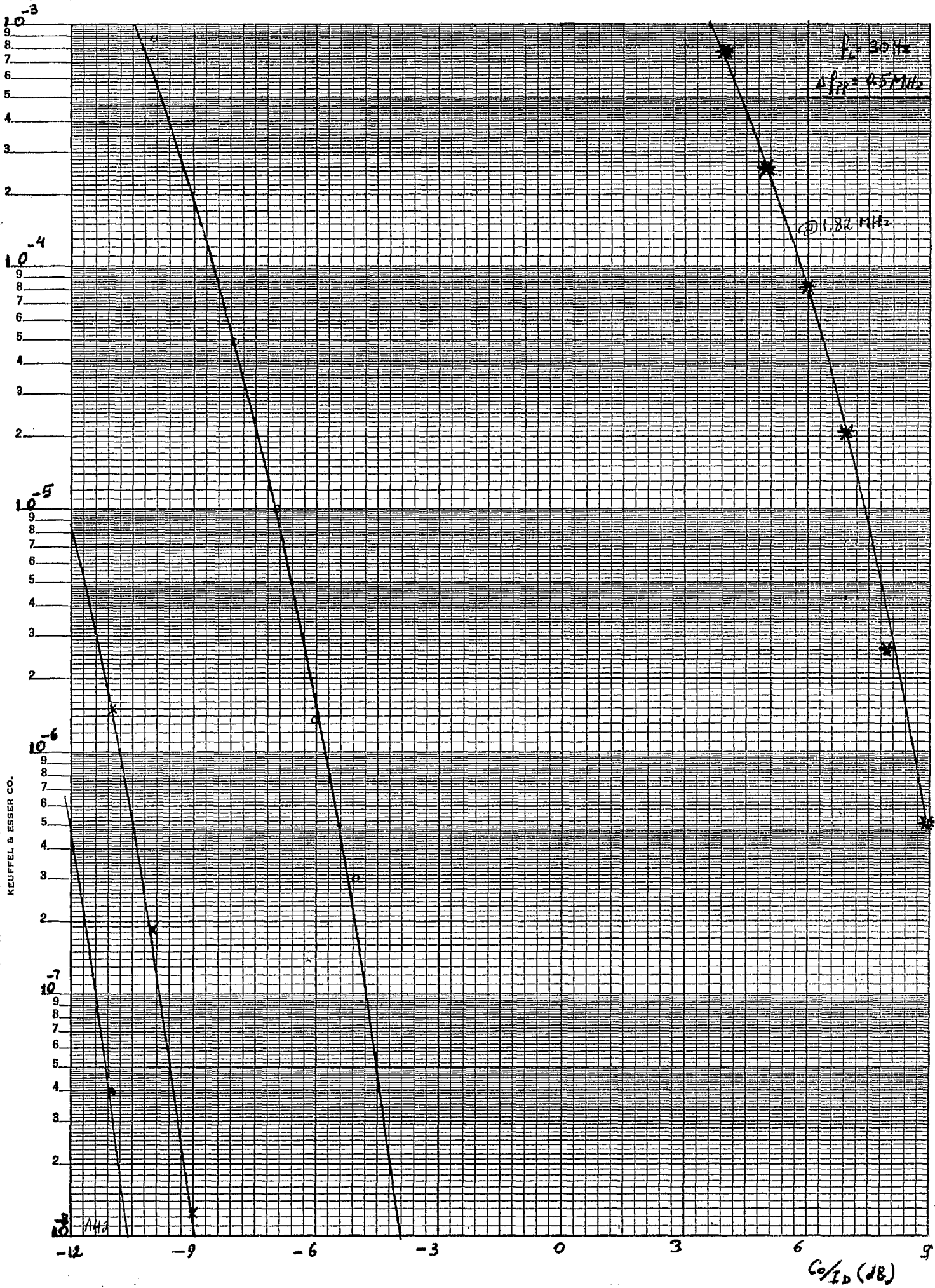
LEGEND:

*	WORST CASE f_c
O	$f_c = 3.6 \text{ MHz}$
X	$f_c = 3 \text{ MHz}$
•	$f_c = 2 \text{ MHz}$
Δ	$f_c = 1 \text{ MHz}$
□	$f_c = 500 \text{ kHz}$
▲	$f_c = 100 \text{ kHz}$
■	$f_c = 22.5 \text{ kHz}$

I_D modulated by Color bar signal

$C_0/N \rightarrow \infty$

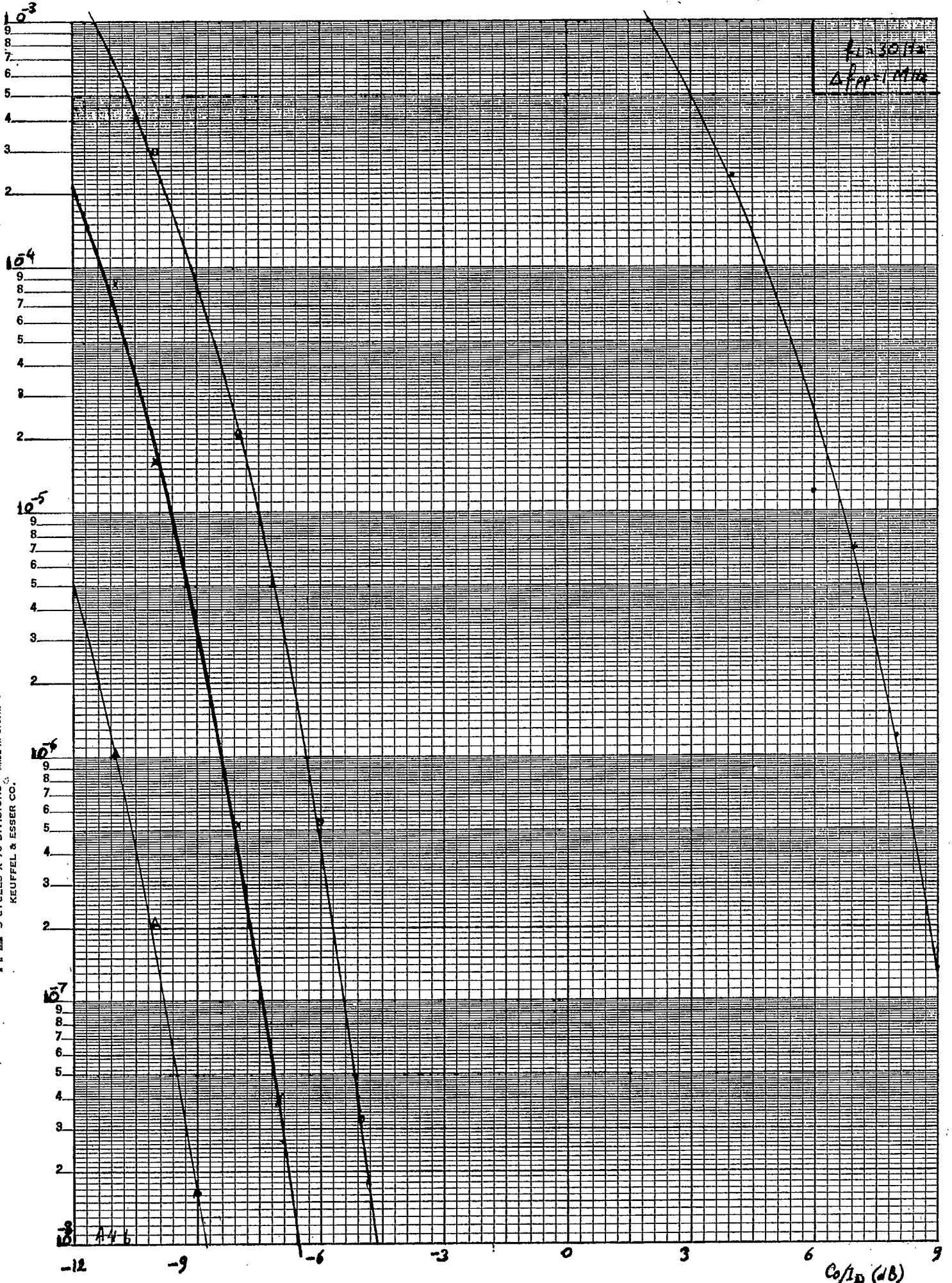
BER



SEMILOGARITHMIC
5 CYCLES X 70 DIVISIONS
46 6213
KEUFFEL & ESSER CO.
MADE IN U.S.A.

BEN

4-30-72
A46

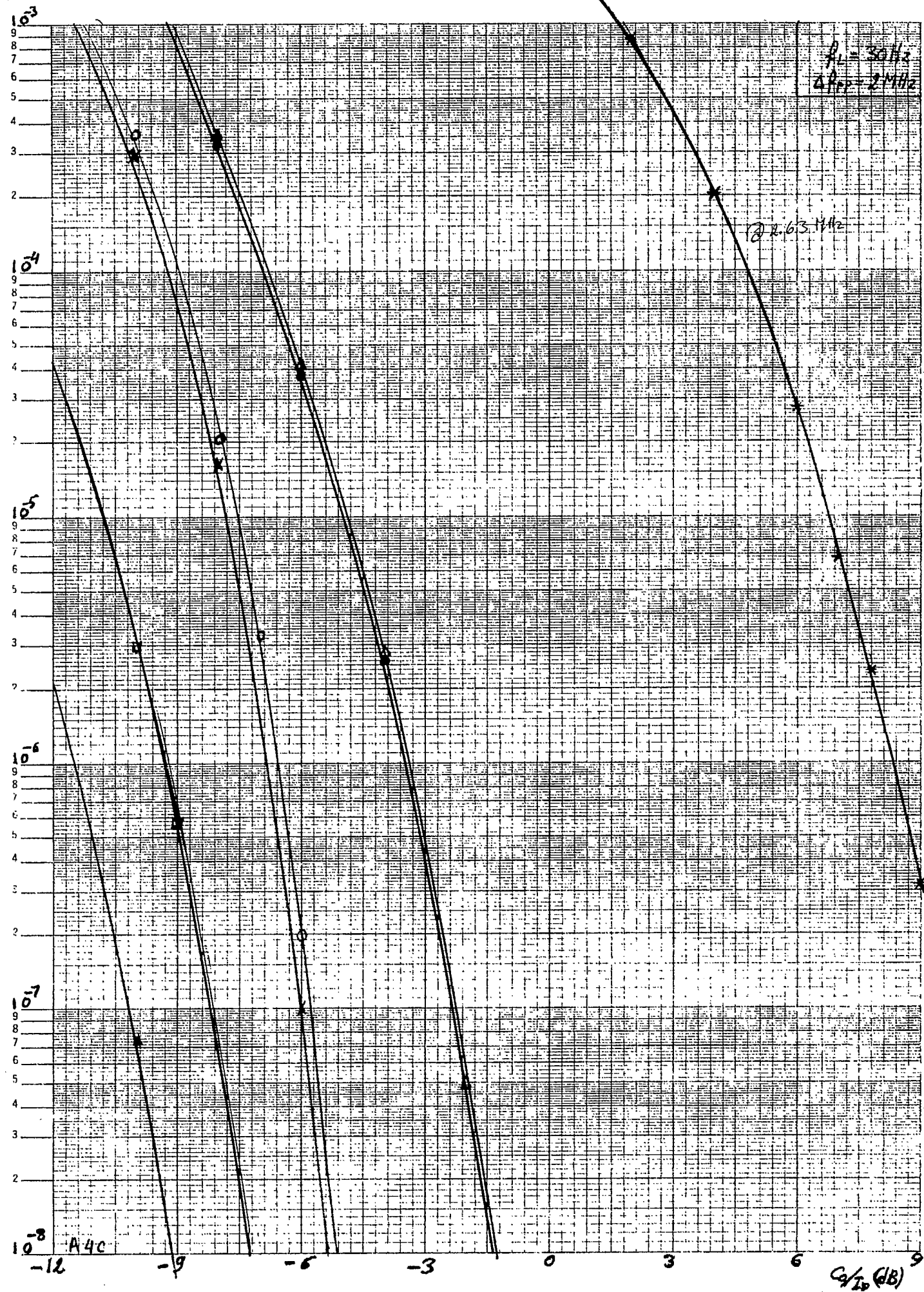


LOG-LOG PAPER
5 CYCLES X 70 DIVISIONS
MADE IN U.S.A.
KEUFFEL & ESSER CO.

A46

Co/10 (dB)

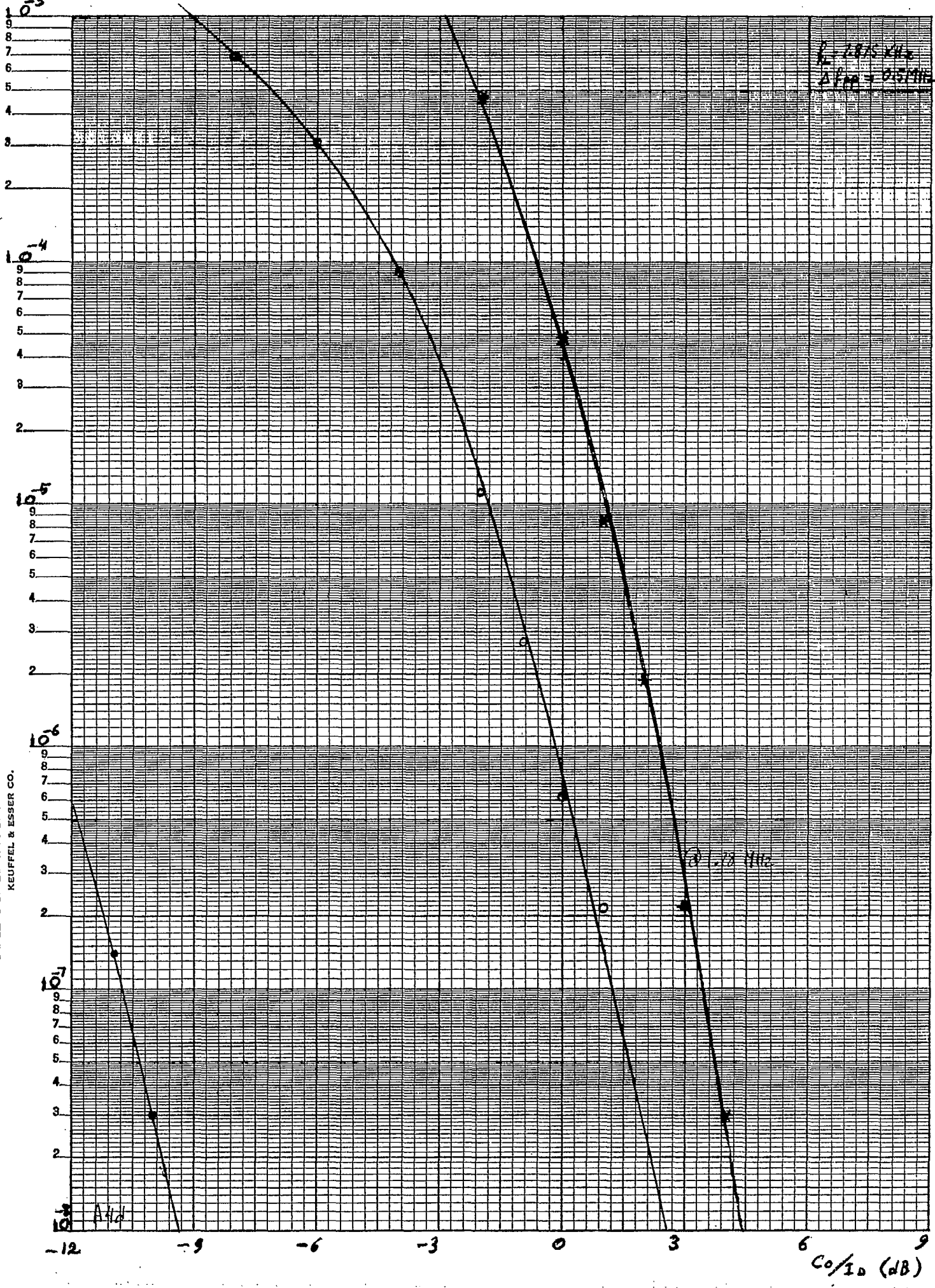
BER



SEAFAR'S COPY X 100% SPECIFY TRACING OF DRAWING PAPER
MADE IN CANADA

BER

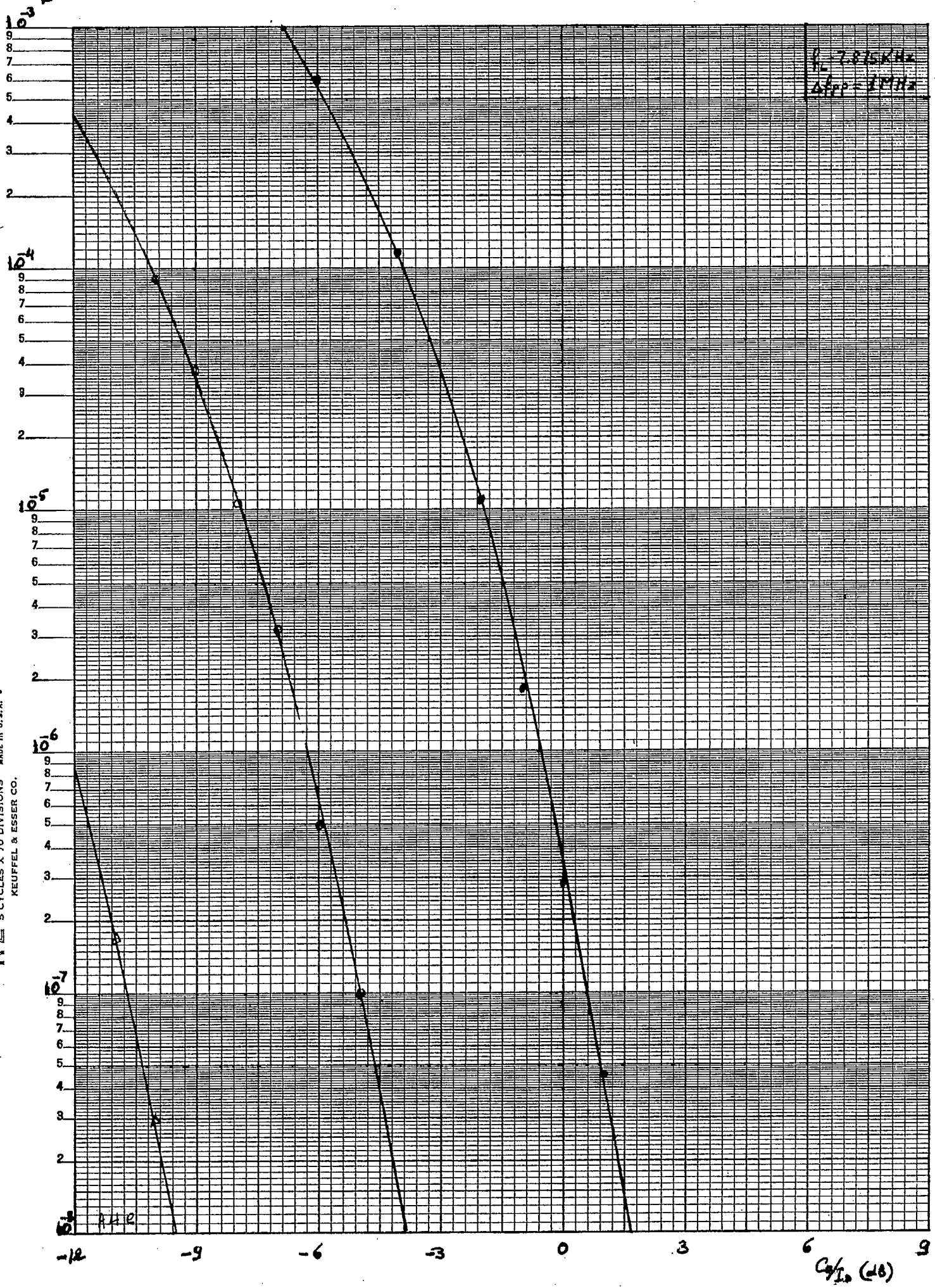
$f_c = 2075 \text{ kHz}$
 $\Delta f_{\text{mod}} = 0.5 \text{ MHz}$



LOG PERMITS 6666
5 CYCLES X 70 DIVISIONS
KEUFFEL & ESSER CO.

BER

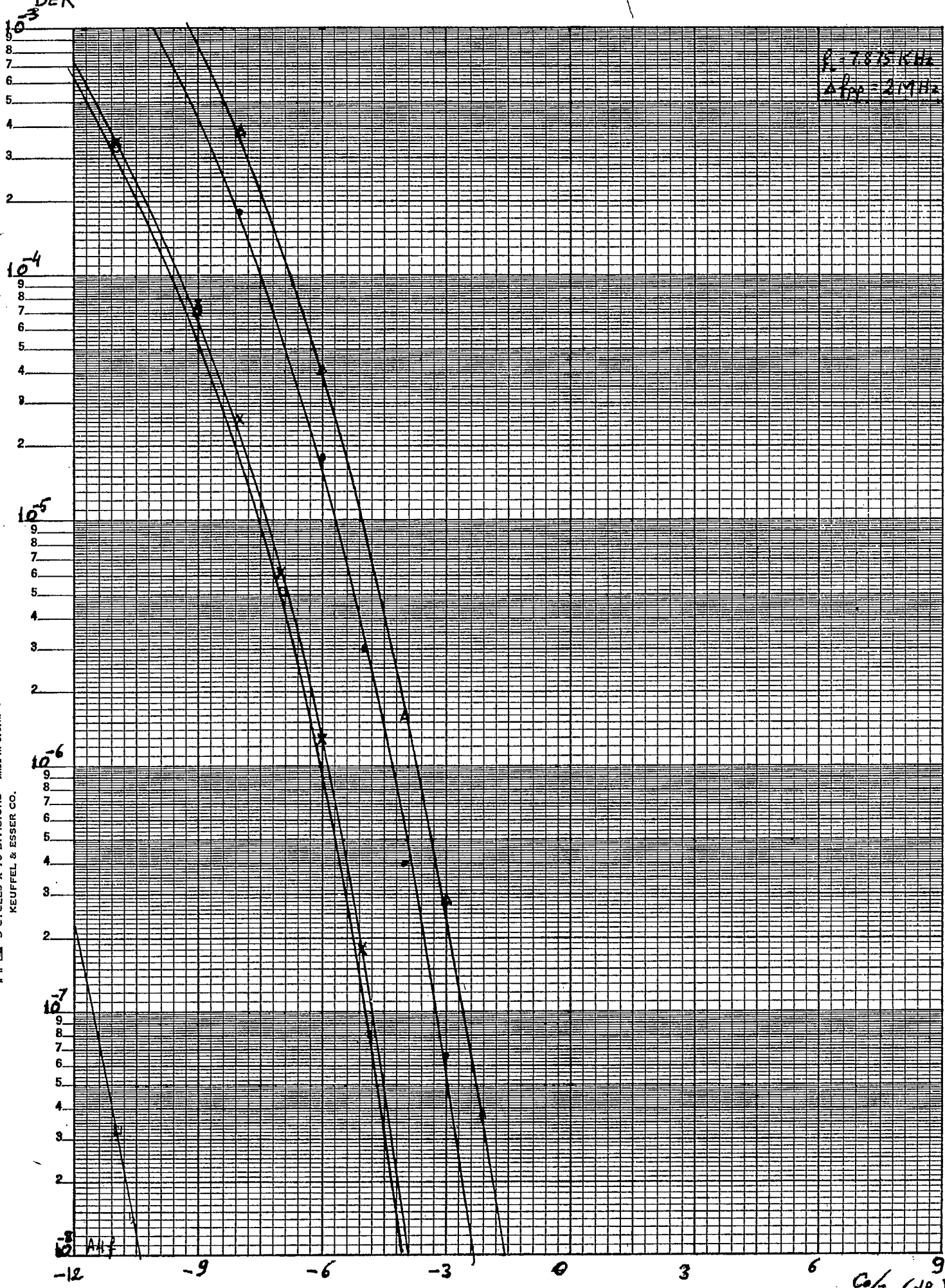
$f_c = 7.815 \text{ kHz}$
 $\Delta f_{pp} = 1 \text{ MHz}$



KEUFFEL & ESSER CO.
SEMI-LOGARITHMIC
5 CYCLES X 70 DIVISIONS
MADE IN U.S.A.

BER

$f_c = 7.875 \text{ MHz}$
 $\Delta f_{\text{mod}} = 2 \text{ MHz}$

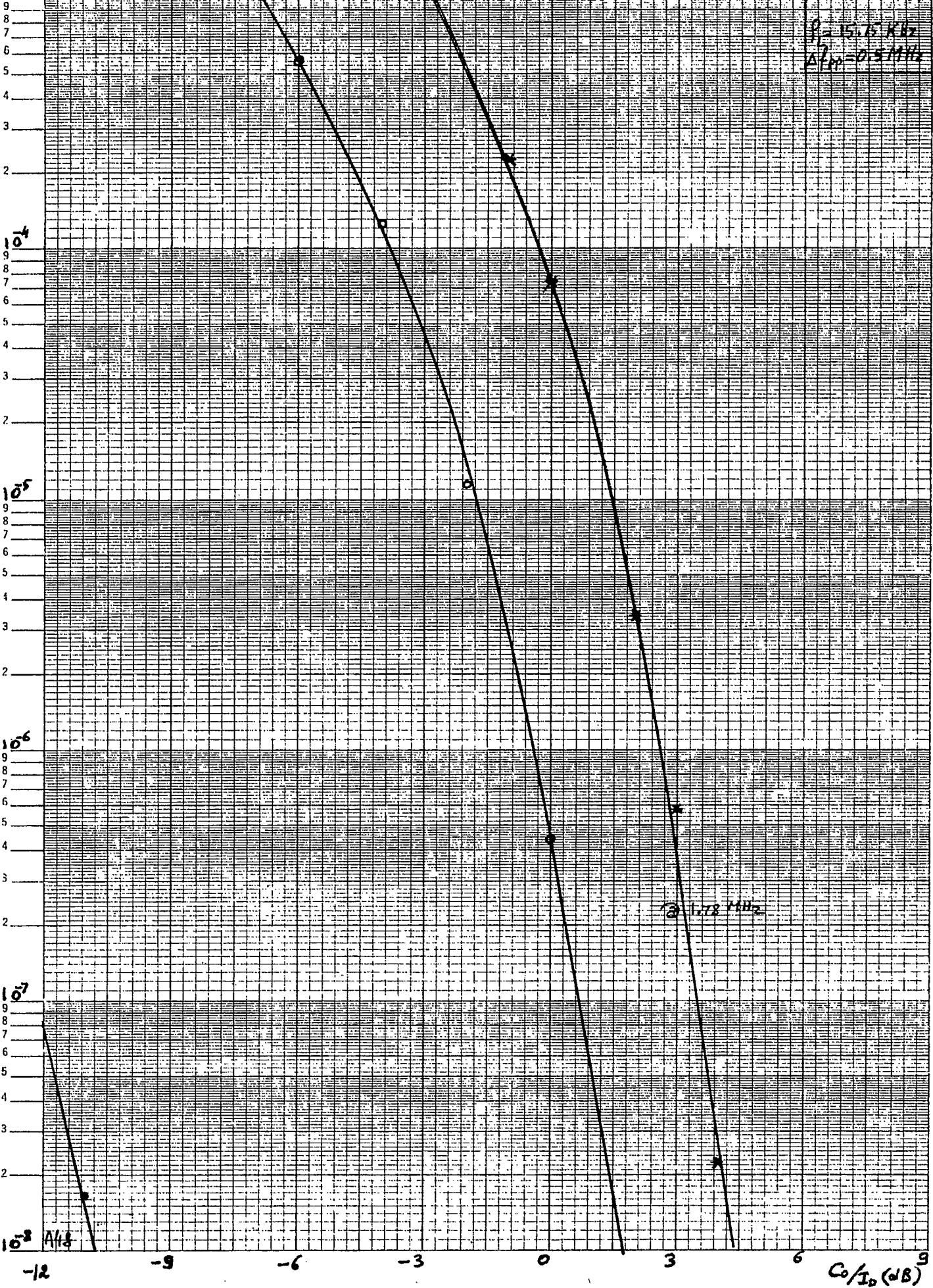


LOG PERIODIC
5 CYCLES X 70 DIVISIONS
MADE IN U.S.A.
KEUFFEL & ESSER CO.

C/I_0 (dB)

BER

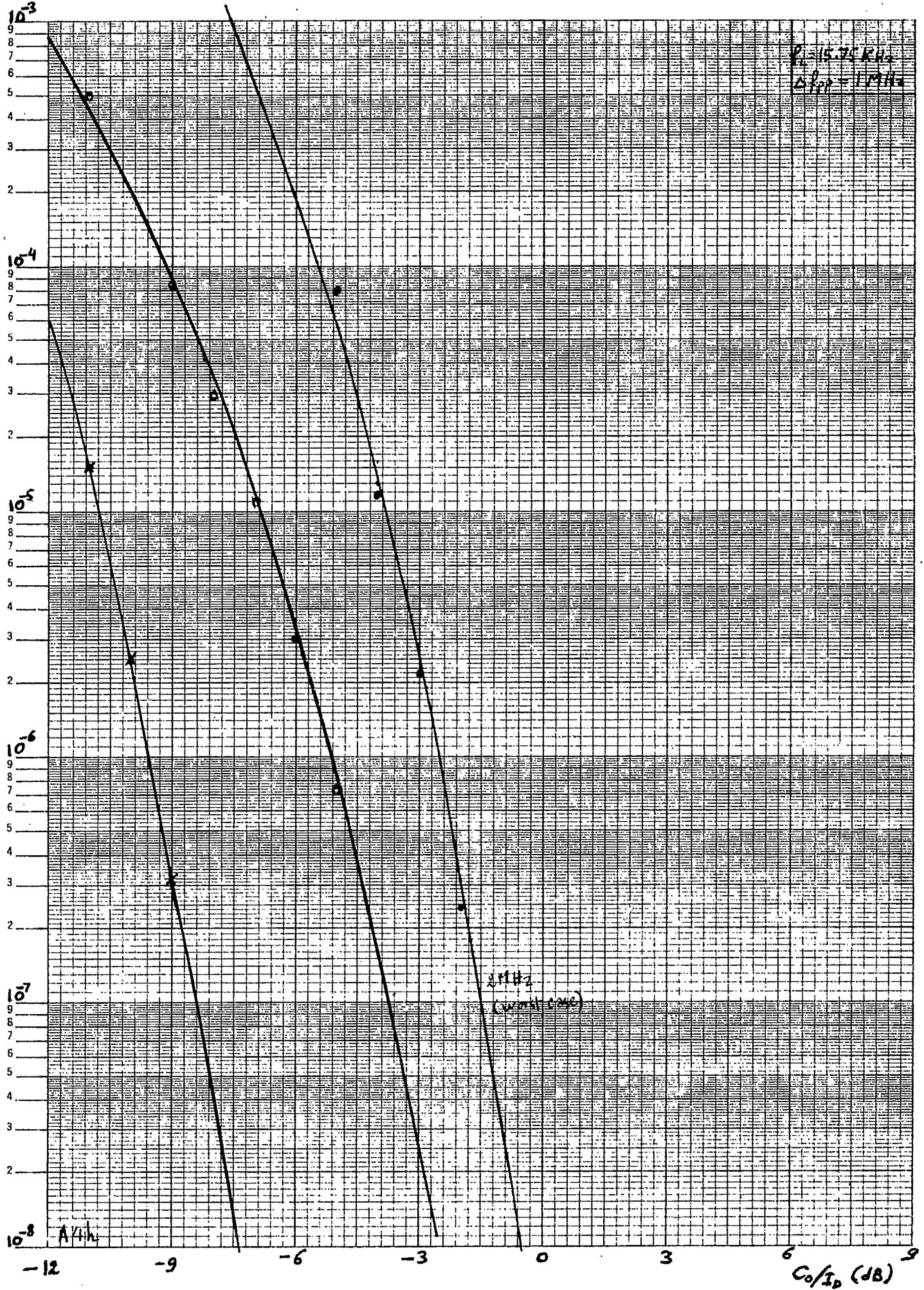
10^3



C_0/I_0 (dB)

BER

$P_r = 15.7 \text{ dBm}$
 $\Delta f_{IF} = 1 \text{ MHz}$



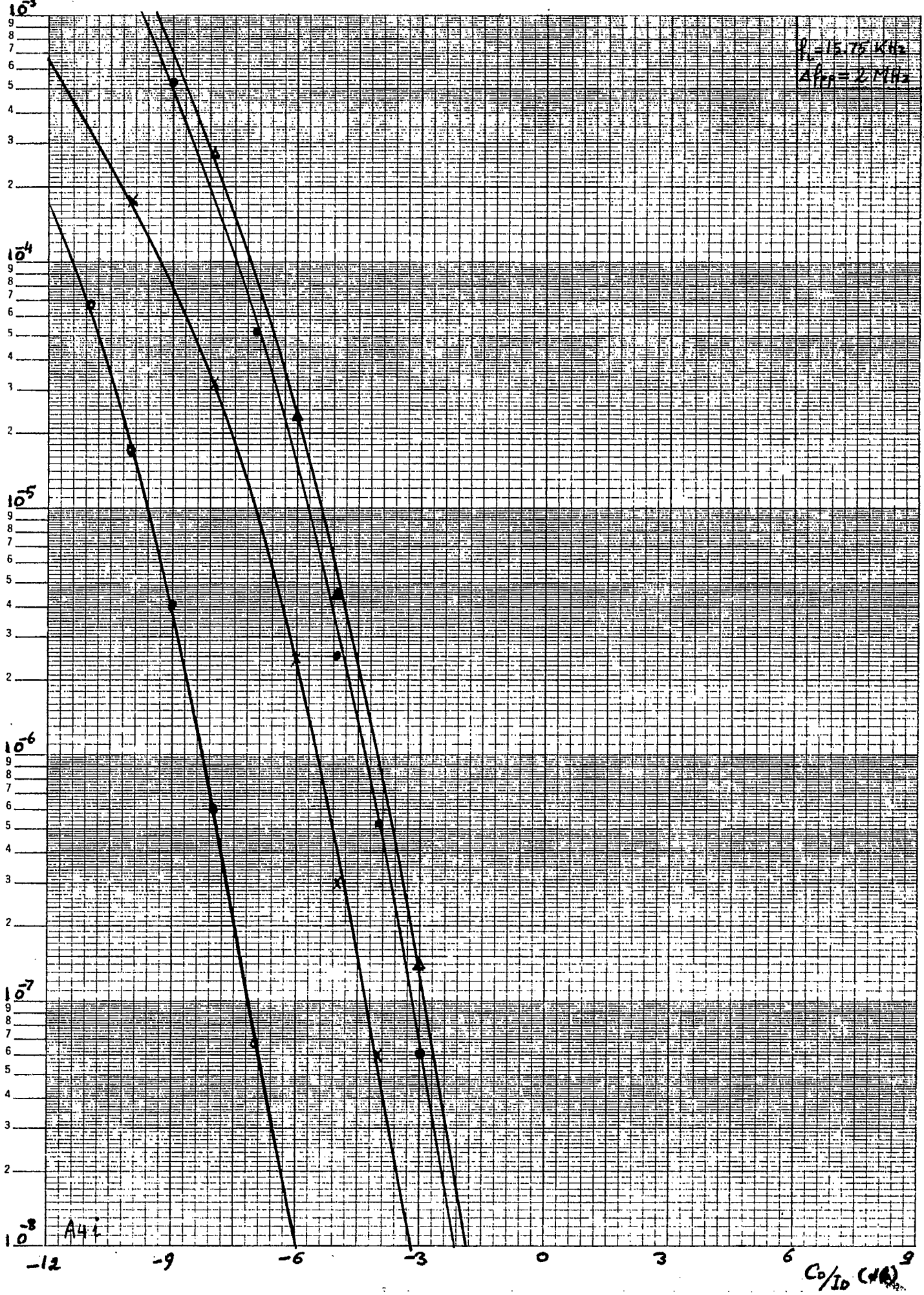
SEHLLASAR... 5... X 10... IN...
SR... MADE...
TRA... OR...
TRA... OR...

A/4h

2 MHz
(worst case)

C_0/I_f (dB)

BER



A5

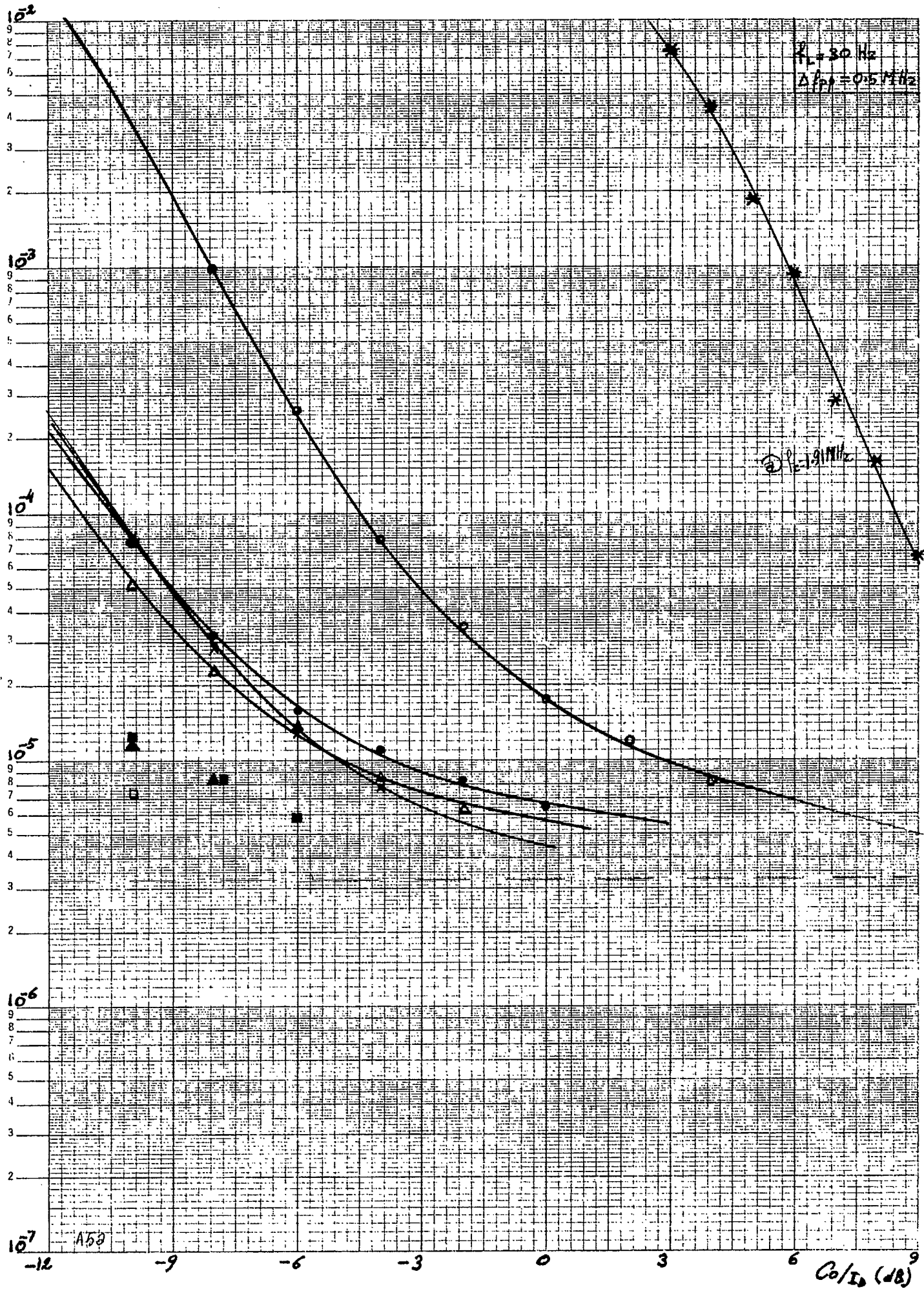
LEGEND:

SYMBOL	f_c
*	WORST CASE
o	3.6 MHz
x	3.0
•	2.0
Δ	1.0 MHz
□	500 kHz
▲	100 kHz
■	22.5 kHz

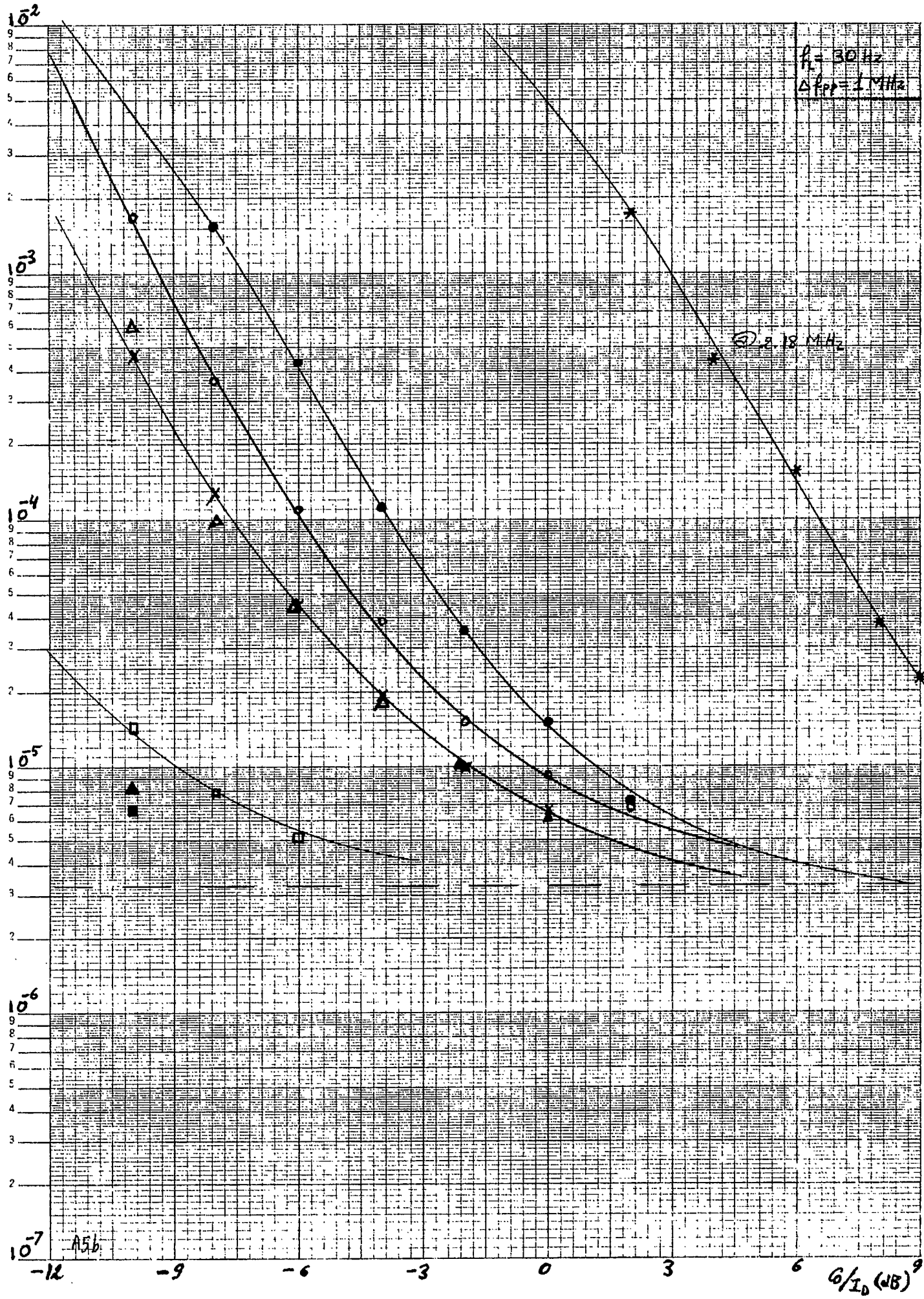
I_p modulated by color bar signal

$C_0/N = 15 \text{ dB}$

BER



BER



BER

10^{-2}

$B_p = 30 \text{ Hz}$
 $\Delta f_{pp} = 2 \text{ MHz}$

10^3

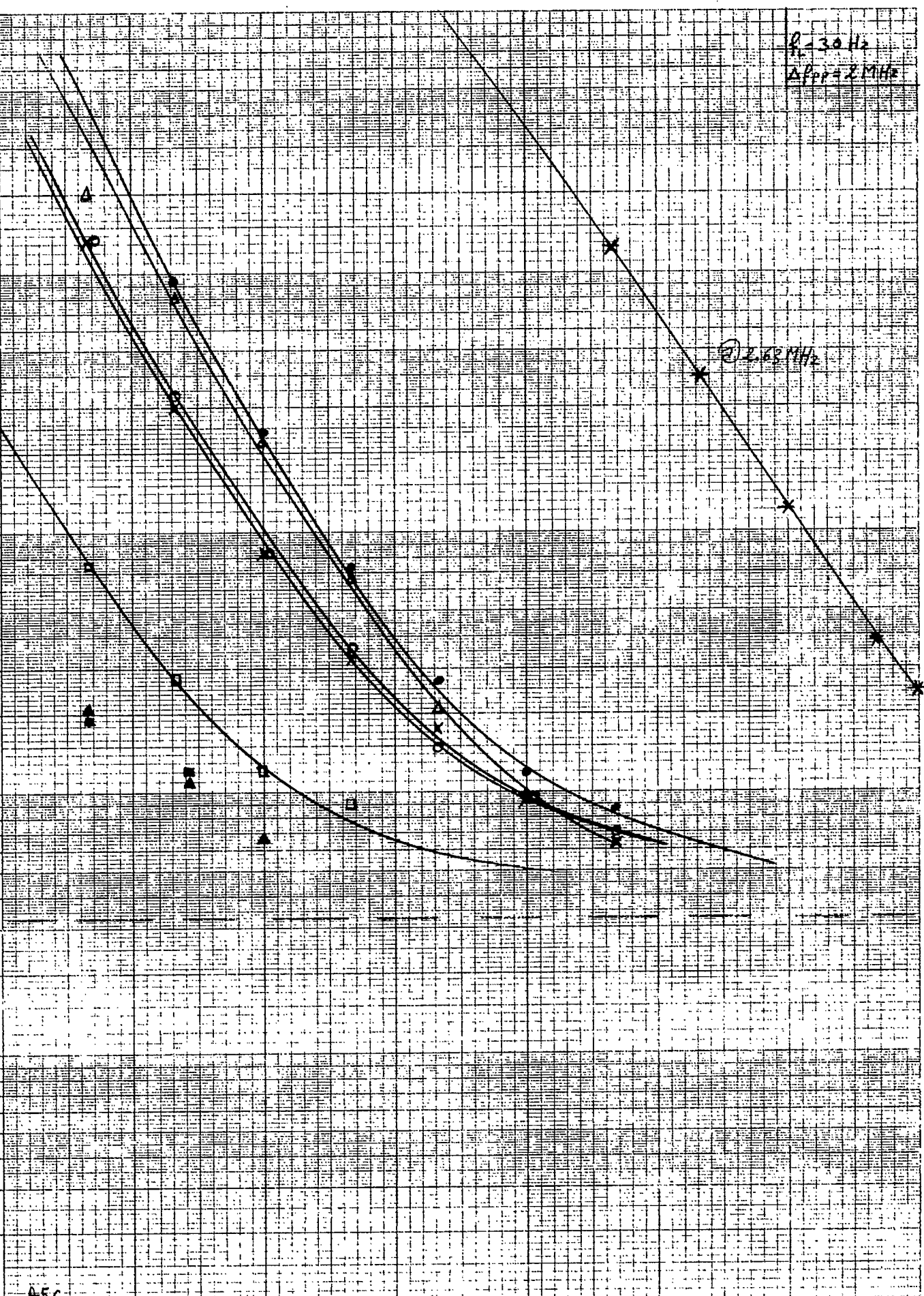
10^4

10^5

10^6

10^7

SEMPER PARITURUS CYCLES A 10 INCHES IN
SPECIFY TRACING OR DRAWING PAPER
SHAPING INSTRUMENTS MADE IN CANADA



A5c

C/I_0 (dB)

BER

10^2

$f_c = 7.975 \text{ kHz}$
 $\Delta f_{IF} = 0.5 \text{ MHz}$

10^3

1.88 MHz

10^4

10^5

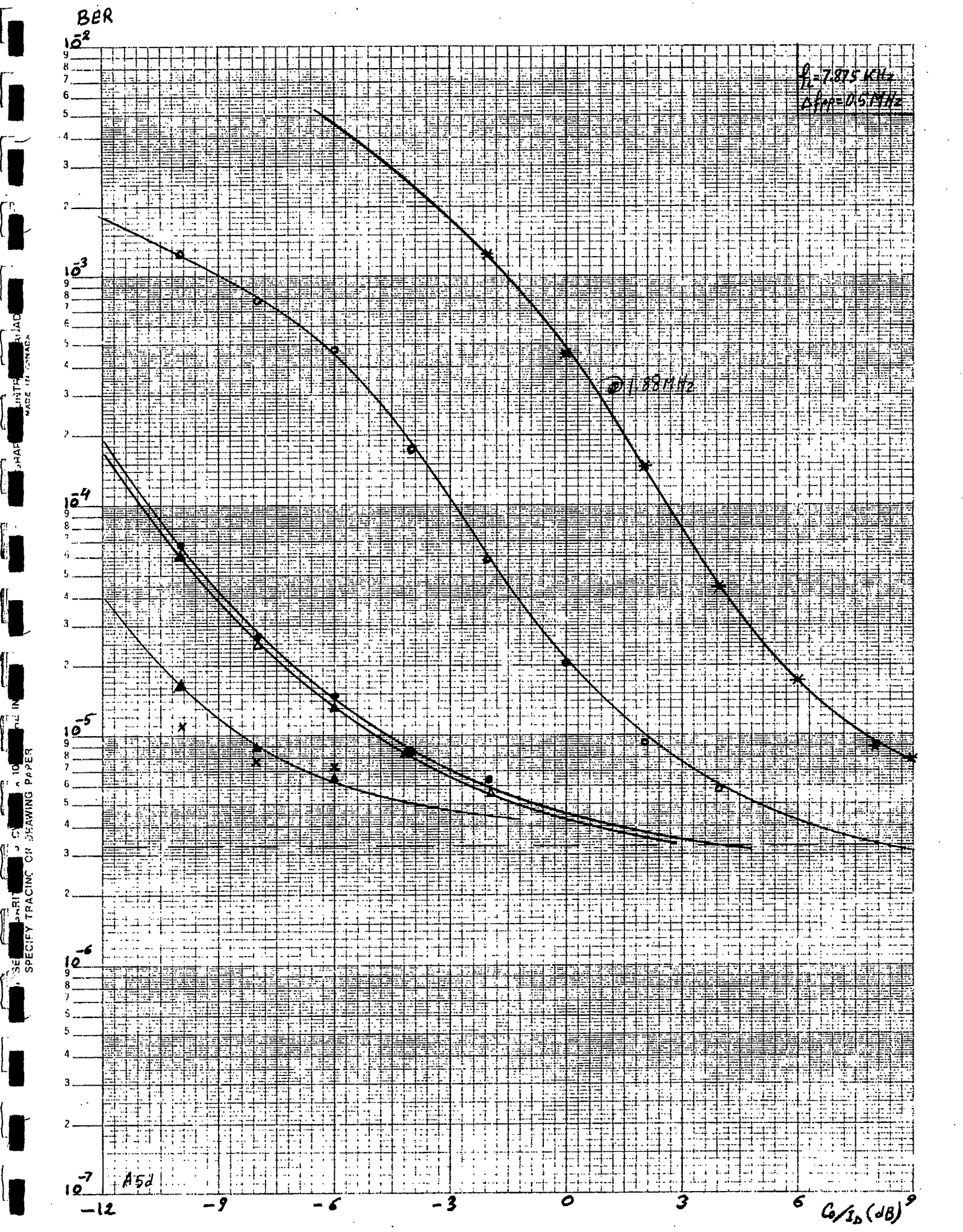
10^6

10^7

A5d

-12 -9 -6 -3 0 3 6 $C_0/S_0 \text{ (dB)}$

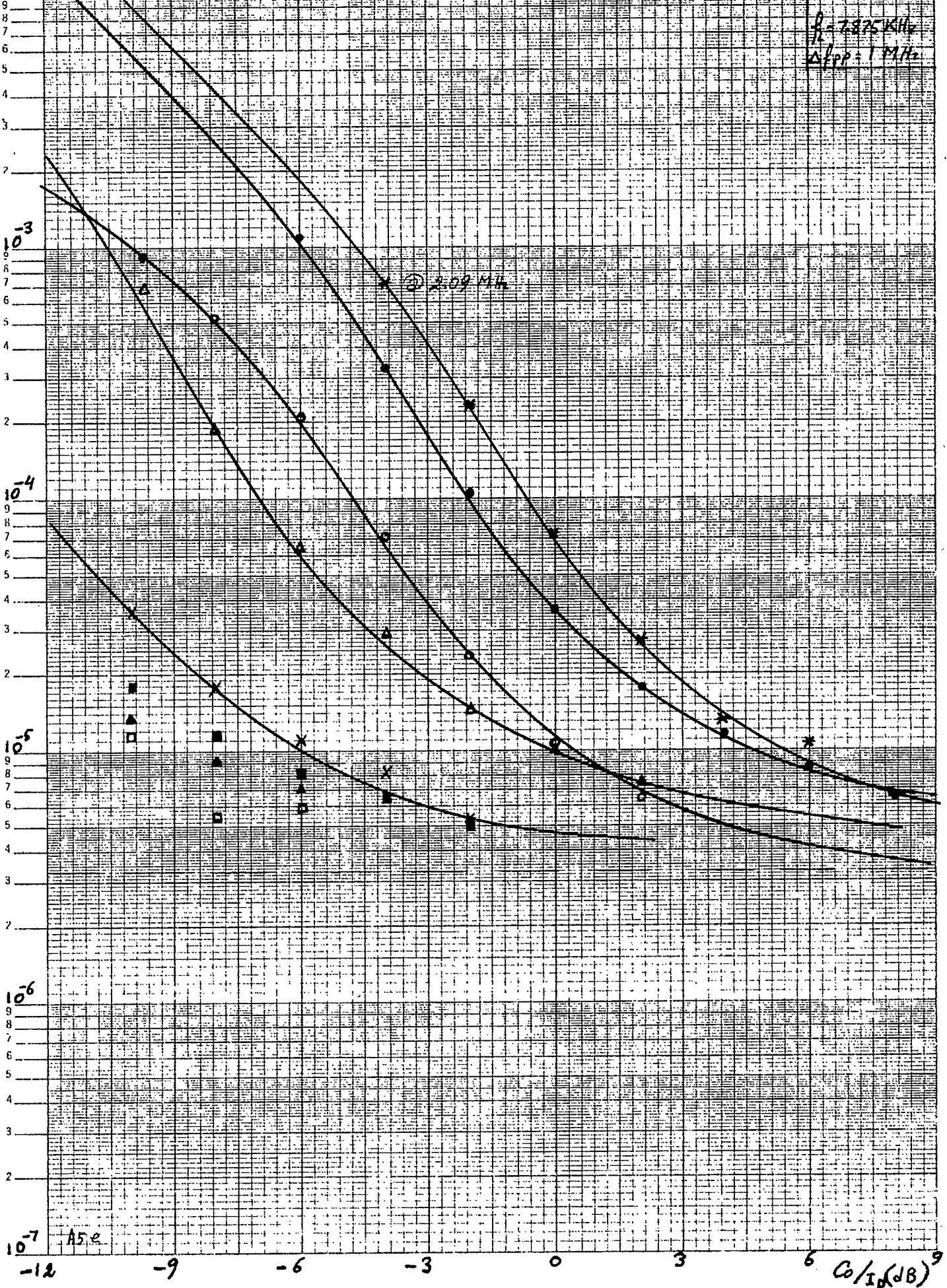
MADE IN CANADA
SPECIFY TRACING OR DRAWING PAPER



BER

10^{-2}

$f_c = 7.875 \text{ MHz}$
 $\Delta f_{pp} = 1 \text{ MHz}$



UNITED STATES OF AMERICA
MADE IN CANADA

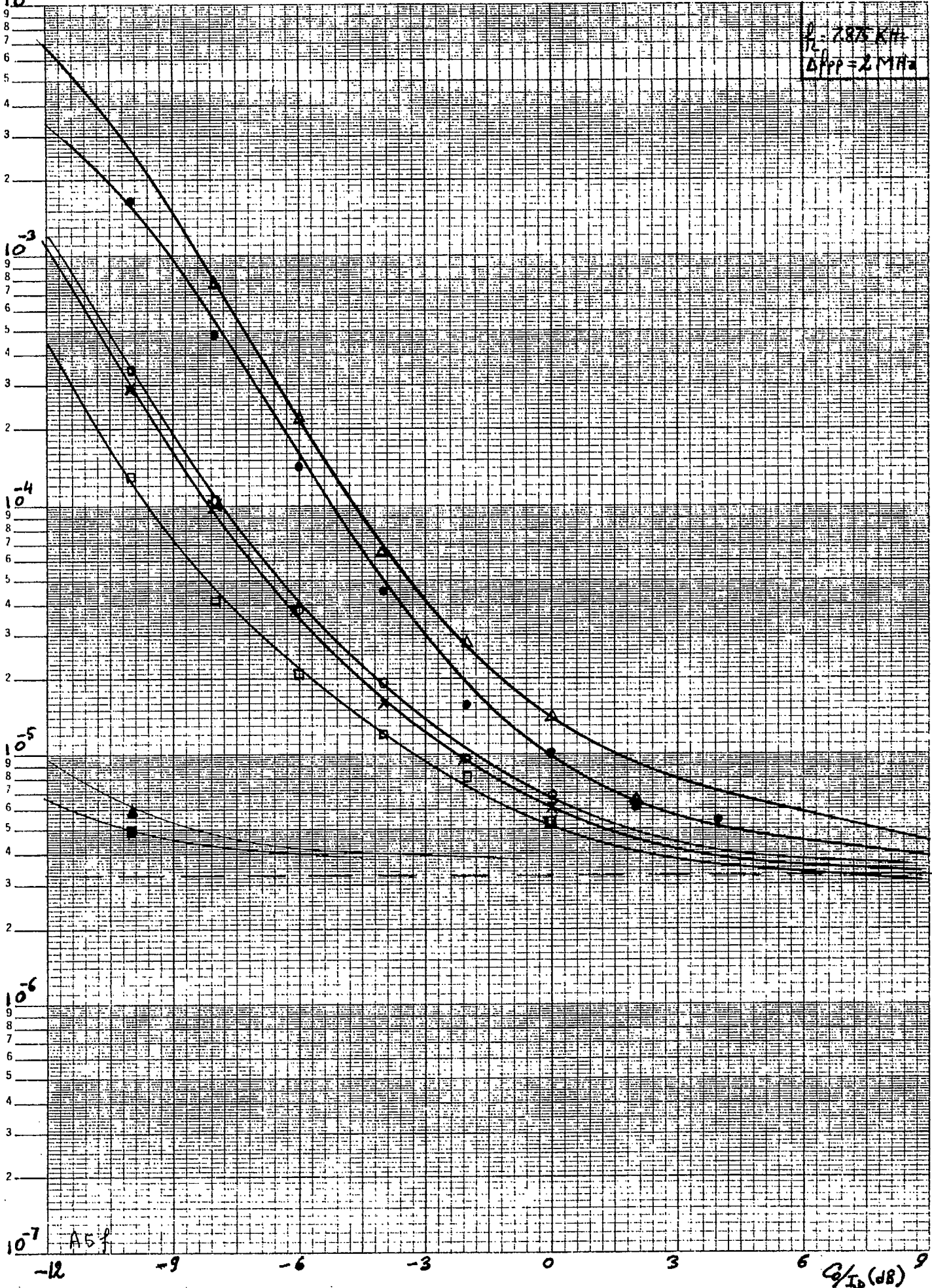
FOR REPRODUCTION OR DRAWING PURPOSES
SPECIFY TRACING OR DRAWING PAPER

10^{-7} $1.5e$

C_0/I_0 (dB)

$f_c = 2875 \text{ kHz}$
 $\Delta f_{pp} = 2 \text{ MHz}$

BER



SEMICONDUCTOR TRACING OR DRAWING PAPER
10 INCHES
AD
DE IN

10^{-7} A5f

C/I_0 (dB)

BER

10^{-2}

$f_c = 15.75 \text{ kHz}$
 $\Delta f_{pp} = 0.5 \text{ MHz}$

10^{-3}

21772 MHz

10^{-4}

10^{-5}

10^{-6}

10^{-7} A53

SE...PT...C...
SPECIFY TRACING OR DRAWING PAPER

-12

-9

-6

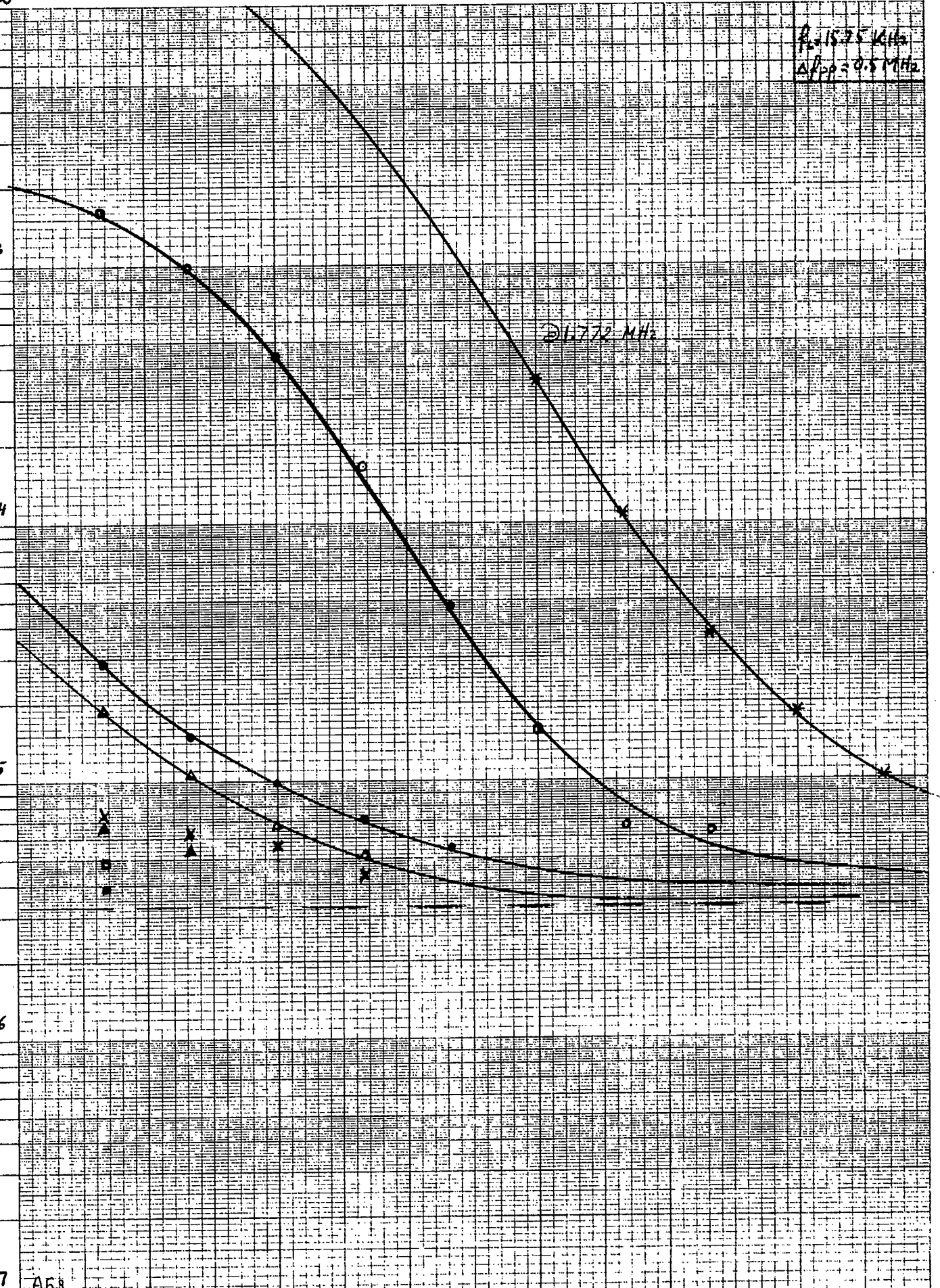
-3

0

3

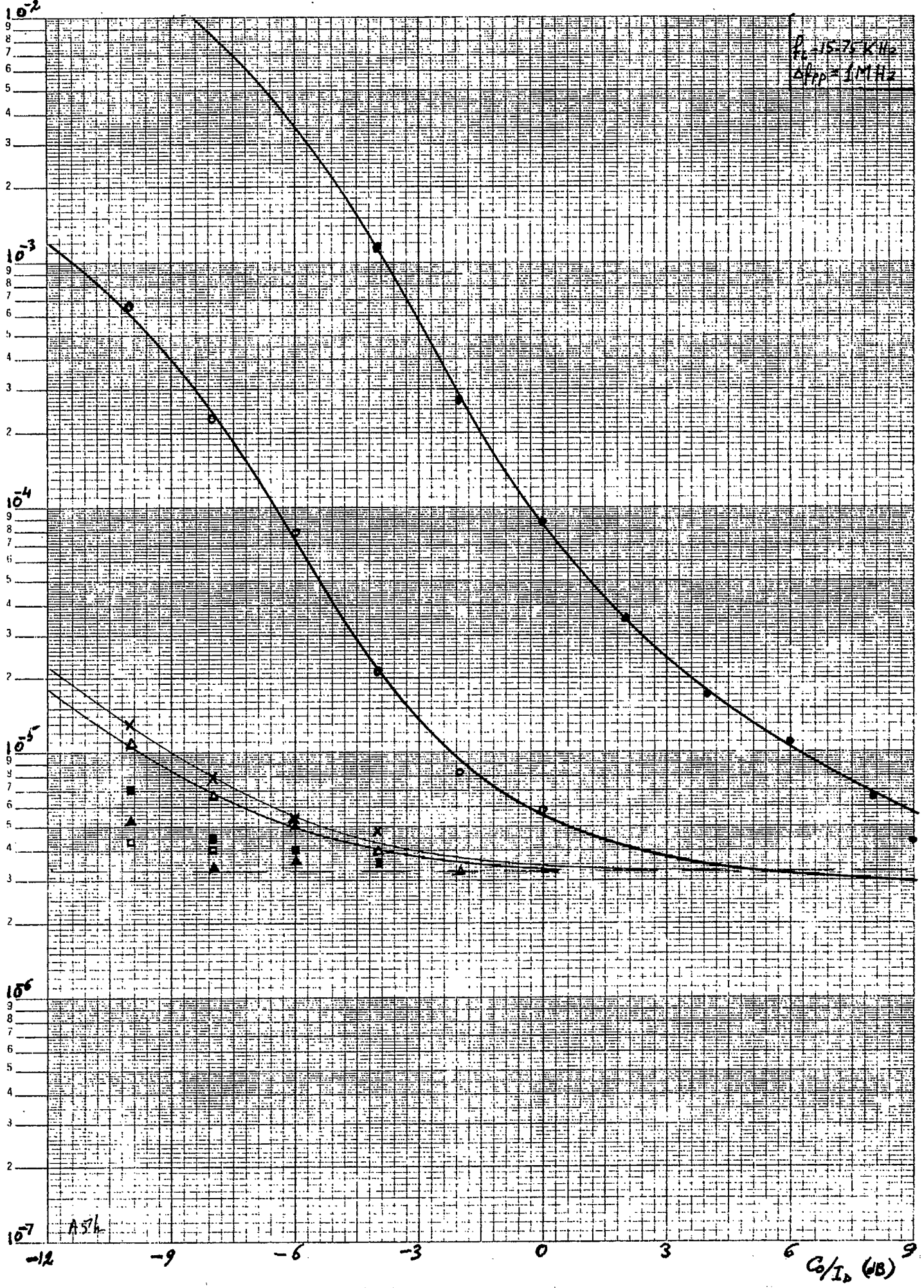
6

$C_0/I_D \text{ (dB)}^9$



BER

$P_{avg} = 15.75 \text{ KHz}$
 $\Delta f_{pp} = 1 \text{ MHz}$



SEMICONDUCTOR INSTRUMENTS
MADE IN CANADA
GRAPHIC INSTRUMENTS
MADE IN CANADA
SERIAL ARITHMETIC CYCLES IN
SPECIFY TRACING OR DRAWING PAPER

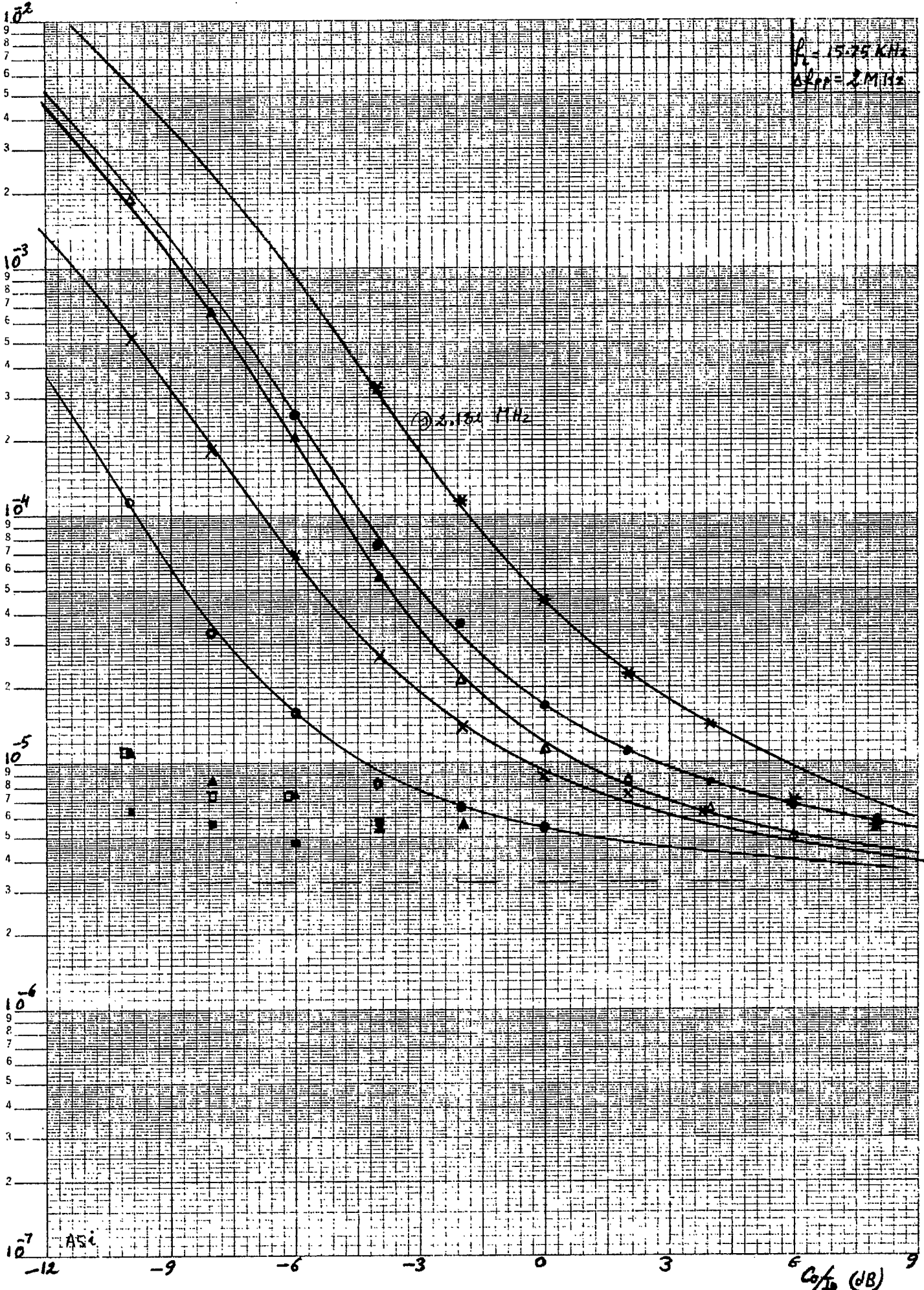
A5h

BER

$f_c = 15.75 \text{ MHz}$
 $\Delta f_{pp} = 2 \text{ MHz}$

② 2.182 MHz

ASi



SEI... R... C... 10... IN... SPECIFY TRACING OR DRAWING PAPER

JAD... MADE IN CANADA

A6

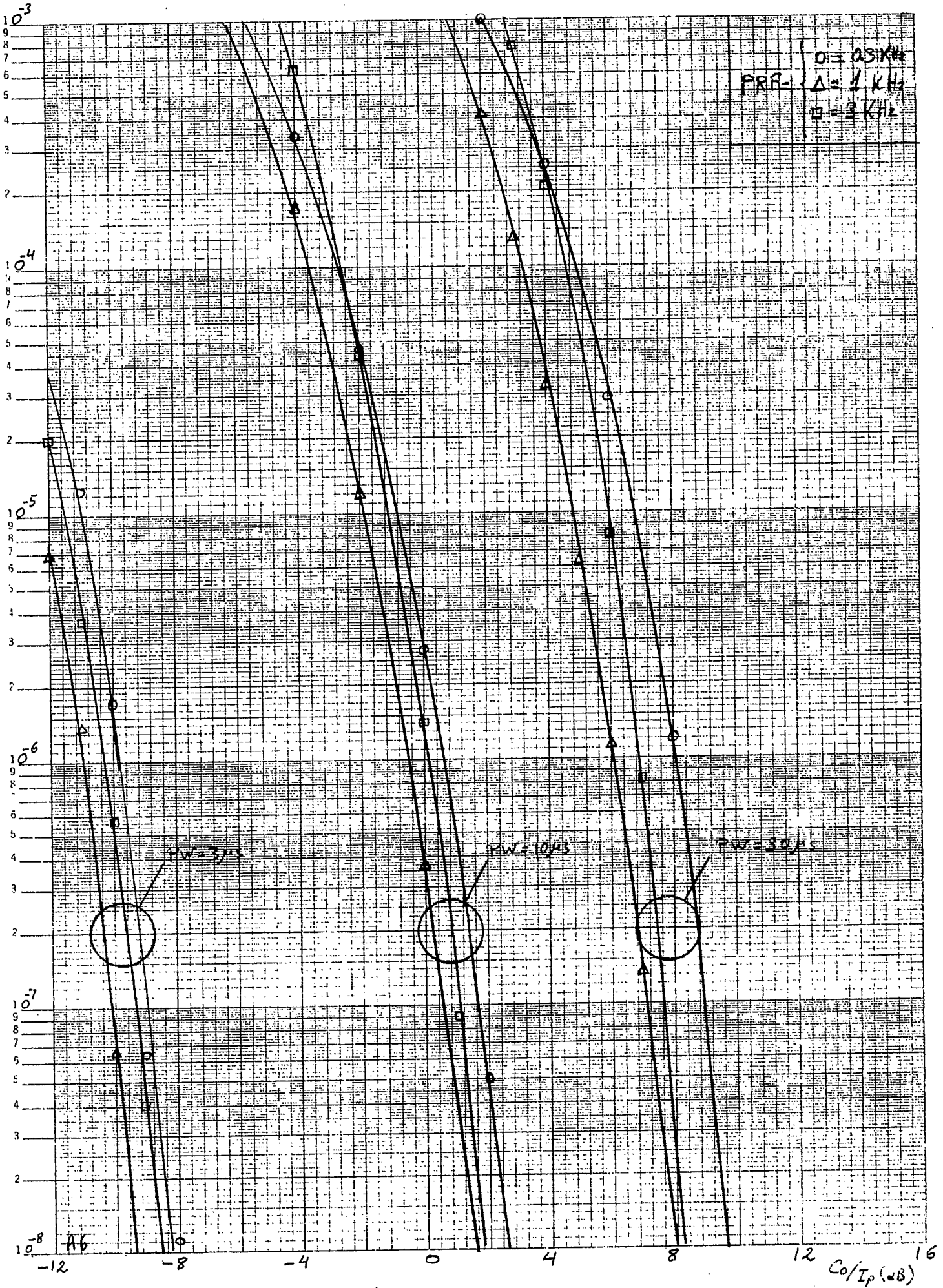
SYMBOL	PRF (kHz)
o	0.3
Δ	1.0
□	3.0

Radax-Like Interfering signal

$C_0/N \rightarrow \infty$

$f_c = 10 \text{ KHz}$

BER



AD
JTR
MADE IN
IAP
10
IN
CY
SER
RIT
SPECIFY TRACING OF DRAWING PAPER

AB

C_o/I_p (dB)

A7

SYMBOL	PRF (kHz)
○	0.3
△	1.0
□	3.0

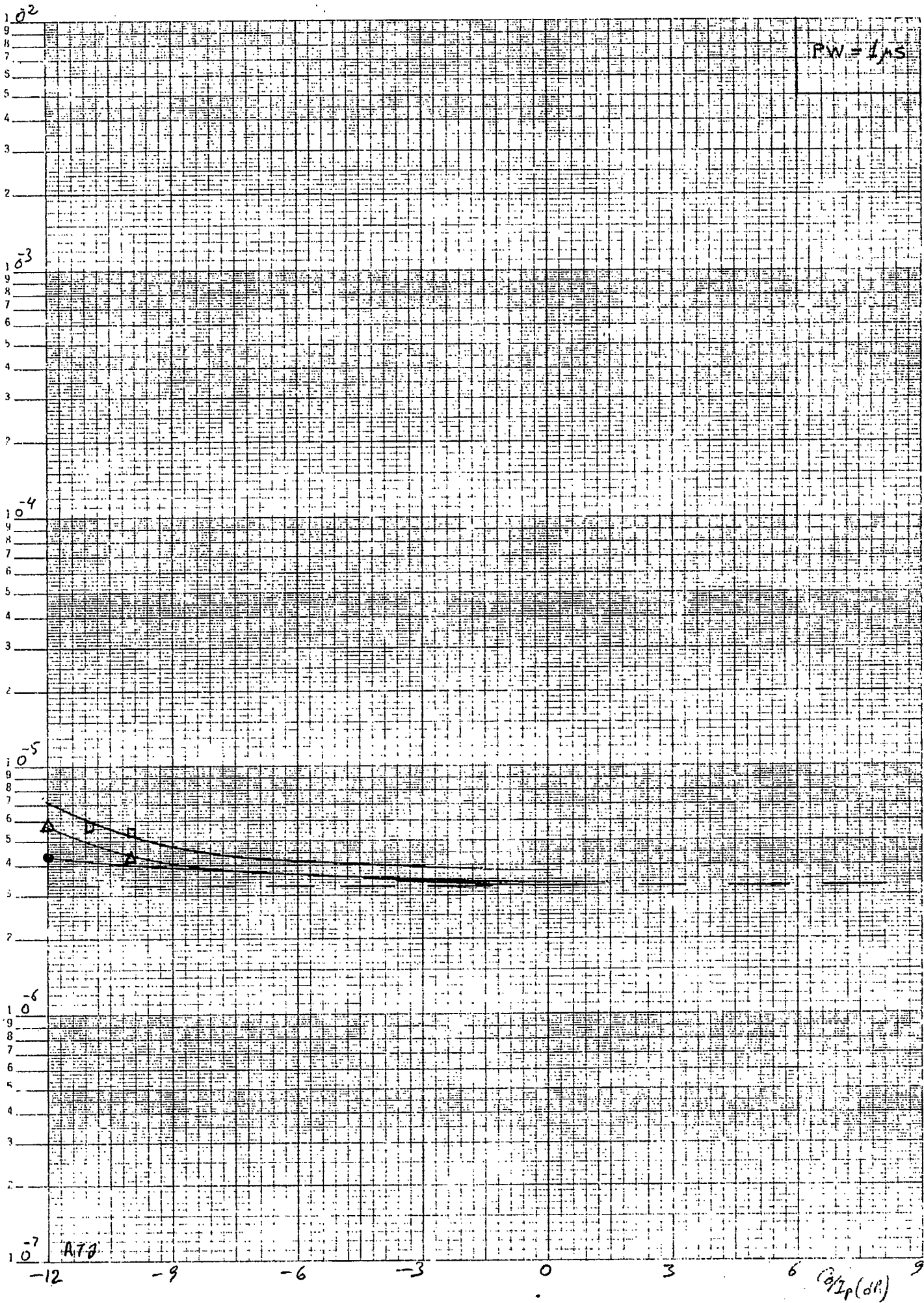
Radior-Like Interfering signal

$$\underline{C_0/N = 15 \text{ dB}}$$

$$\underline{f_c = 10 \text{ KHz}}$$

BER

PW = 1 μs



MADE IN U.S.A.
JNT
3RA
15
X
5
SPECIFY TRACING OR DRAWING PAPER

A70

$(d/L)P$ (dB)

BER

PW-315

10^{-2}

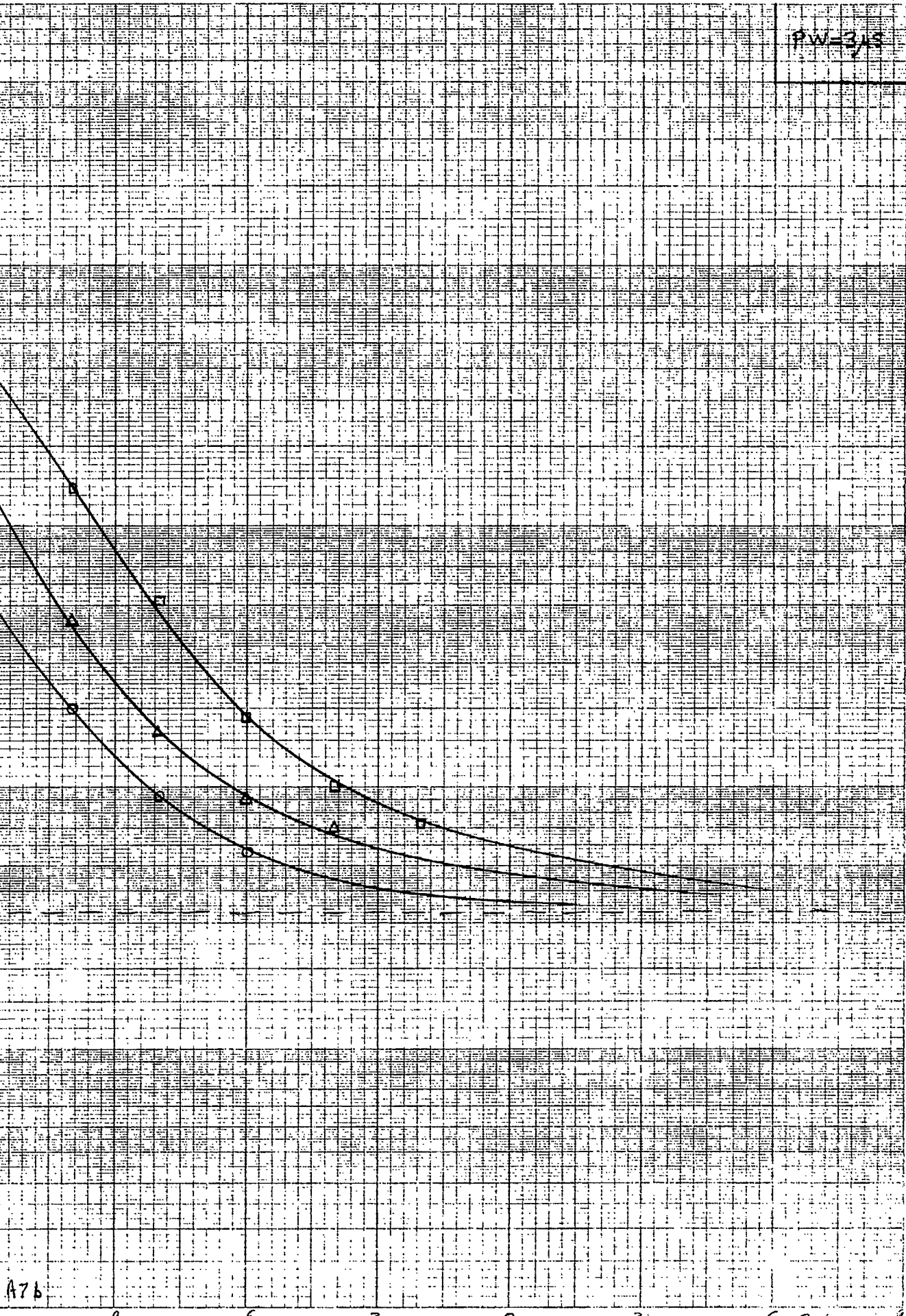
10^{-3}

10^{-4}

10^{-5}

10^{-6}

10^{-7}

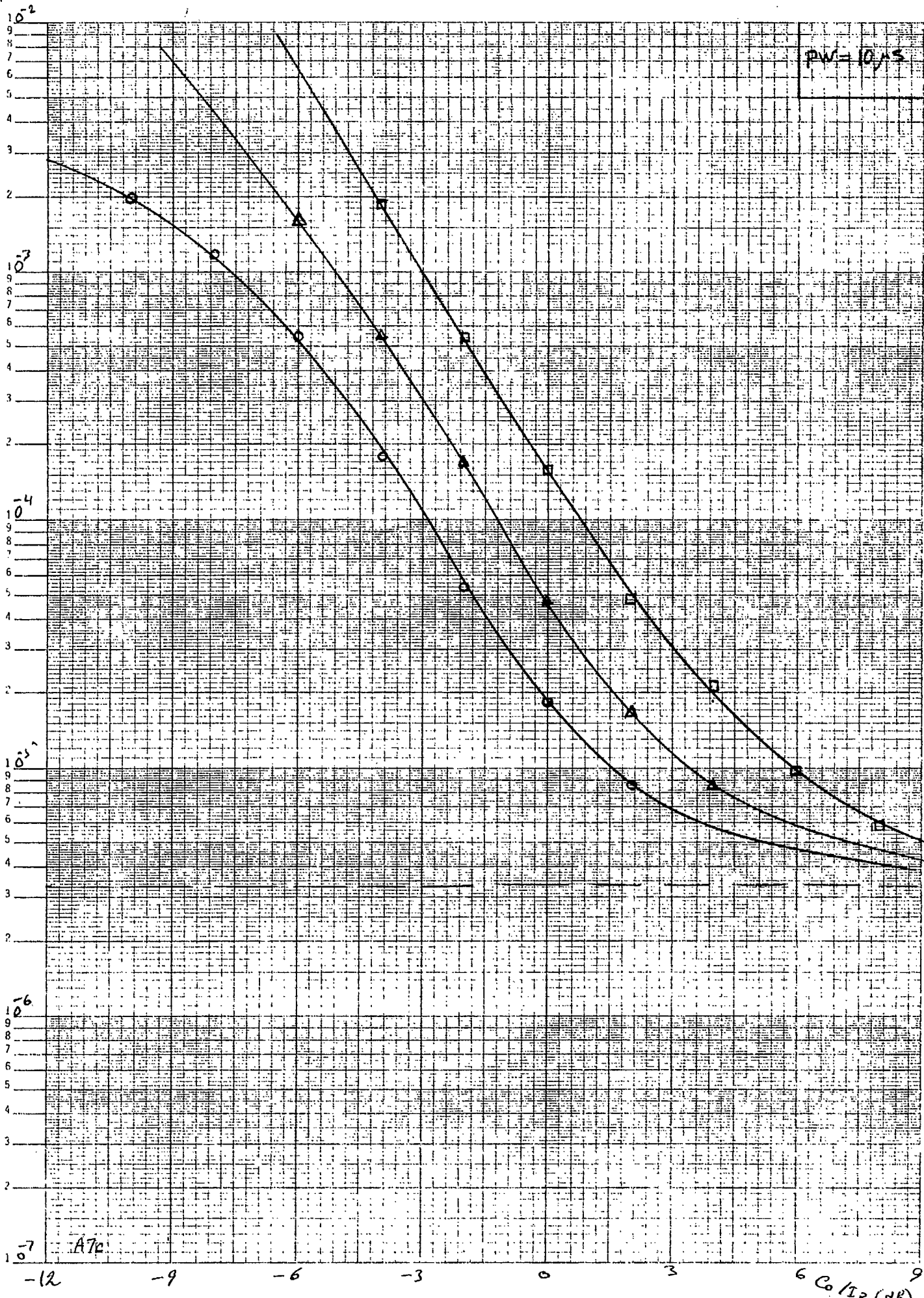


A7b

6 C/Ip (dB)

BER

PW = 10/15

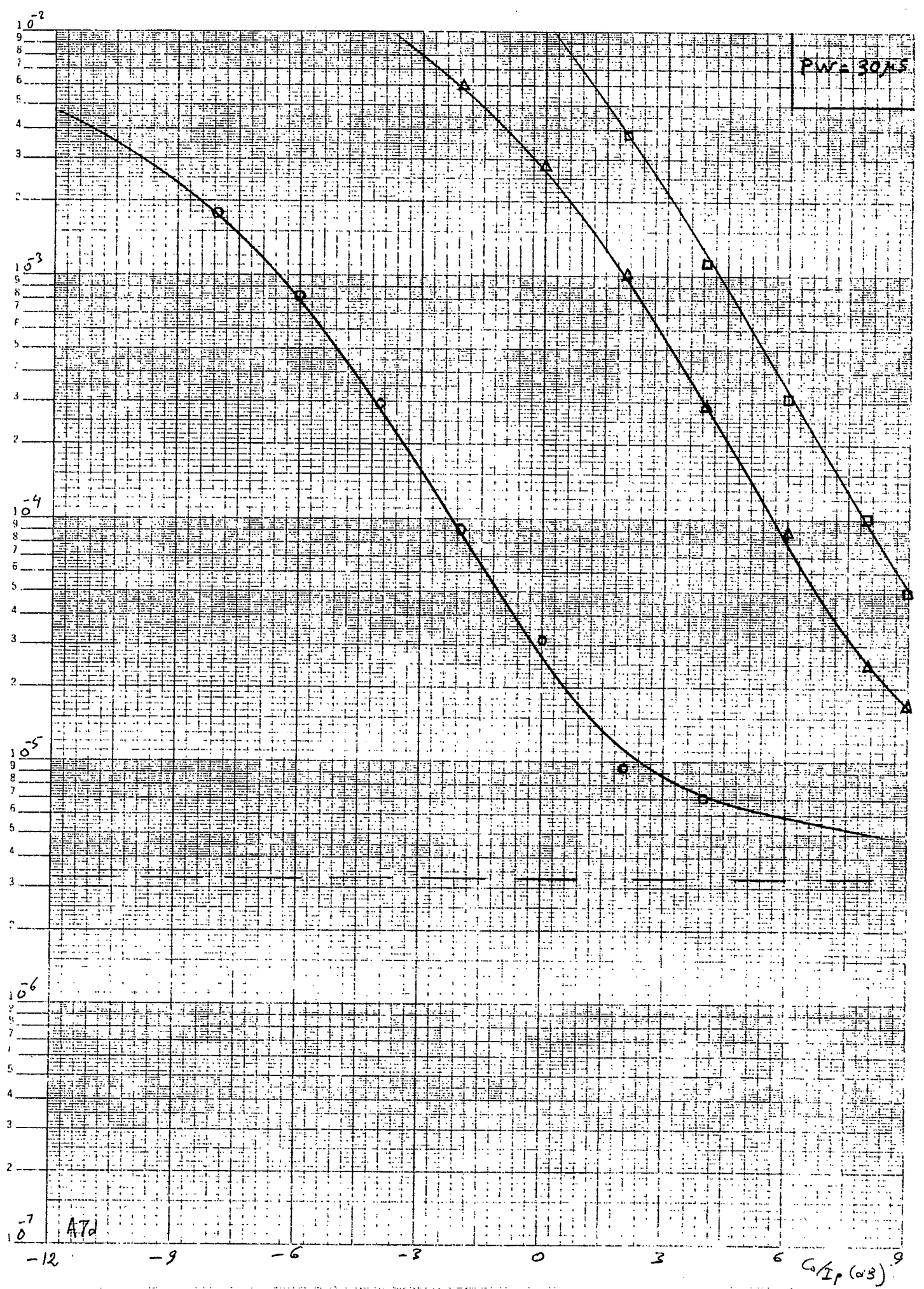


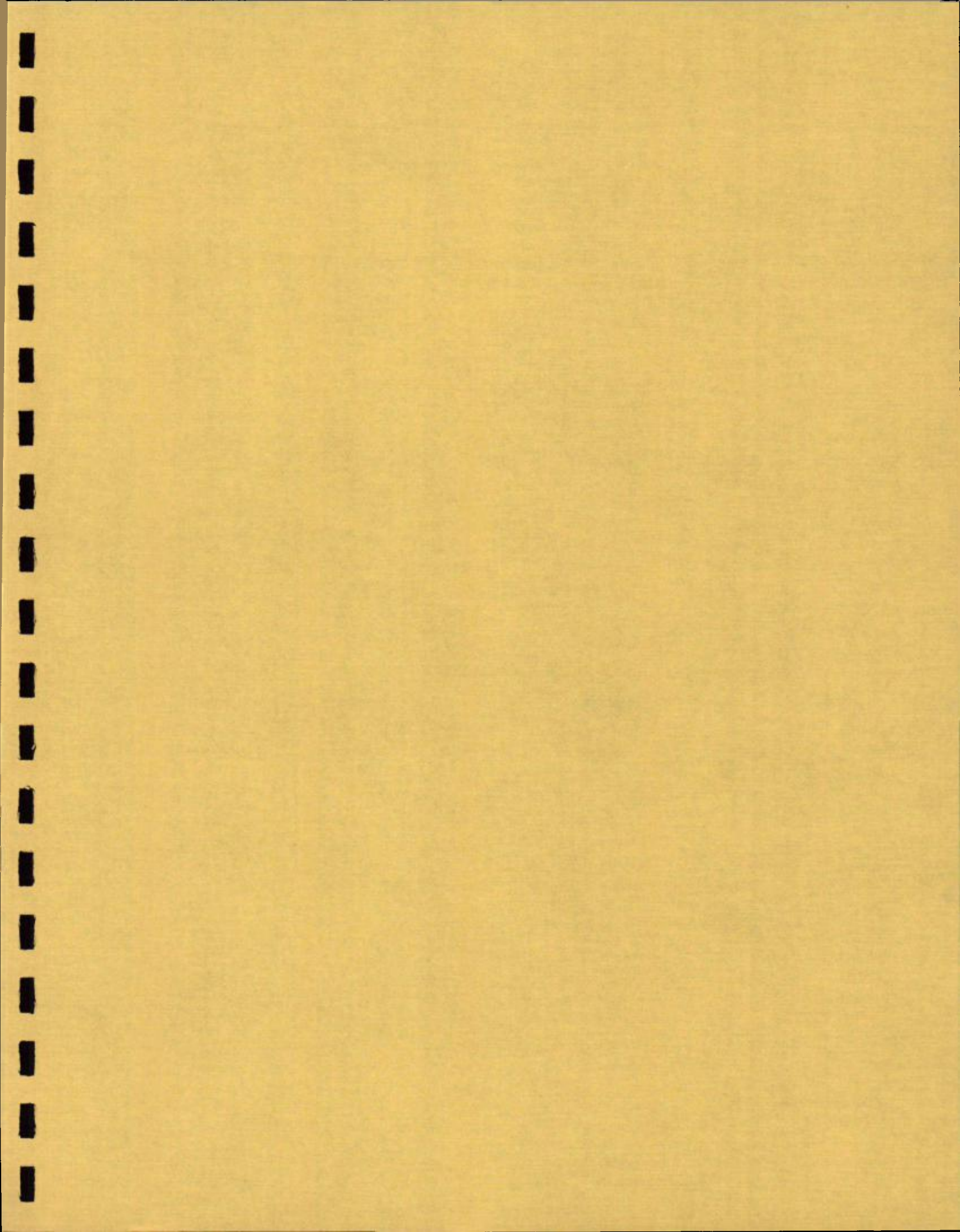
A7c

SEMI-LOGARITHMIC GRAPHING PAPER
TRADE OR SERVICE MARK
MADE IN U.S.A.

BER

PWR = 30/15



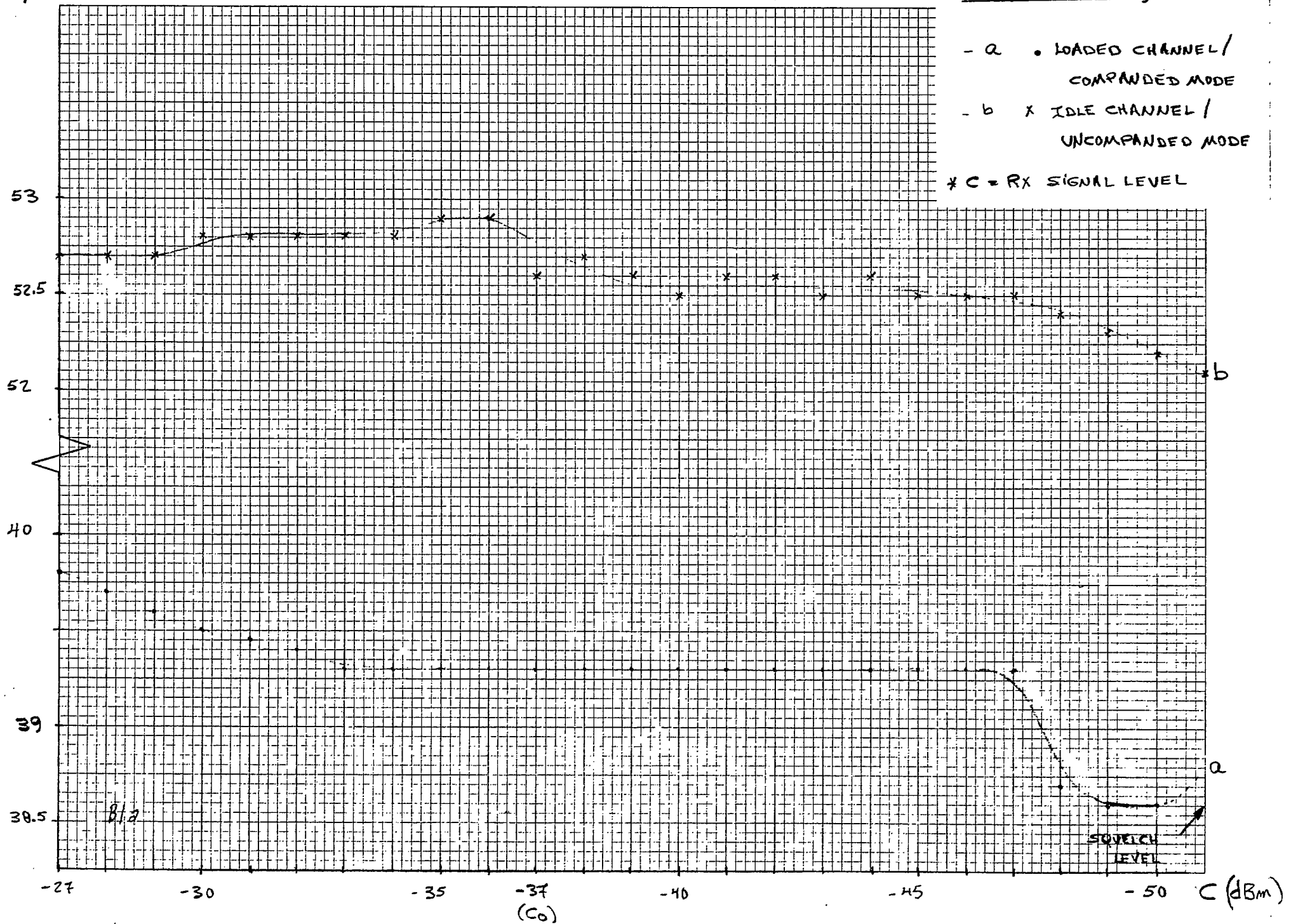


B1

S/N (dB)

S/N (C-WEIGHTED) vs C*

- a • LOADED CHANNEL / COMPANDED MODE
- b x IDLE CHANNEL / UNCOMPANDED MODE
- * C = RX SIGNAL LEVEL



S/N (dB) = 88(dB) - Threshold (dBm)

49

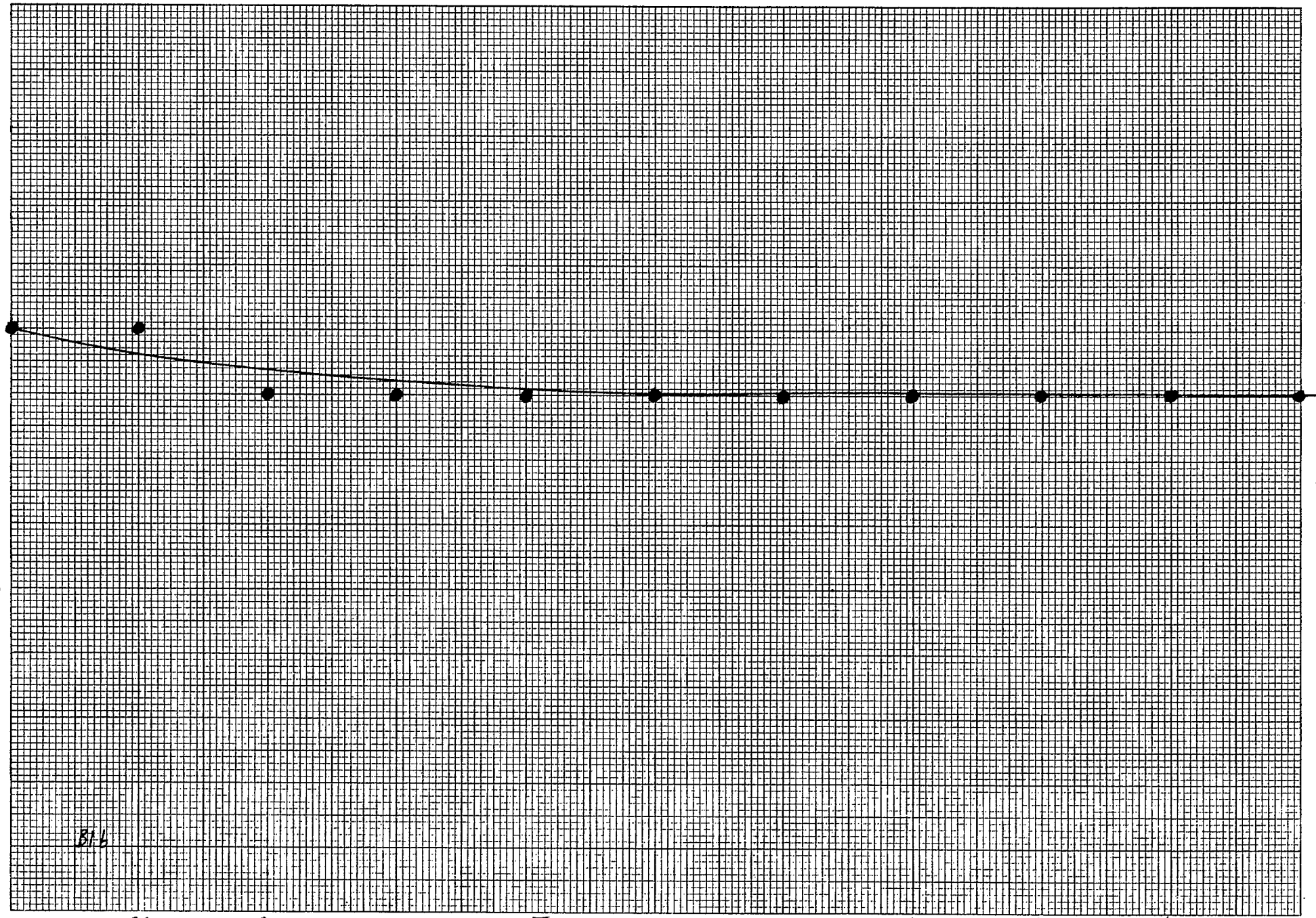
45

40

35

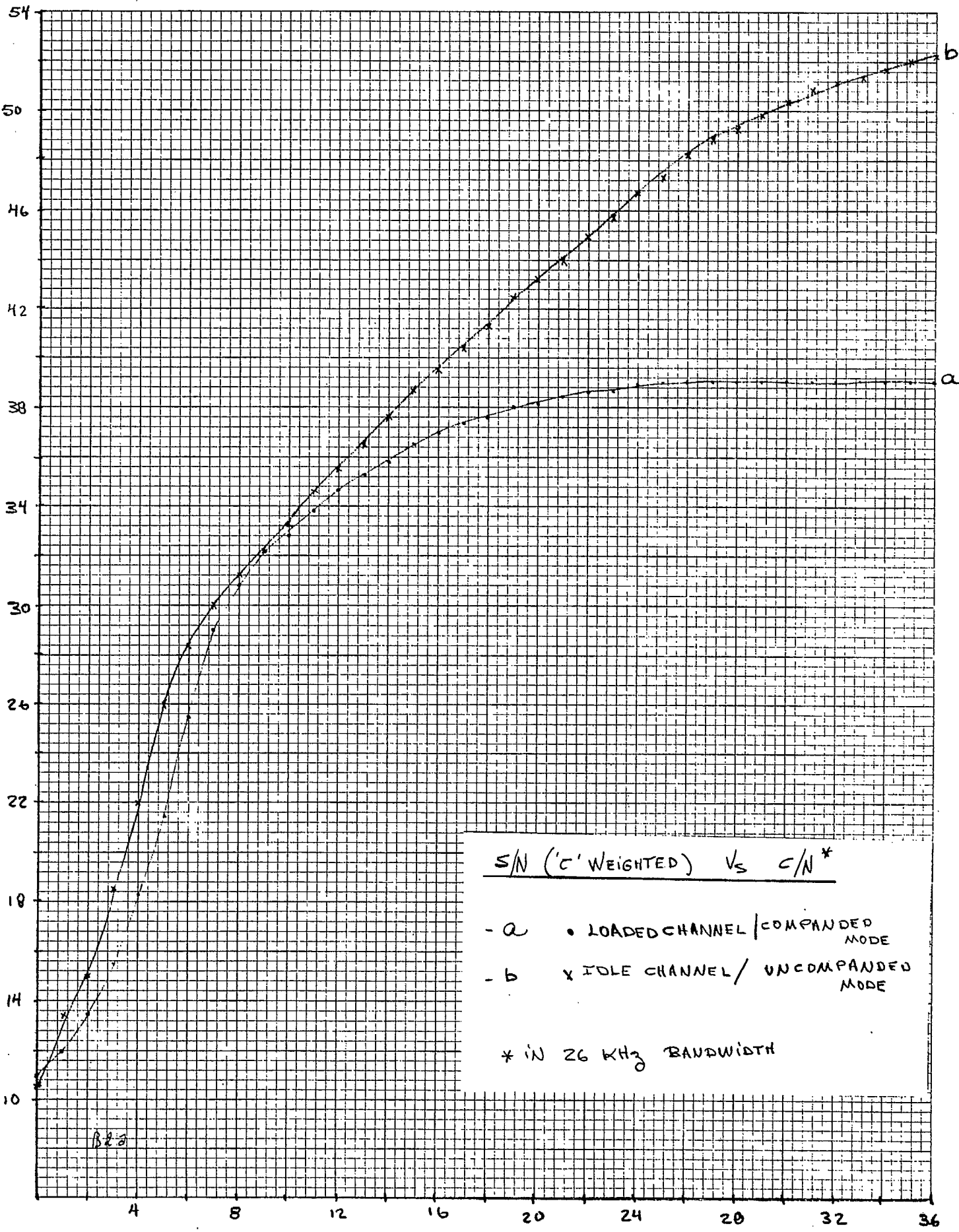
-27 -29 -31 -33 -35 -37 -39 -41 -43 -45 -47
C₀ C (dBm)

81.6



B2

SN (dB)



SN ('C' WEIGHTED) vs C/N*

- a • LOADED CHANNEL / COMPANDED MODE
- b x IDLE CHANNEL / UNCOMPANDED MODE

* IN 26 KHz BANDWIDTH

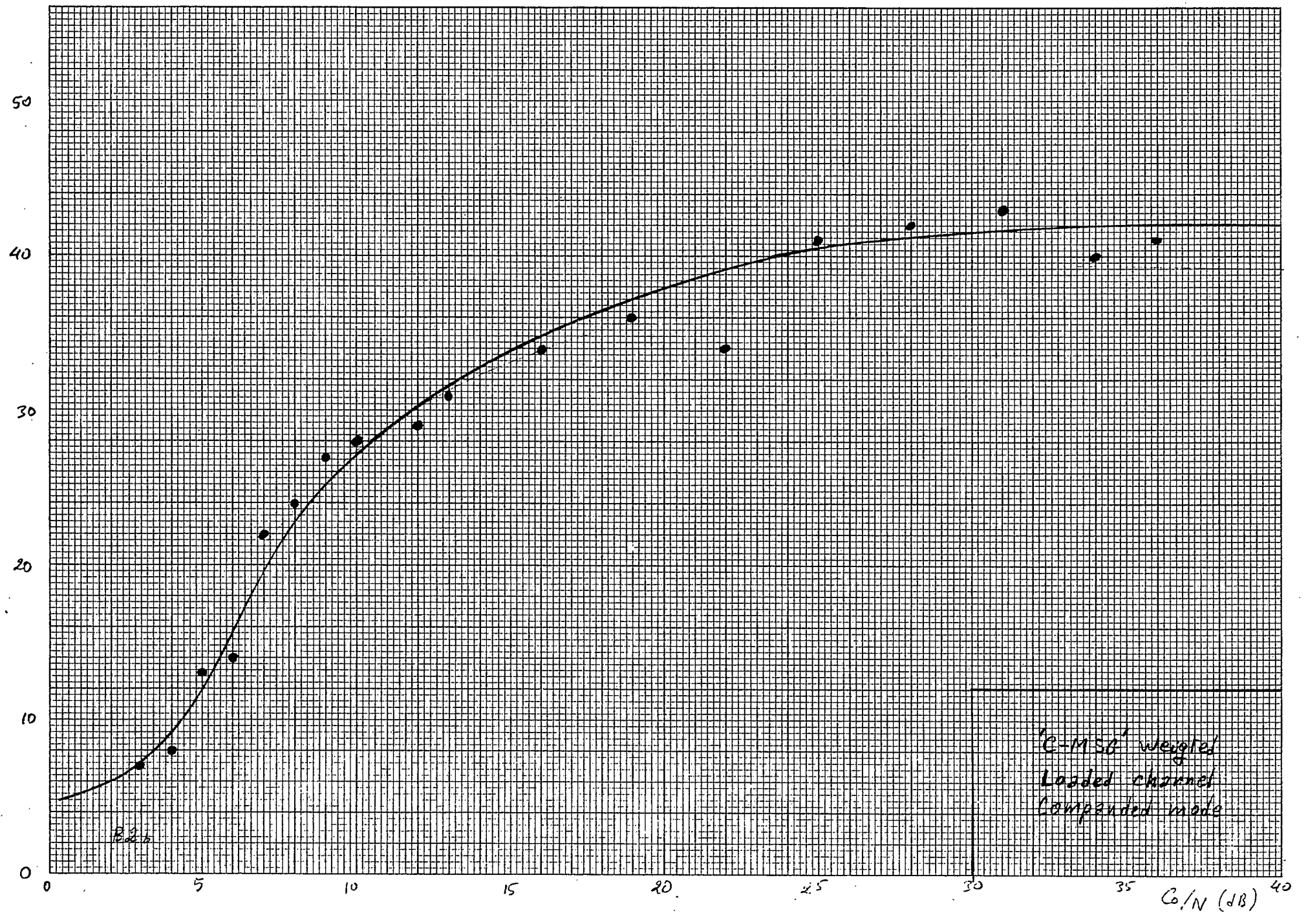
B2a

C/N

46 1610

5 X 5 TO THE CENTIMETER 18 X 24 CM.
KEUFFEL & ESSER CO. MADE IN U.S.A.

S/N (dB) = 88(dB) - Threshold (dB RN)

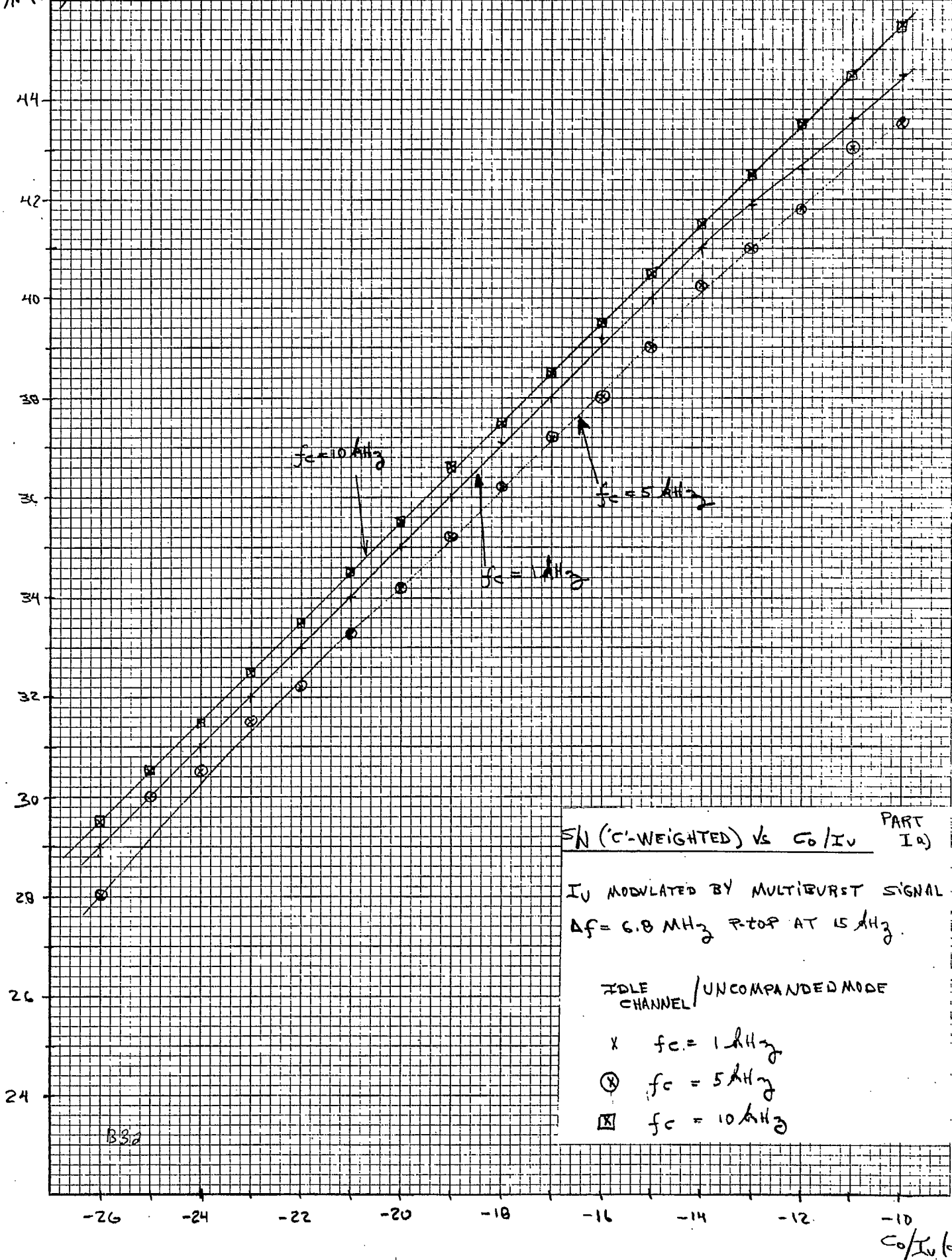


B3

S/N (dB)

46 1610

K&E
5 X 5 TO THE CENTIMETER 18 X 24 CM.
KEUFFEL & ESSER CO. MADE IN U.S.A.



S/N (C-WEIGHTED) vs C_0/I_v PART I(a)

I_v MODULATED BY MULTIBURST SIGNAL
 $\Delta f = 6.8$ MHz P-TOP AT 15 kHz

IDLE CHANNEL / UNCOMPANDED MODE

- x $f_c = 1$ kHz
- ⊗ $f_c = 5$ kHz
- ⊠ $f_c = 10$ kHz

B32

C_0/I_v (dB)

S/N (dB)

44

42

40

38

36

34

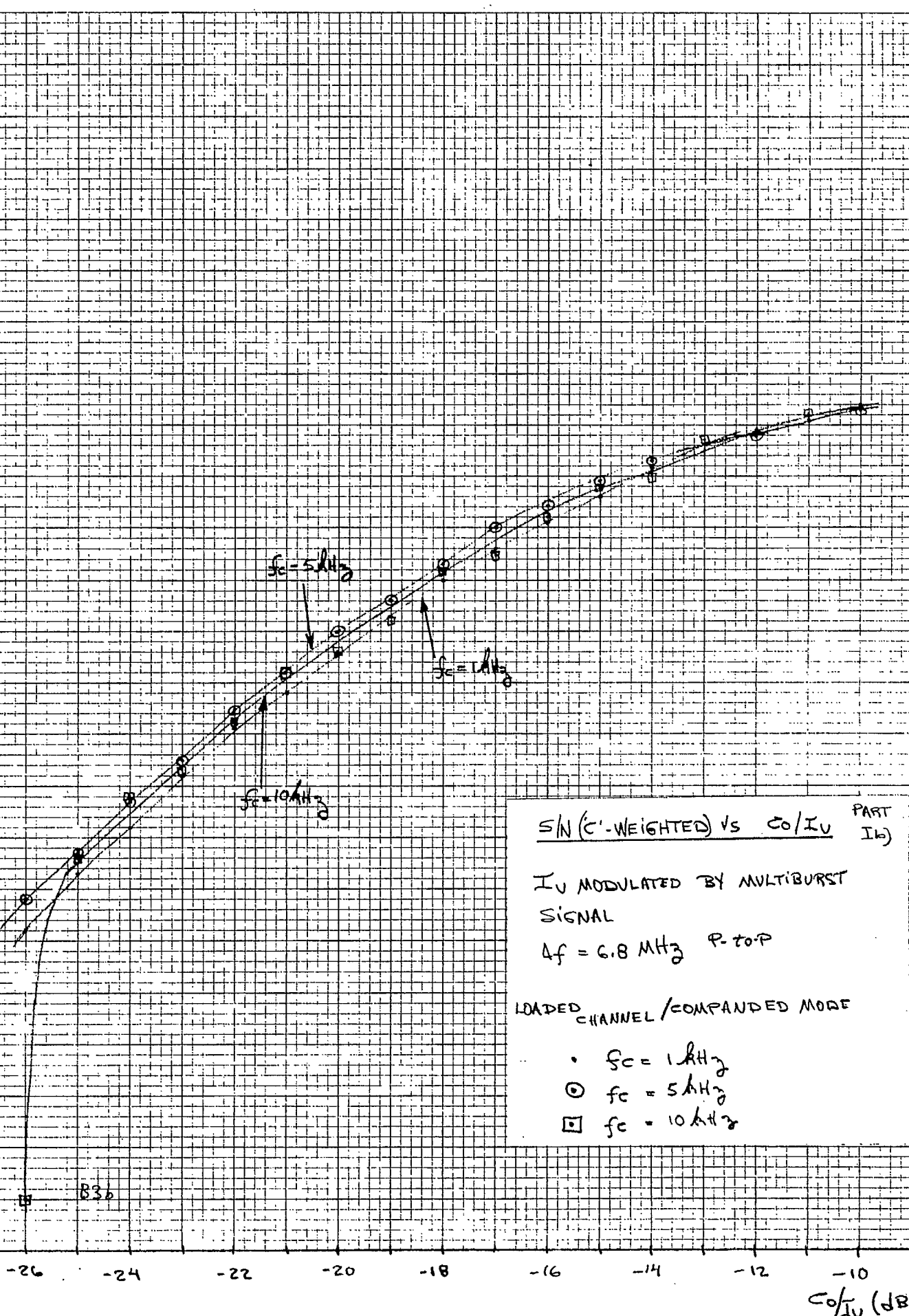
32

30

28

26

24



S/N (C'-WEIGHTED) vs C_0/I_U PART I(b)

I_U MODULATED BY MULTIBURST SIGNAL

$\Delta f = 6.8 \text{ MHz}$ P-to-P

LOADED CHANNEL / COMPANDED MORE

- $f_c = 1 \text{ kHz}$
- $f_c = 5 \text{ kHz}$
- $f_c = 10 \text{ kHz}$

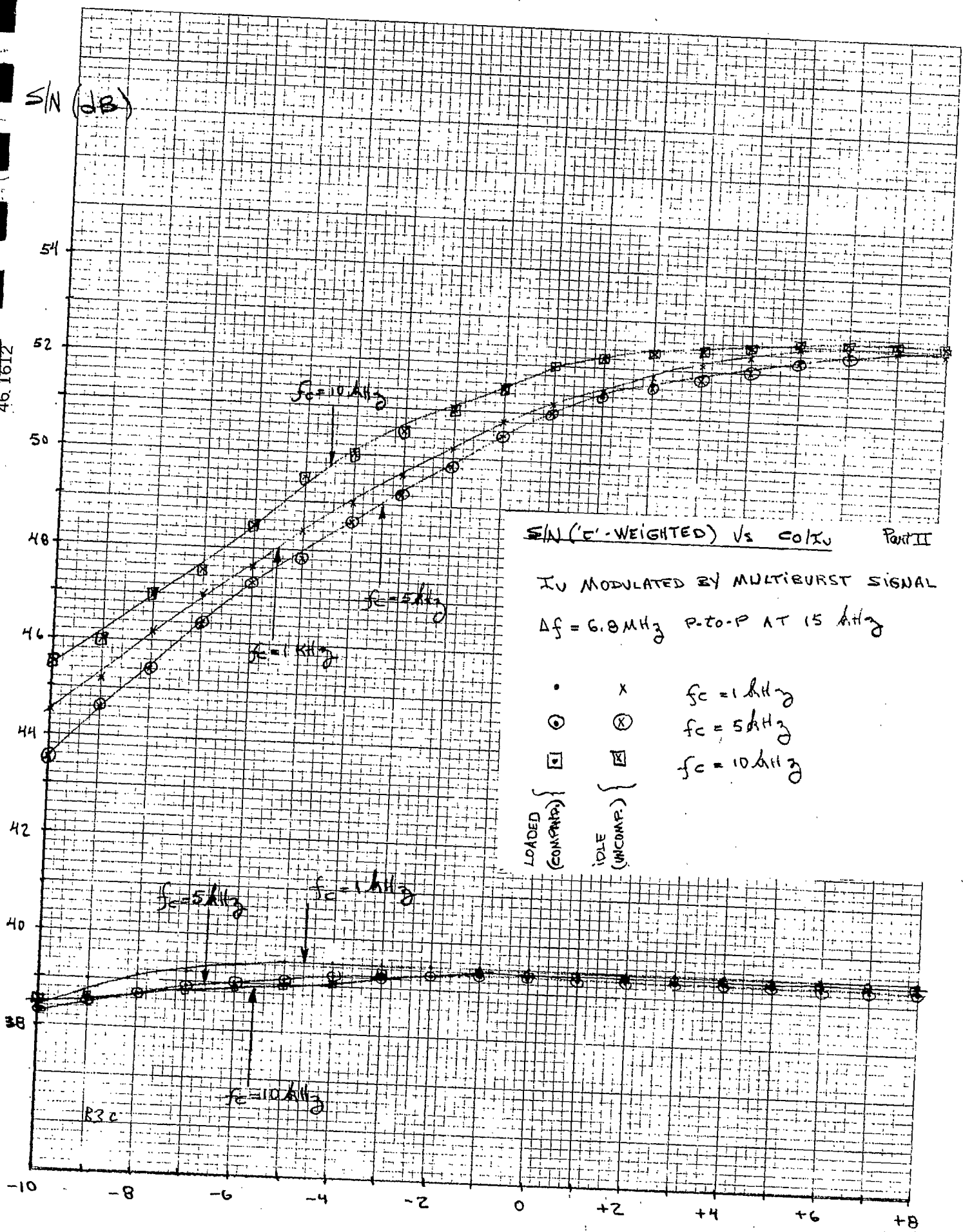
-26 -24 -22 -20 -18 -16 -14 -12 -10
 C_0/I_U (dB)

K 2 X 5 KEUFFEL & ESSER CO. MADE IN U.S.A. 46 1612

S/N (dB)

46.1612

J.M. L.R. KEUFFEL & ESSER CO. MADE IN U.S.A.



S/N ('C'-WEIGHTED) vs C_0/I_0 Part II

I_0 MODULATED BY MULTIBURST SIGNAL
 $\Delta f = 6.8 \text{ MHz}$ P-to-P AT 15 kHz

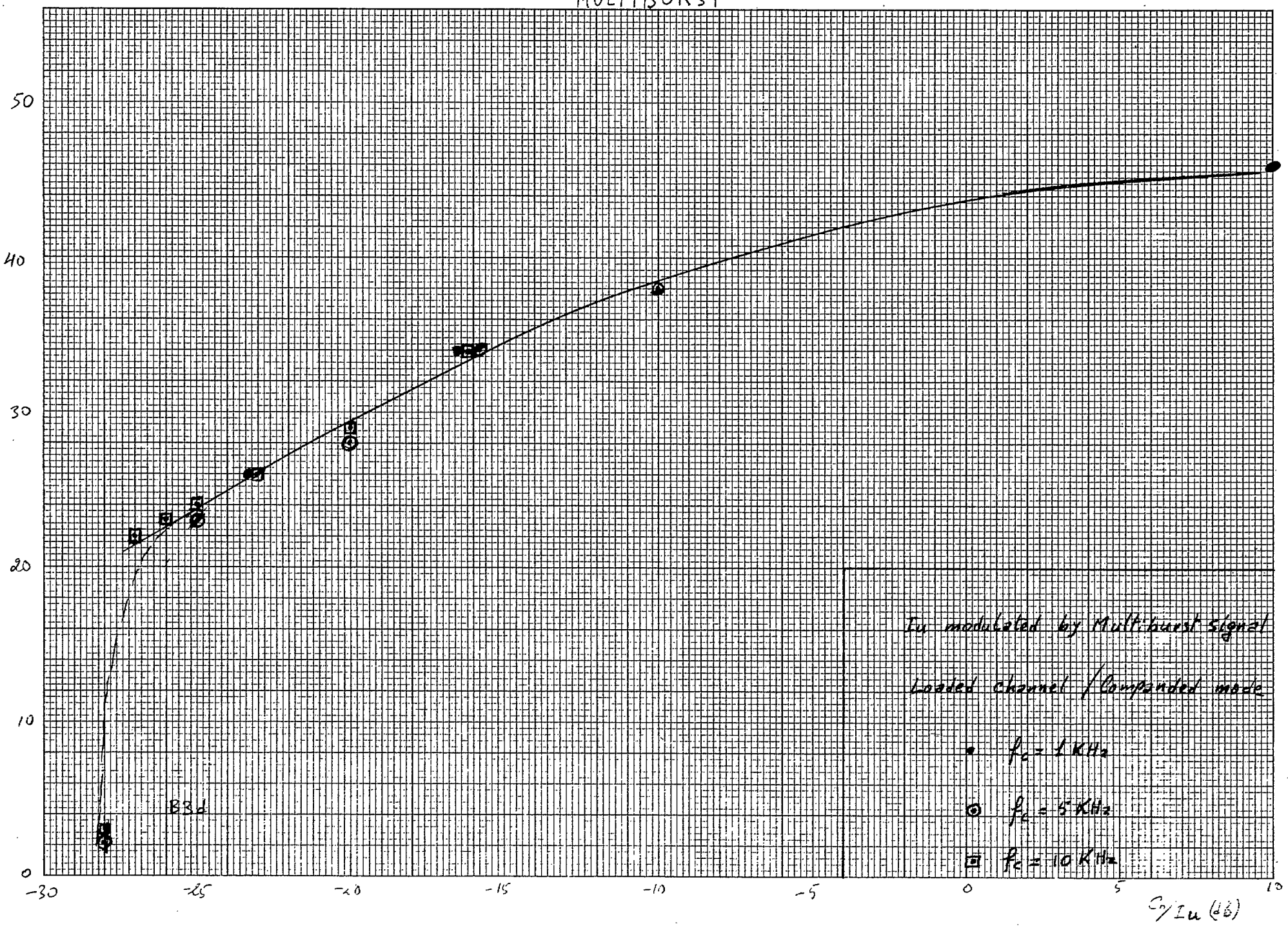
- $f_c = 1 \text{ kHz}$
- ⊙ $f_c = 5 \text{ kHz}$
- ⊠ $f_c = 10 \text{ kHz}$
- x (LOADED (COMP.)) $f_c = 1 \text{ kHz}$
- ⊗ (IDLE (UNCOMP.)) $f_c = 5 \text{ kHz}$
- ⊠ (IDLE (UNCOMP.)) $f_c = 10 \text{ kHz}$

B3c

C_0/I_0 (dB)

$S/N^{(dB)} = 88(dB) - \text{Threshold (dB)}_{RN}$

"MULTIBURST"

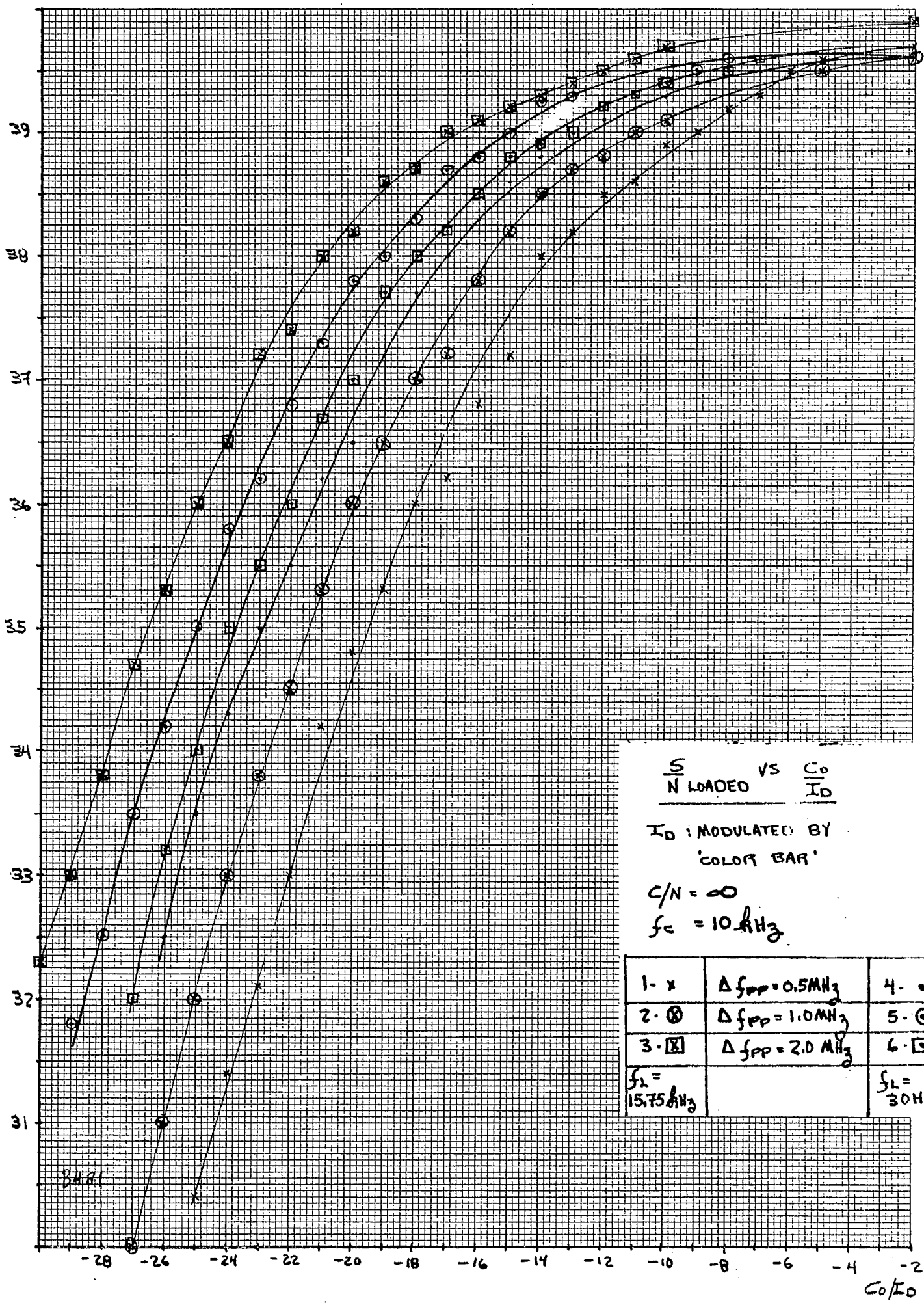


B4

S/N LOADED

40 1331

KEUFFEL & ESSER CO. MADE IN U.S.A.



$\frac{S}{N \text{ LOADED}} \text{ VS } \frac{C_0}{I_D}$

I_D : MODULATED BY 'COLOR BAR'

$C/N = \infty$
 $f_c = 10 \text{ MHz}$

1 - x	$\Delta f_{pp} = 0.5 \text{ MHz}$	4 - o
2 - ⊗	$\Delta f_{pp} = 1.0 \text{ MHz}$	5 - ⊙
3 - ⊠	$\Delta f_{pp} = 2.0 \text{ MHz}$	6 - ⊡
$f_L = 15.75 \text{ MHz}$		$f_L = 30 \text{ MHz}$

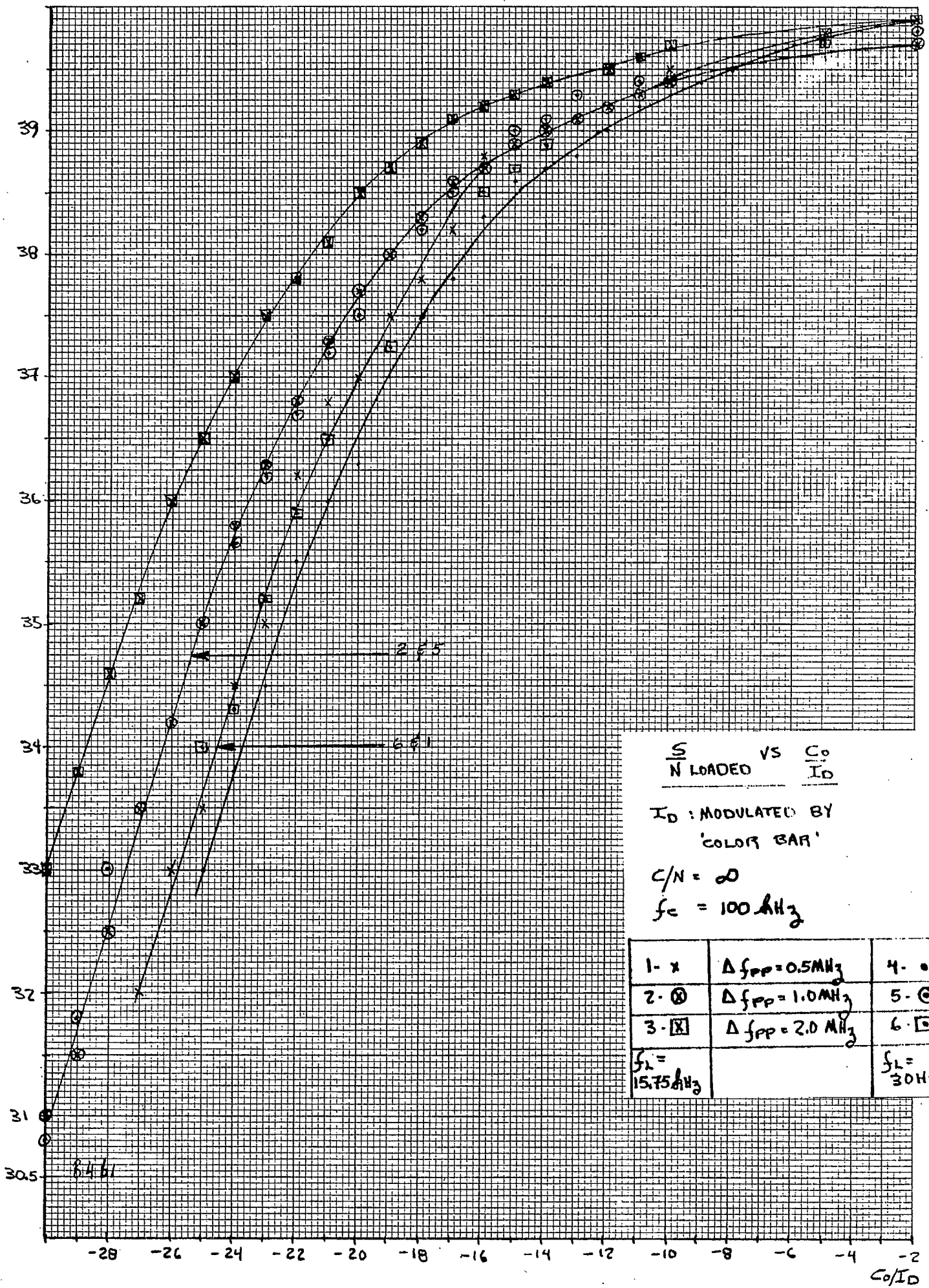
3471

C_0/I_D

S/N LOADED

46 1331

10 X 10 1/2 INCH 7 X 10 INCHES
K & Z KEUFFEL & ESSER CO. MADE IN U.S.A.



$\frac{S}{N \text{ LOADED}}$ VS $\frac{C_0}{I_D}$

I_D : MODULATED BY 'COLOR BAR'

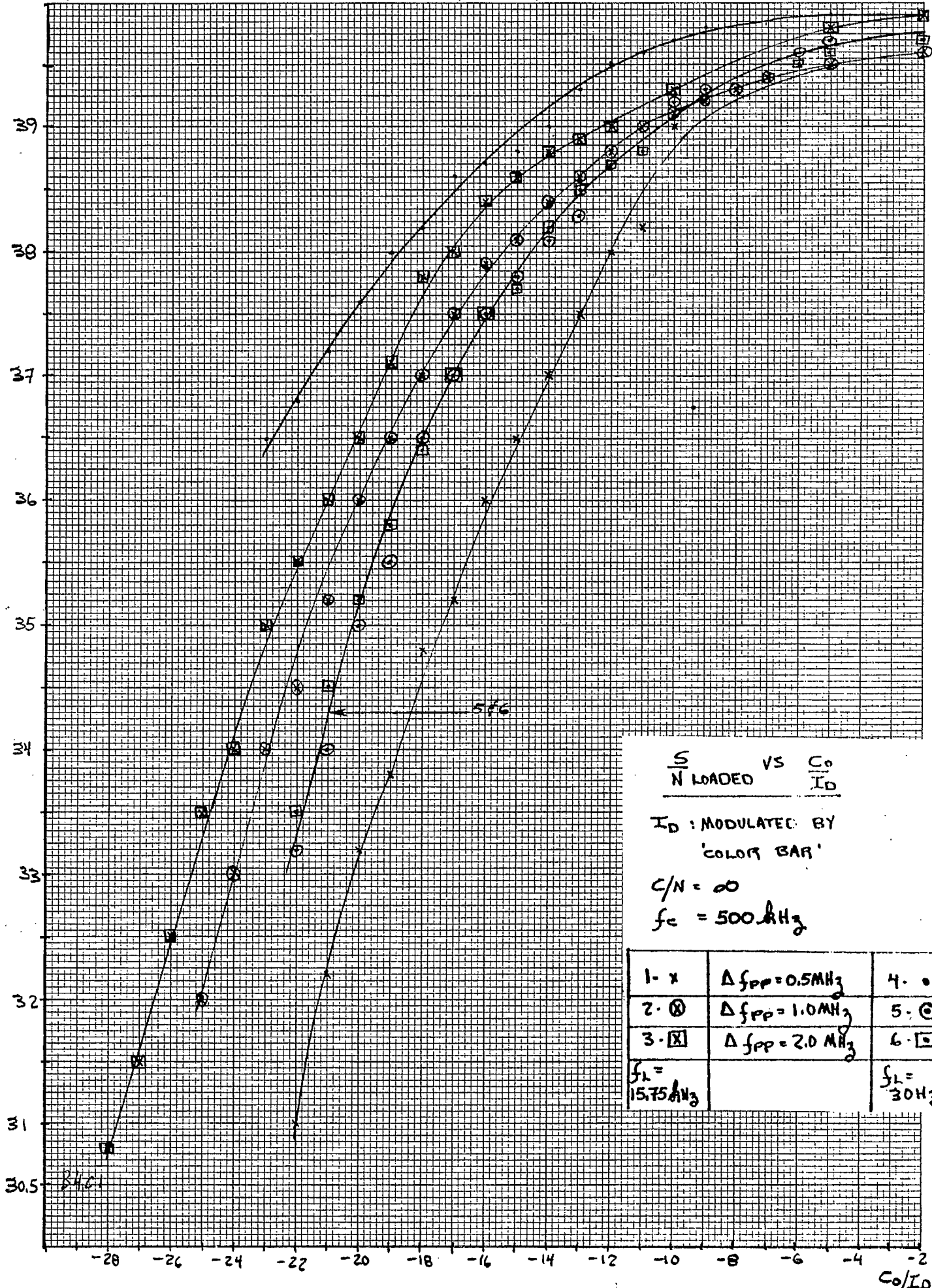
$C/N = \infty$

$f_c = 100 \text{ MHz}$

1-x	$\Delta f_{PP} = 0.5 \text{ MHz}$	4-•
2-⊙	$\Delta f_{PP} = 1.0 \text{ MHz}$	5-⊙
3-⊠	$\Delta f_{PP} = 2.0 \text{ MHz}$	6-⊠
$f_L = 15.75 \text{ MHz}$		$f_L = 30 \text{ MHz}$

C_0/I_D

S/N LOADED



$\frac{S}{N}$ LOADED VS $\frac{C_0}{I_D}$

I_D : MODULATED BY 'COLOR BAR'

$C/N = \infty$

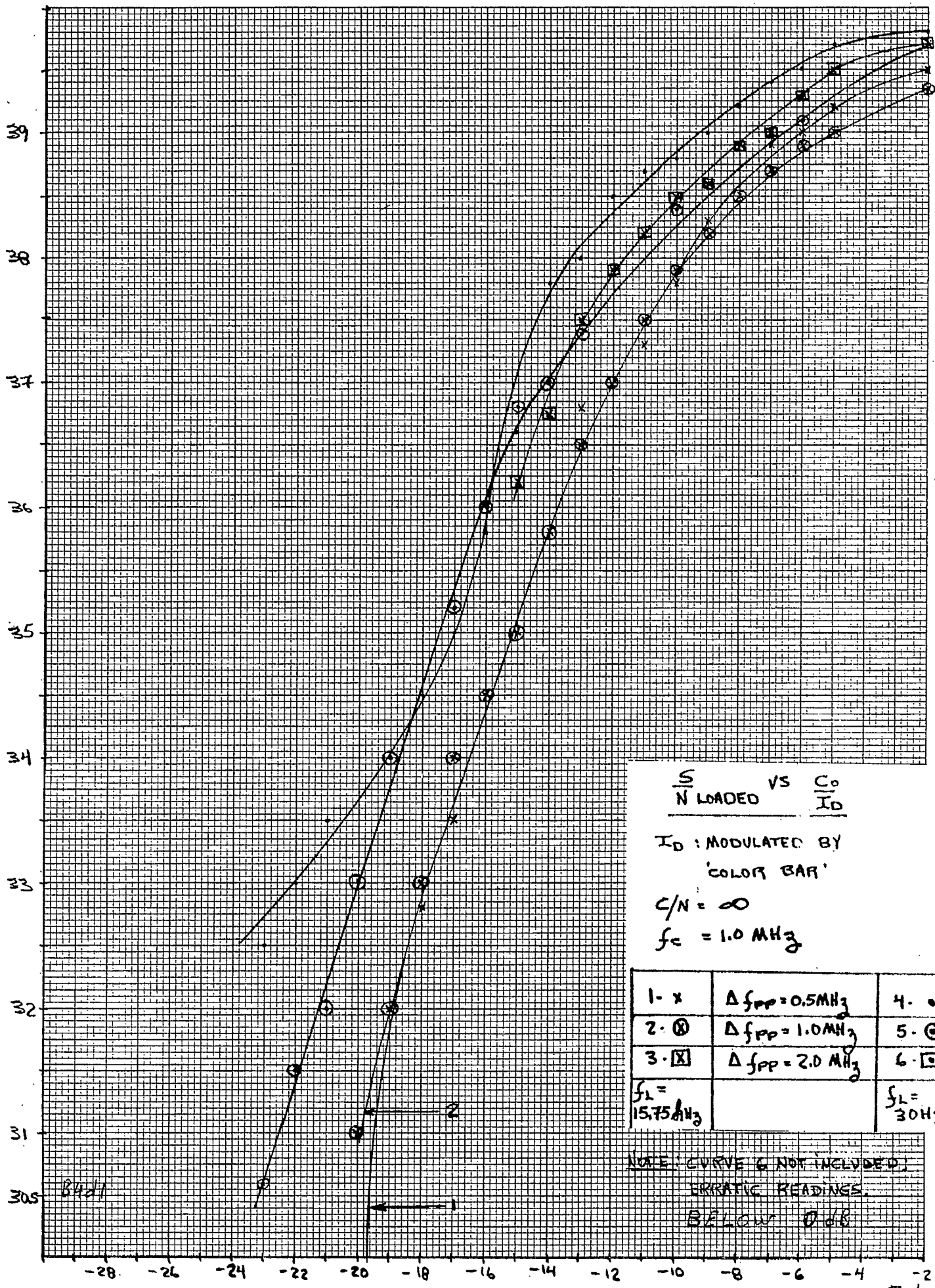
$f_c = 500 \text{ kHz}$

1. x	$\Delta f_{pp} = 0.5 \text{ MHz}$	4. o
2. o	$\Delta f_{pp} = 1.0 \text{ MHz}$	5. o
3. x	$\Delta f_{pp} = 2.0 \text{ MHz}$	6. o
$f_L = 15.75 \text{ MHz}$		$f_L = 30 \text{ MHz}$

KEUFFEL & ESSER CO. MADE IN U.S.A.

S/N LOADED

KEUFFEL & ESSER CO. MADE IN U.S.A.
 10 INCHES
 46 1331



$\frac{S}{N} \text{ LOADED VS } \frac{C_0}{I_0}$

I_0 : MODULATED BY 'COLOR BAR'

$C/N = \infty$

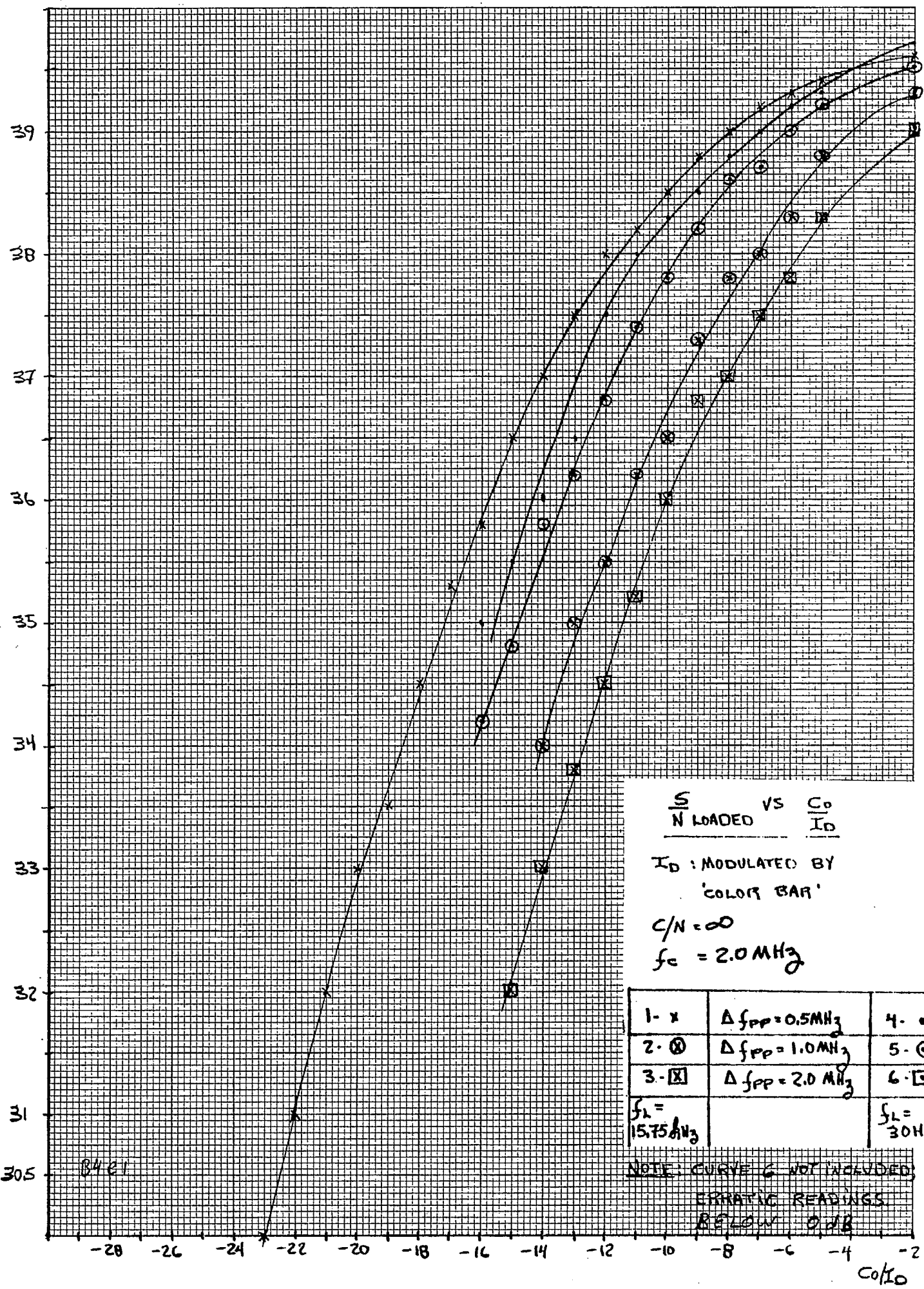
$f_c = 1.0 \text{ MHz}$

1 - x	$\Delta f_{pp} = 0.5 \text{ MHz}$	4 - o
2 - @	$\Delta f_{pp} = 1.0 \text{ MHz}$	5 - @
3 - [X]	$\Delta f_{pp} = 2.0 \text{ MHz}$	6 - []
$f_L = 15.75 \text{ kHz}$		$f_L = 30 \text{ kHz}$

NOTE: CURVE 6 NOT INCLUDED;
ERRATIC READINGS
BELOW 0.68

C_0/I_0

S/N LOADED



$\frac{S}{N \text{ LOADED}} \text{ VS } \frac{C_0}{I_D}$

I_D : MODULATED BY 'COLOR BAR'

$C/N = \infty$

$f_c = 2.0 \text{ MHz}$

1 - x	$\Delta f_{pp} = 0.5 \text{ MHz}$	4 - o
2 - ⊙	$\Delta f_{pp} = 1.0 \text{ MHz}$	5 - ⊖
3 - ⊠	$\Delta f_{pp} = 2.0 \text{ MHz}$	6 - ⊡
$f_L = 15.75 \text{ MHz}$		$f_L = 30 \text{ Hz}$

NOTE: CURVE 6 NOT INCLUDED;
ERRATIC READINGS
BELOW 0 dB

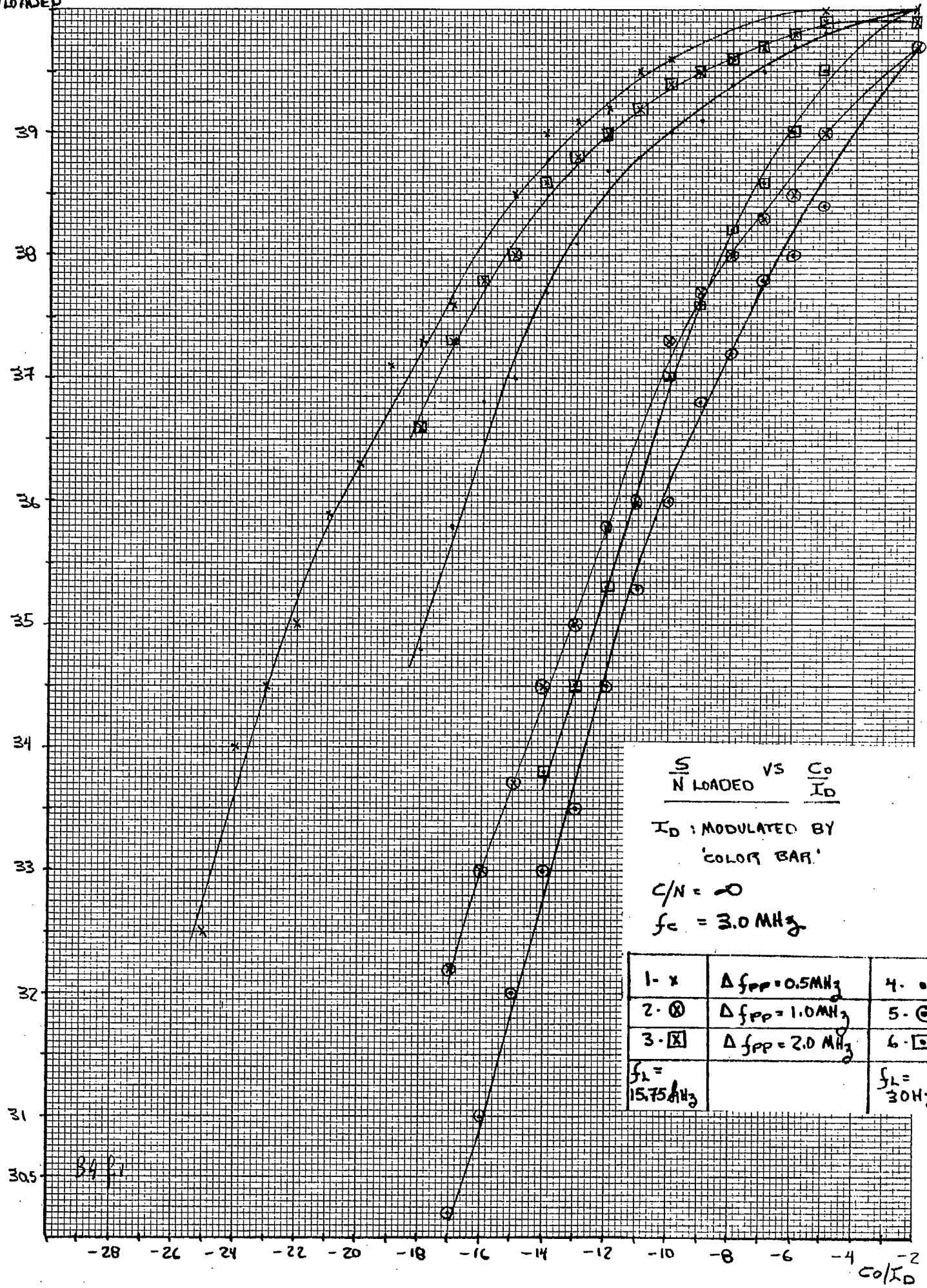
8421

C_0/I_D

UNLOADED

46 1331

KEUFFEL & ESSER CO. MADE IN U.S.A.



$\frac{S}{N \text{ LOADED}}$ VS $\frac{C_0}{I_D}$

I_D : MODULATED BY 'COLOR BAR'

$C/N = \infty$

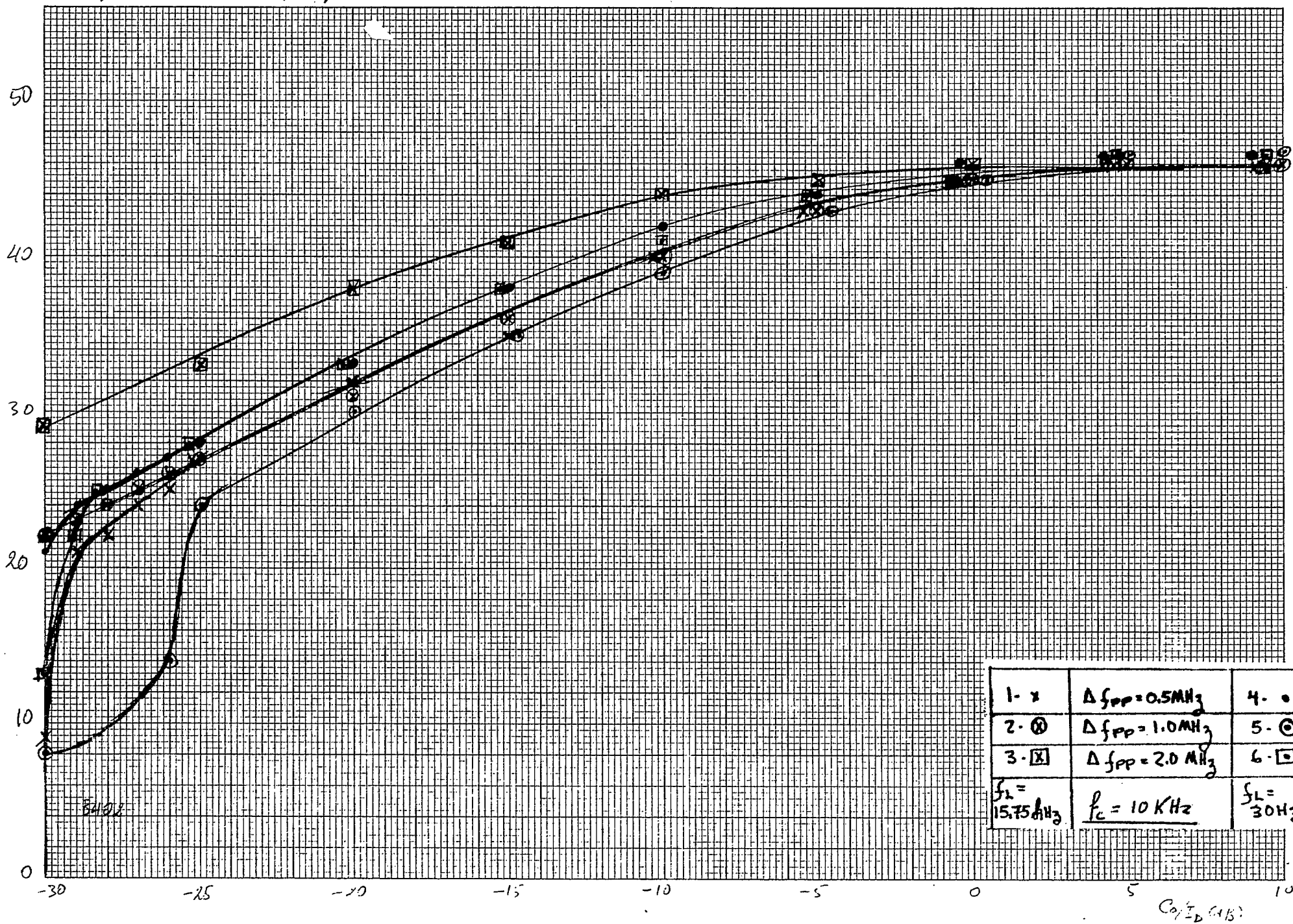
$f_c = 3.0 \text{ MHz}$

1. x	$\Delta f_{PP} = 0.5 \text{ MHz}$	4. o
2. ⊗	$\Delta f_{PP} = 1.0 \text{ MHz}$	5. ⊙
3. ⊠	$\Delta f_{PP} = 2.0 \text{ MHz}$	6. ⊡
$f_L = 15.75 \text{ MHz}$		$f_L = 30 \text{ MHz}$

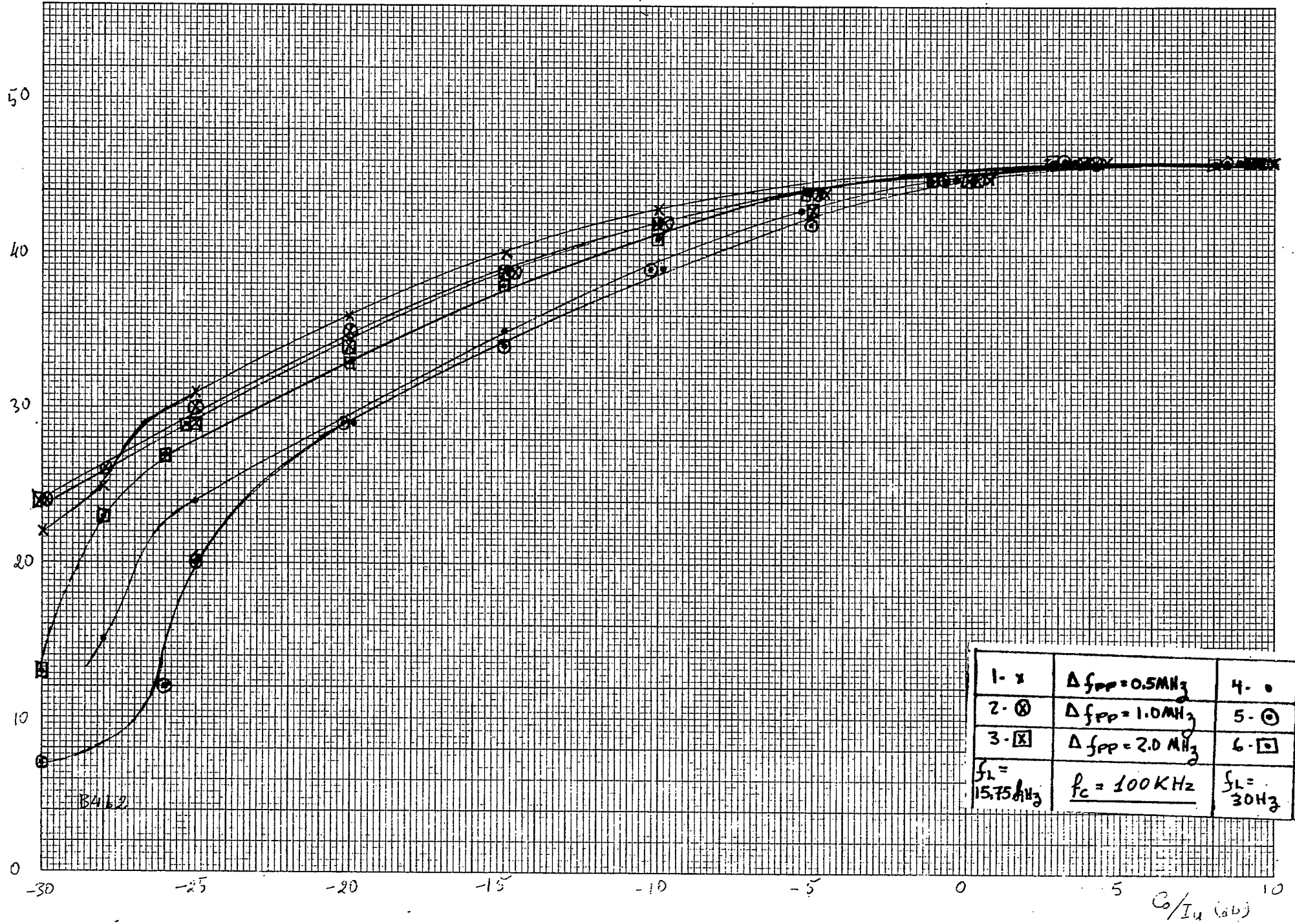
B4 fr

C_0/I_D

$$S/N \text{ (dB)} = 88 \text{ dB} - \text{Threshold (dB)} \text{ (BRN)}$$

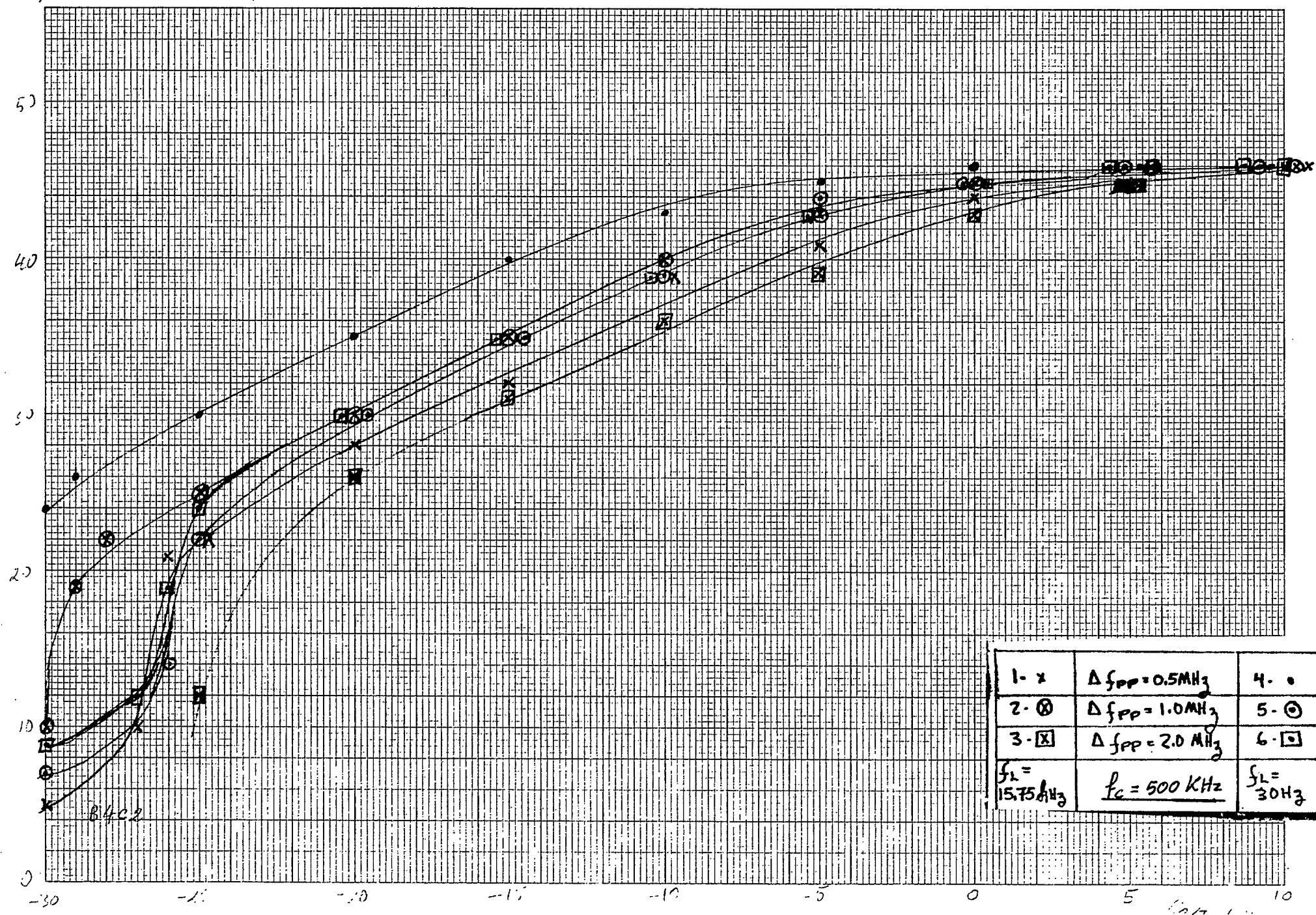


S/N (dB) = 88.8 - T_h (dBm)



1 - x	$\Delta f_{PP} = 0.5 \text{ MHz}$	4 - o
2 - ⊗	$\Delta f_{PP} = 1.0 \text{ MHz}$	5 - ⊙
3 - ⊠	$\Delta f_{PP} = 2.0 \text{ MHz}$	6 - □
$f_L = 15.75 \text{ MHz}$	$f_c = 100 \text{ KHz}$	$f_L = 30 \text{ Hz}$

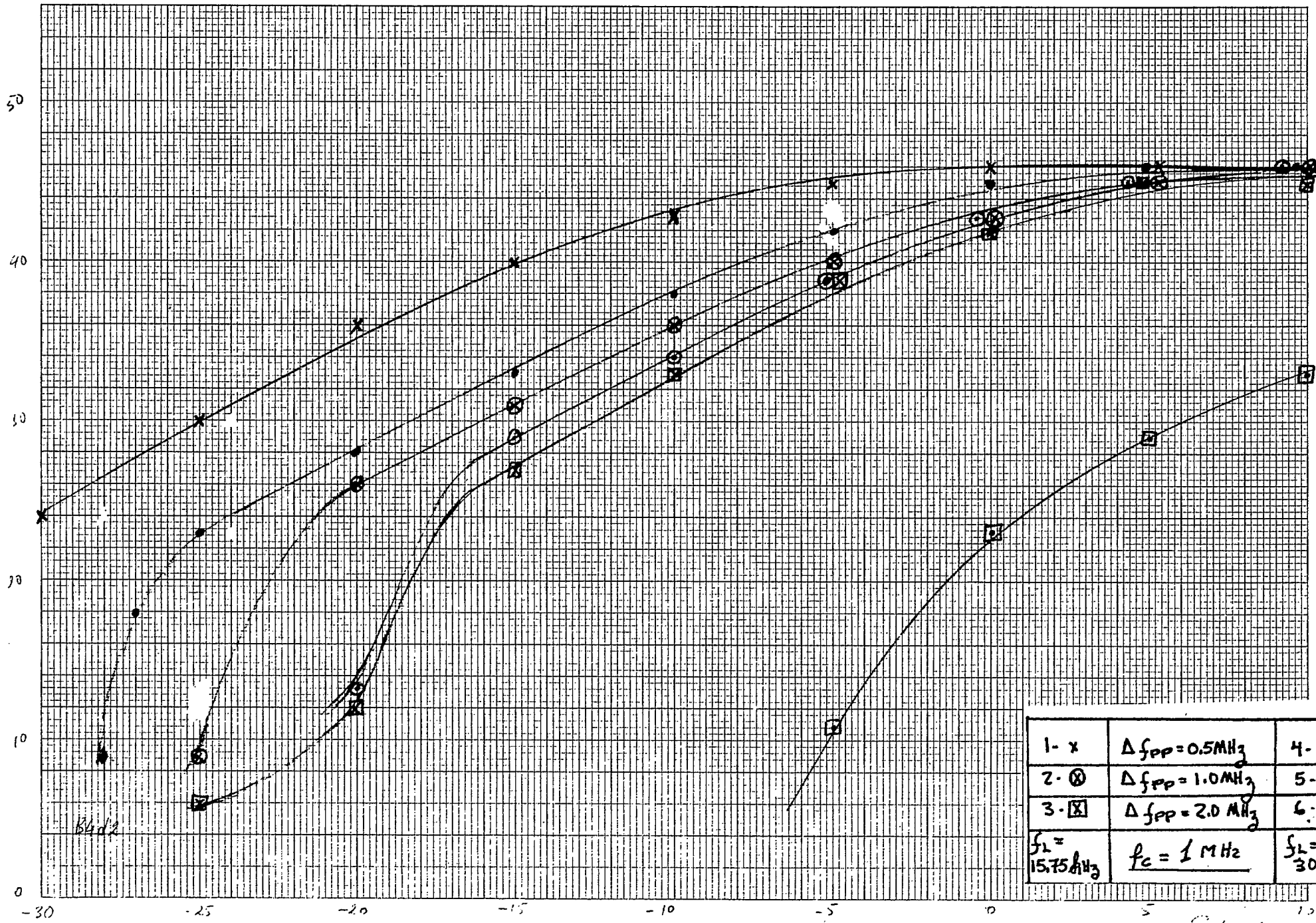
$S/N (dB) = 88$ to 76 (dB)



1 - x	$\Delta f_{PP} = 0.5 \text{ MHz}$	4 - •
2 - ⊗	$\Delta f_{PP} = 1.0 \text{ MHz}$	5 - ⊙
3 - ⊠	$\Delta f_{PP} = 2.0 \text{ MHz}$	6 - ⊡
$f_L = 15.75 \text{ MHz}$	$f_c = 500 \text{ KHz}$	$f_L = 30 \text{ Hz}$

$10/10$

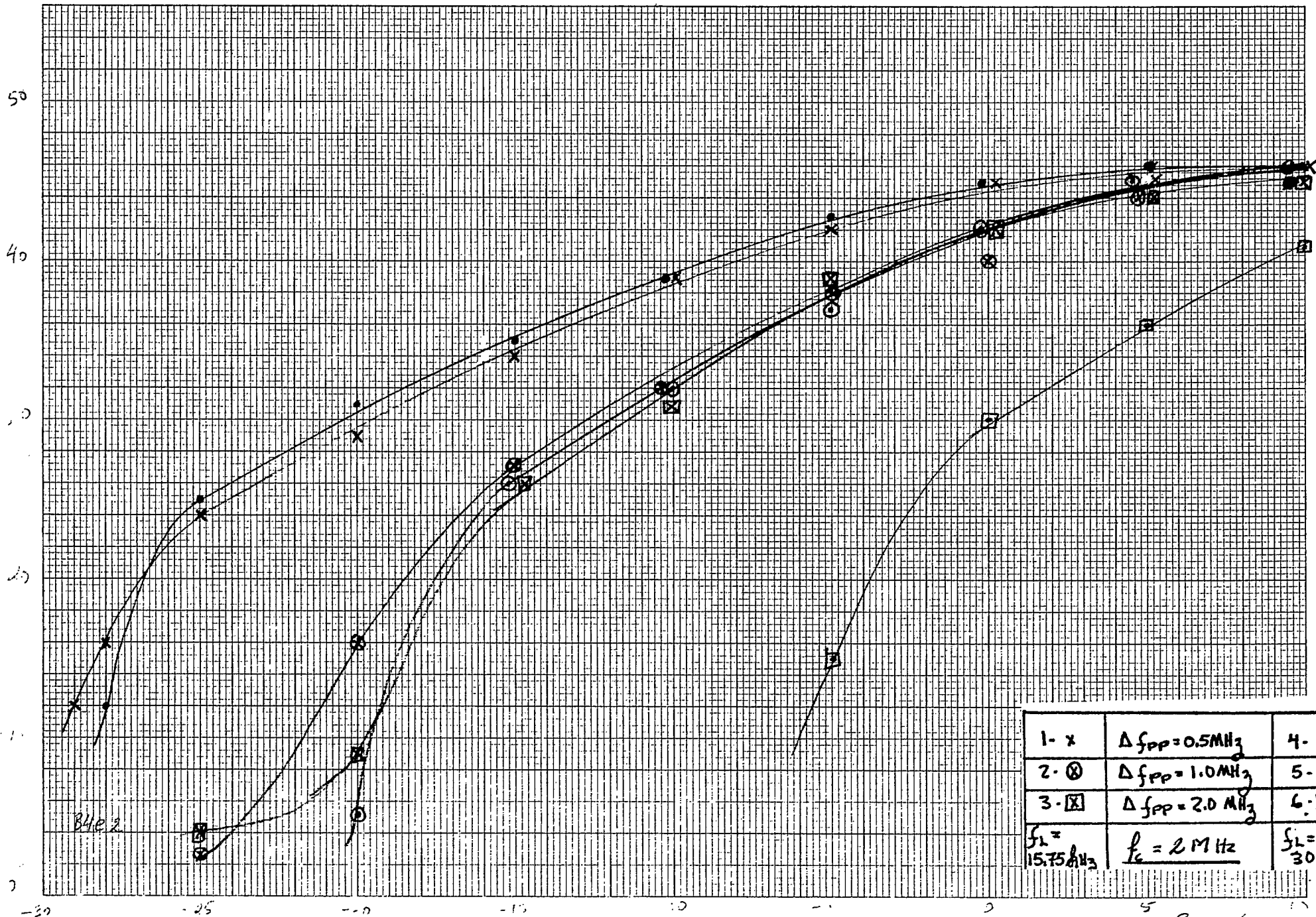
$S/N (dB) = 88 - T_n (dB)$



1- x	$\Delta f_{pp} = 0.5 \text{ MHz}$	4- •
2- ⊗	$\Delta f_{pp} = 1.0 \text{ MHz}$	5- ⊙
3- ⊠	$\Delta f_{pp} = 2.0 \text{ MHz}$	6- □
$f_L = 15.75 \text{ MHz}$	$f_c = 1 \text{ MHz}$	$f_L = 30 \text{ Hz}$

$T_n (dB)$

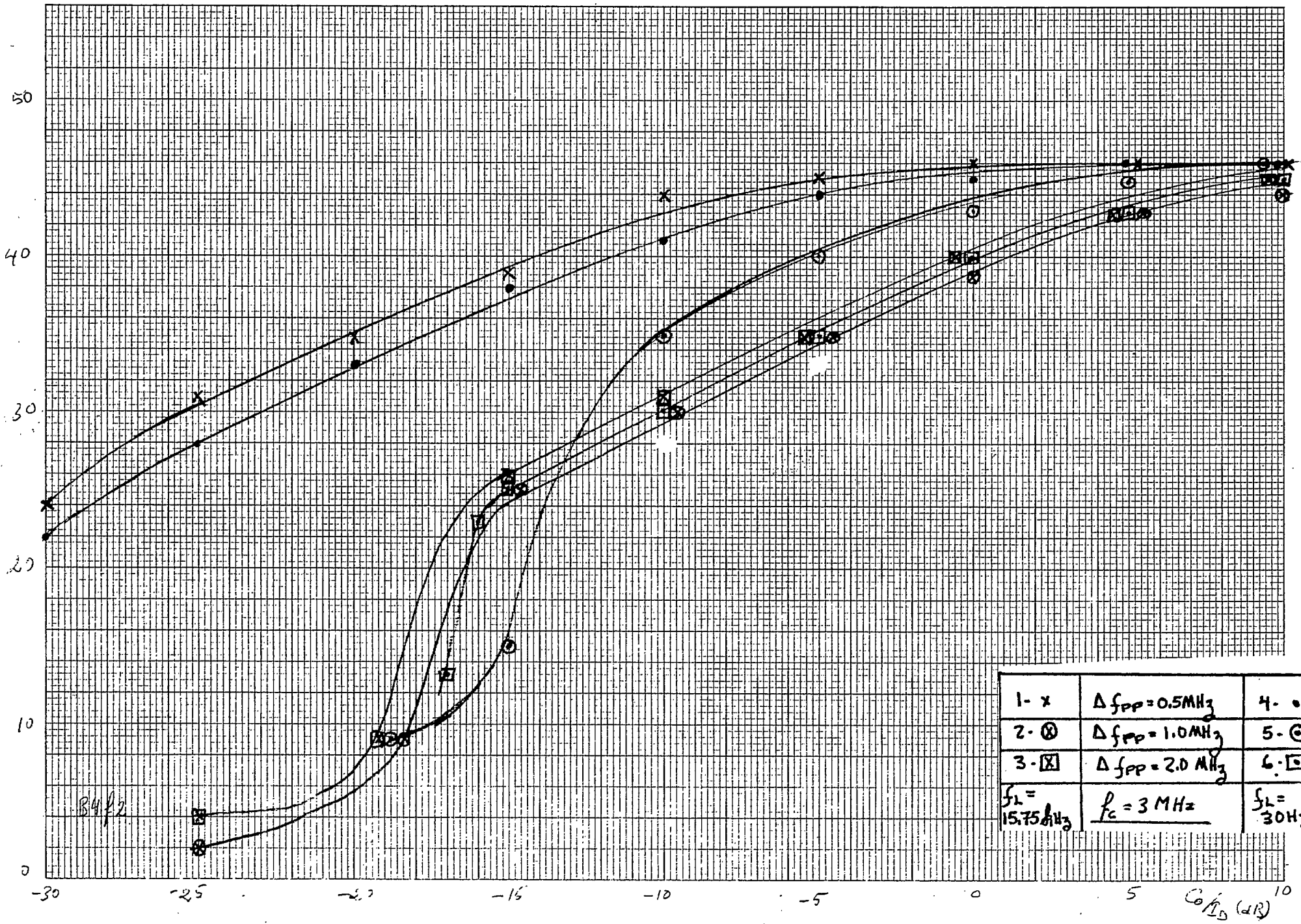
$S/N (dB) = 89 \text{ dB} - \gamma; (\text{dB RN})$



1 - x	$\Delta f_{pp} = 0.5 \text{ MHz}$	4 - •
2 - ⊗	$\Delta f_{pp} = 1.0 \text{ MHz}$	5 - ⊙
3 - ⊠	$\Delta f_{pp} = 2.0 \text{ MHz}$	6 - □
$f_L = 15.75 \text{ MHz}$	$f_c = 2 \text{ MHz}$	$f_L = 30 \text{ Hz}$

Com20 (dB)

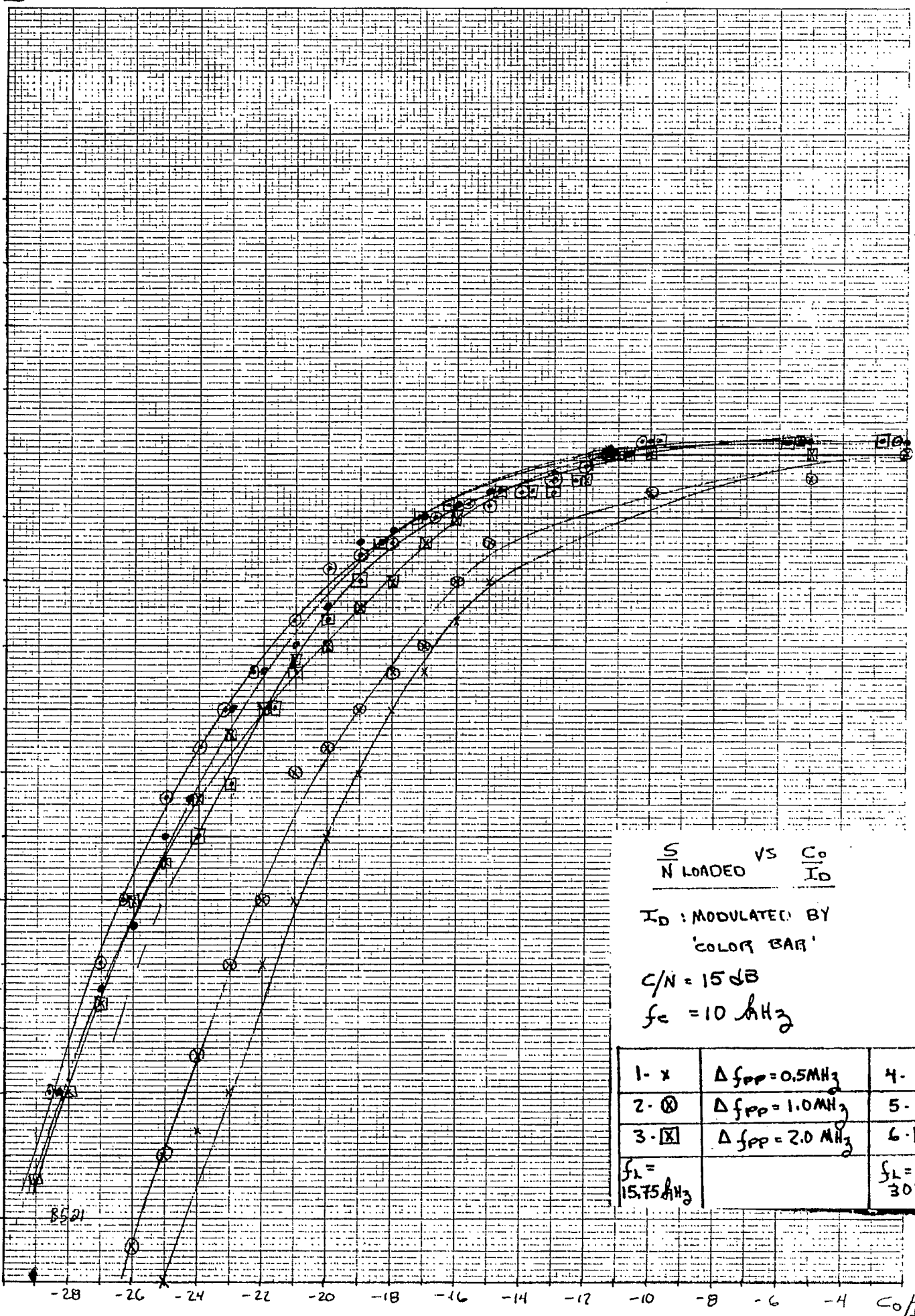
S.N (dB) = 88(dB) - Tn (dB RN)



B5

$\approx N_{LOADED}$

39
38
37
36
35
34
33
32
31
30.5



$\frac{S}{N_{LOADED}}$ VS $\frac{C_0}{I_0}$

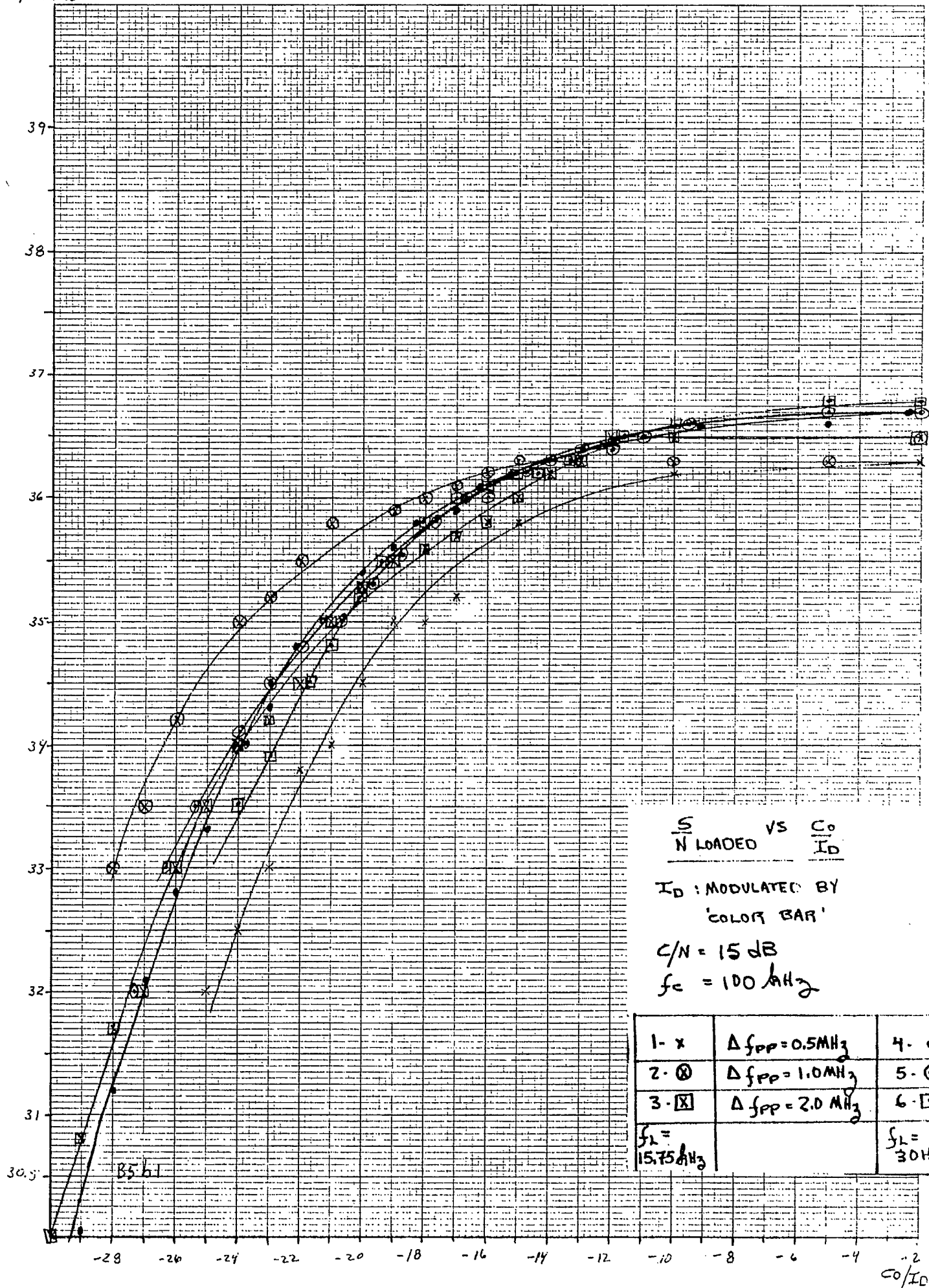
I_D : MODULATED BY 'COLOR BAR'

$C/N = 15 \text{ dB}$
 $f_c = 10 \text{ MHz}$

1. x	$\Delta f_{PP} = 0.5 \text{ MHz}$	4. •
2. ⊗	$\Delta f_{PP} = 1.0 \text{ MHz}$	5. ⊙
3. ⊠	$\Delta f_{PP} = 2.0 \text{ MHz}$	6. ⊡
$f_L = 15.75 \text{ MHz}$		$f_L = 30 \text{ MHz}$

C_0/I_0

S/N LOADED



$\frac{S}{N} \text{ LOADED VS } \frac{C_0}{I_D}$

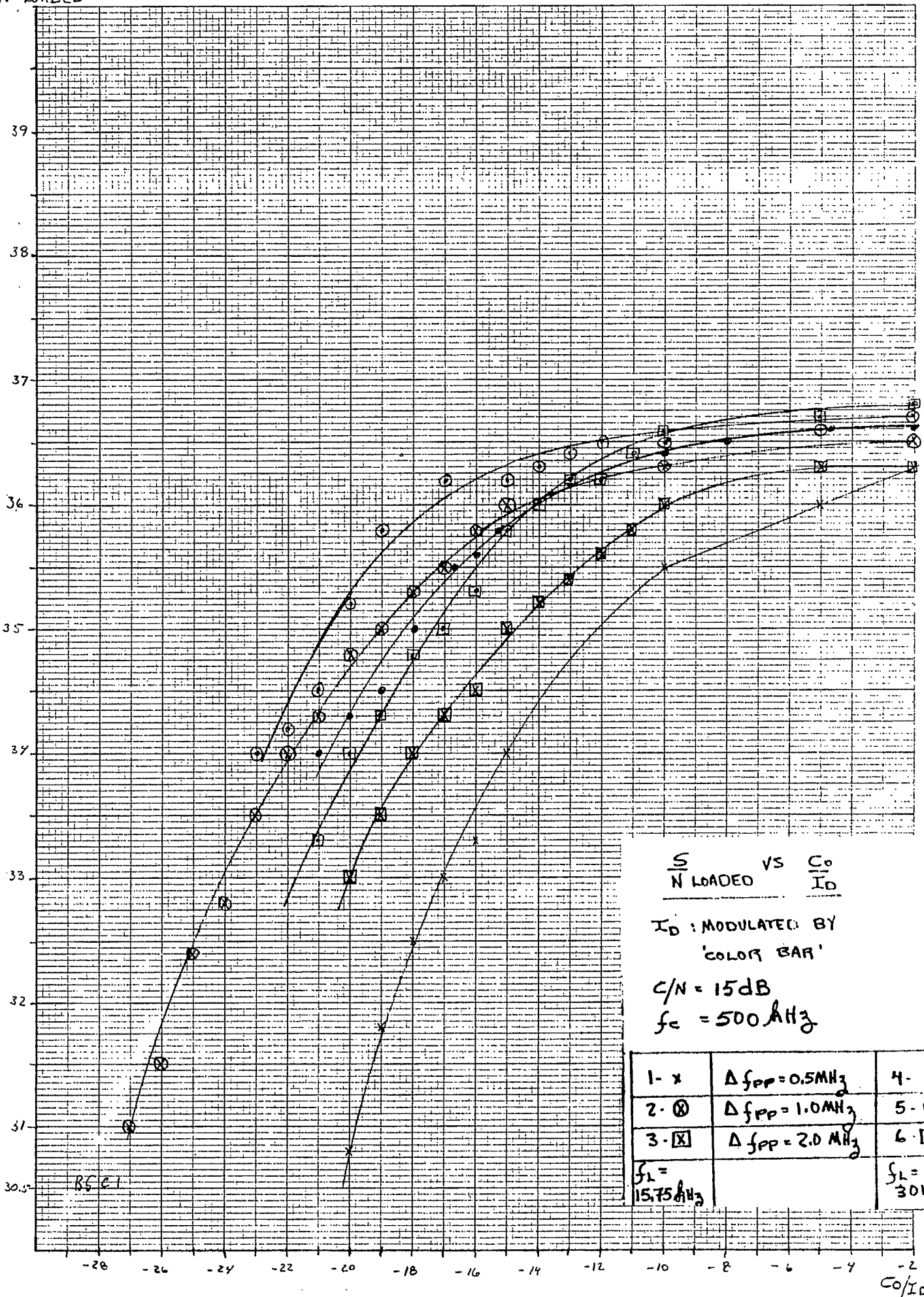
I_D : MODULATED BY 'COLOR BAR'

$C/N = 15 \text{ dB}$
 $f_c = 100 \text{ MHz}$

1-x	$\Delta f_{pp} = 0.5 \text{ MHz}$	4-•
2-⊗	$\Delta f_{pp} = 1.0 \text{ MHz}$	5-⊙
3-⊠	$\Delta f_{pp} = 2.0 \text{ MHz}$	6-⊡
$f_L = 15.75 \text{ MHz}$		$f_L = 30 \text{ Hz}$

$\frac{C_0}{I_D}$

S/N LOADED



S/N LOADED

$\frac{S}{N \text{ LOADED}}$ VS $\frac{C_0}{I_D}$

I_D : MODULATED BY
'COLOR BAR'

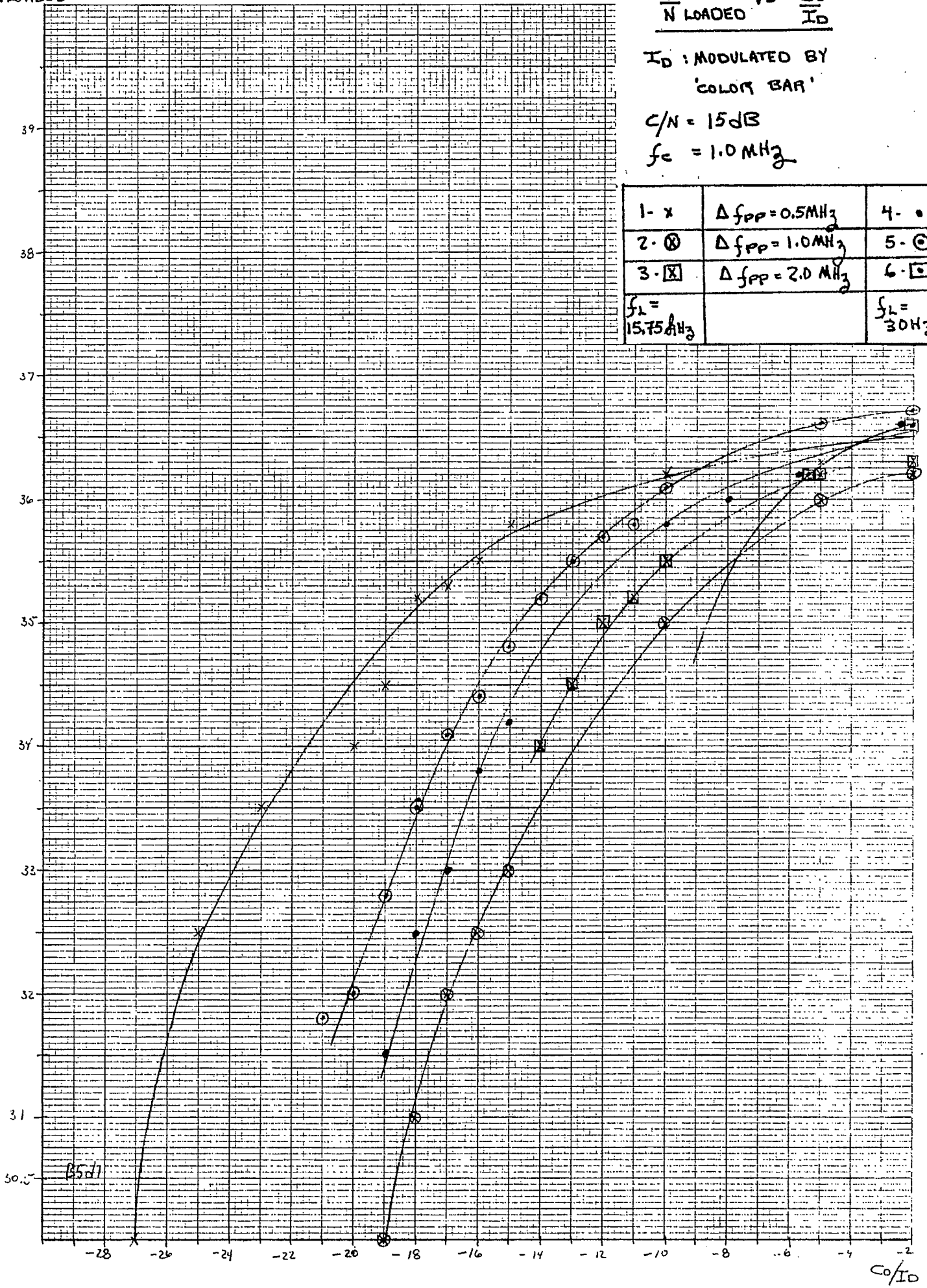
$C/N = 15 \text{ dB}$

$f_c = 1.0 \text{ MHz}$

1 - x	$\Delta f_{pp} = 0.5 \text{ MHz}$	4 - •
2 - ⊗	$\Delta f_{pp} = 1.0 \text{ MHz}$	5 - ⊙
3 - ⊠	$\Delta f_{pp} = 2.0 \text{ MHz}$	6 - ⊡
$f_L = 15.75 \text{ MHz}$		$f_L = 30 \text{ MHz}$

46 1331

10 X 10 INCH KEUFFEL & ESSER CO. MADE IN U.S.A.



B5d1

C_0/I_D

S/N LOADED

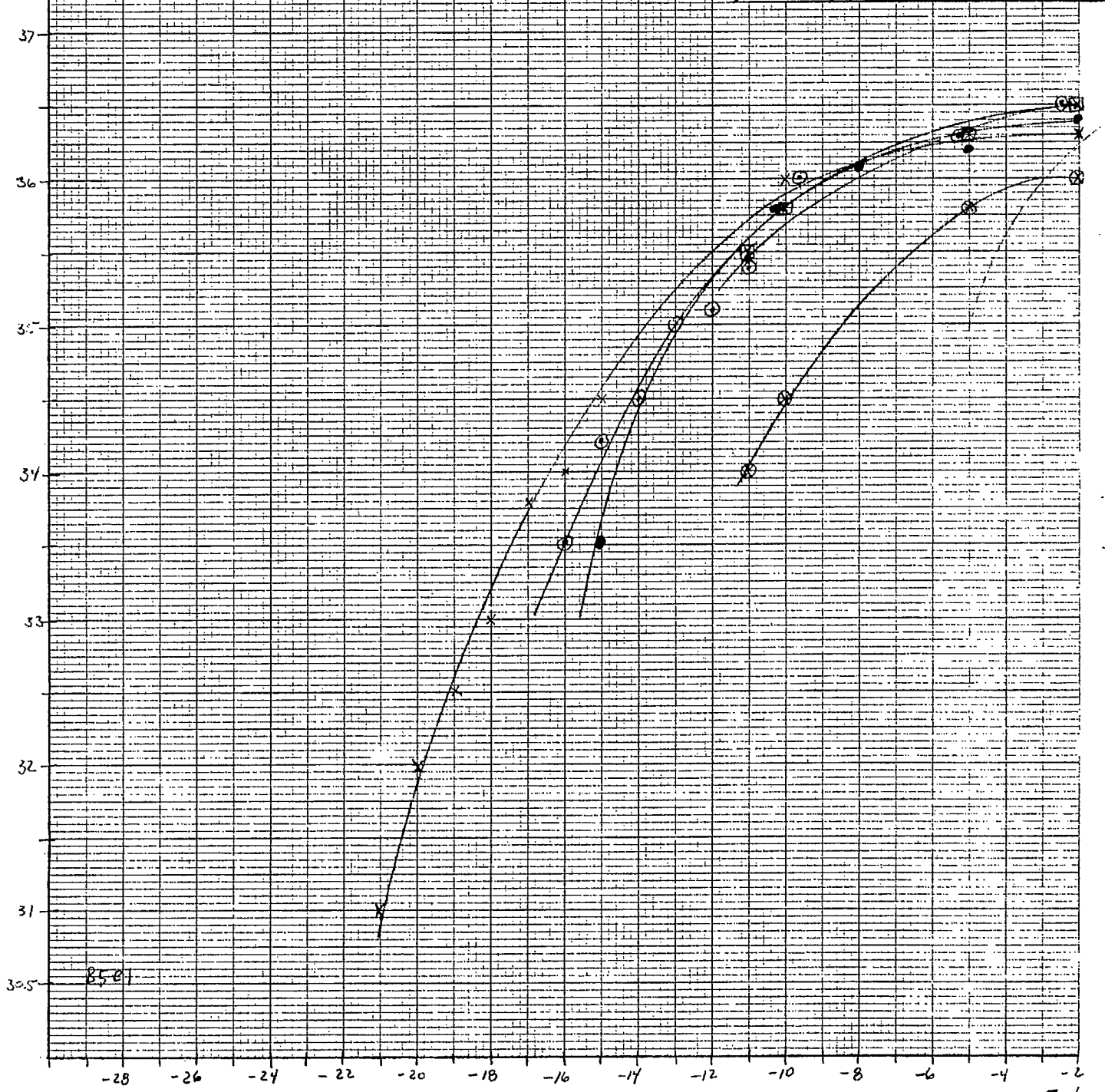
$\frac{S}{N}$ LOADED VS $\frac{C_0}{I_0}$

I_0 : MODULATED BY
'COLOR BAR'

$C/N = 15 \text{ dB}$

$f_c = 2 \text{ MHz}$

1 - x	$\Delta f_{PP} = 0.5 \text{ MHz}$	4 - o
2 - ⊗	$\Delta f_{PP} = 1.0 \text{ MHz}$	5 - ⊙
3 - ⊠	$\Delta f_{PP} = 2.0 \text{ MHz}$	6 - ⊡
$f_L = 15.75 \text{ MHz}$		$f_L = 30 \text{ MHz}$



1331
 NEUFEL & ESSER CO. MADE IN U.S.A.

C_0/I_0

S/N LOADED

$\frac{S}{N \text{ LOADED}} \text{ VS } \frac{C_0}{I_0}$

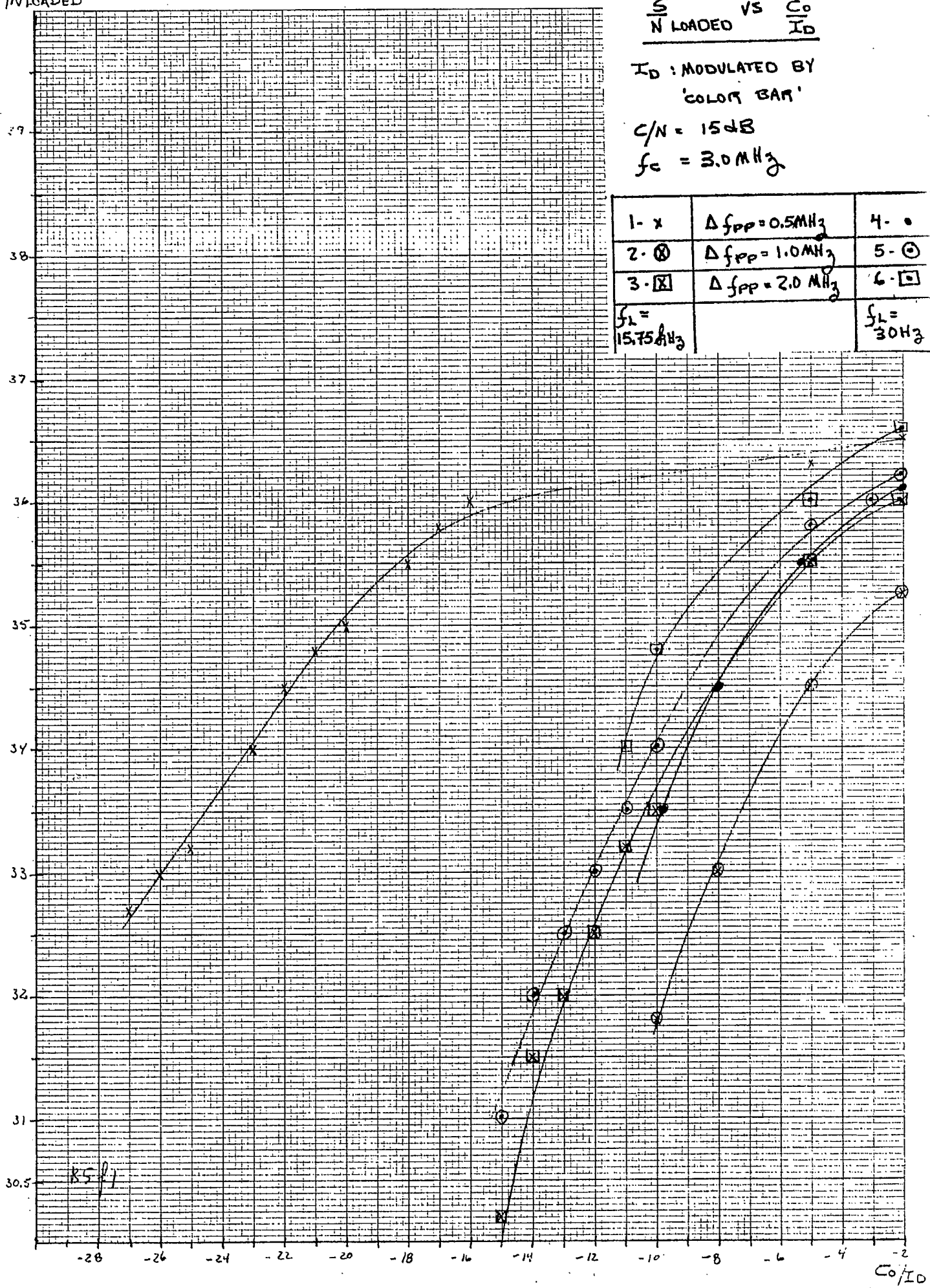
I_0 : MODULATED BY 'COLOR BAR'

$C/N = 15 \text{ dB}$

$f_c = 3.0 \text{ MHz}$

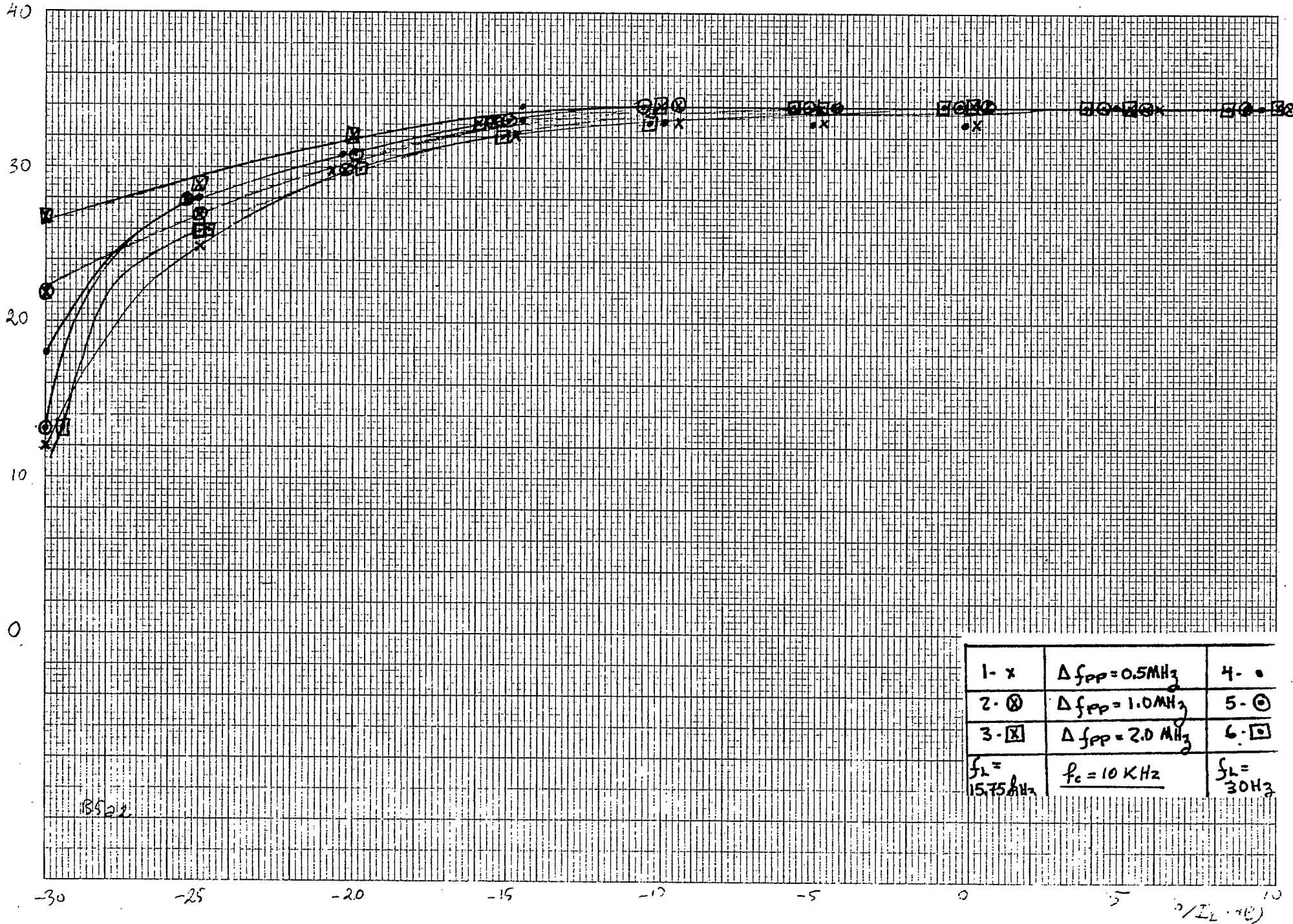
1 - x	$\Delta f_{PP} = 0.5 \text{ MHz}$	4 - o
2 - ⊗	$\Delta f_{PP} = 1.0 \text{ MHz}$	5 - ⊙
3 - ⊠	$\Delta f_{PP} = 2.0 \text{ MHz}$	6 - □
$f_L = 15.75 \text{ MHz}$		$f_L = 30 \text{ Hz}$

KEUFFEL & ESSER CO. MADE IN U.S.A.
 BOX 10, 10 1/2 INCH
 46 1331



C_0/I_0

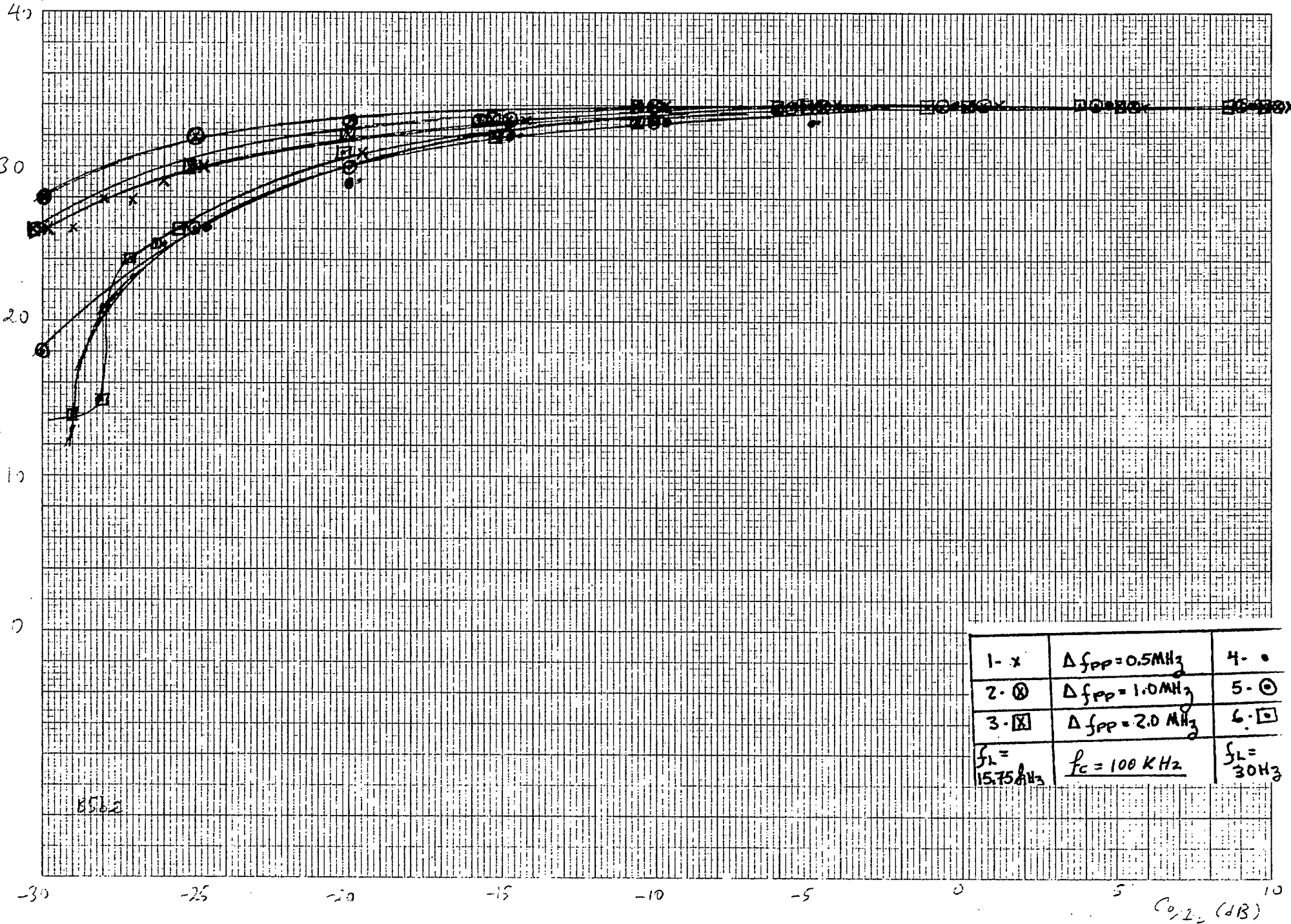
$S/N(dB) = 88(dB) - T_w(dB/N)$



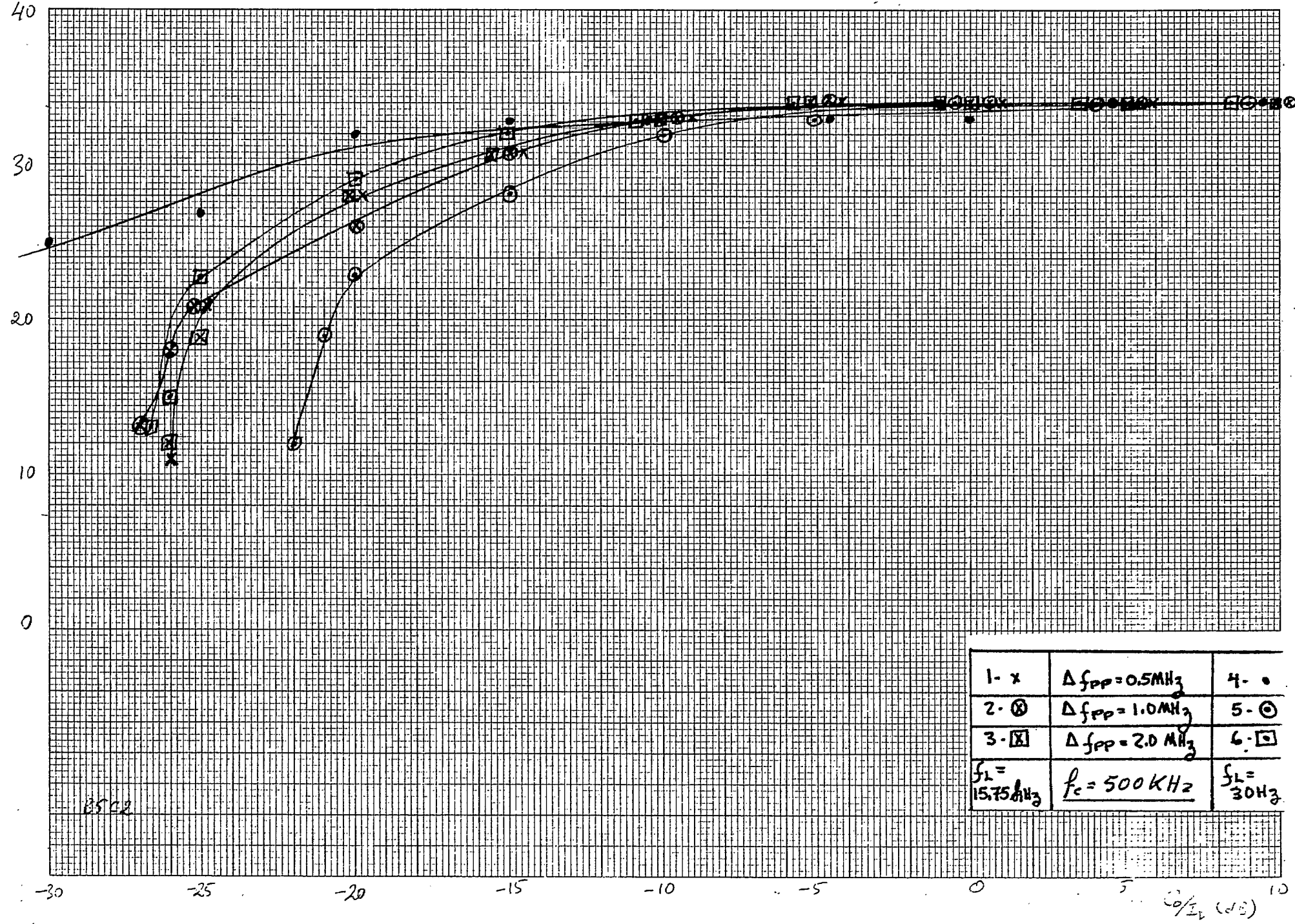
185212

$T_w(dB/N)$

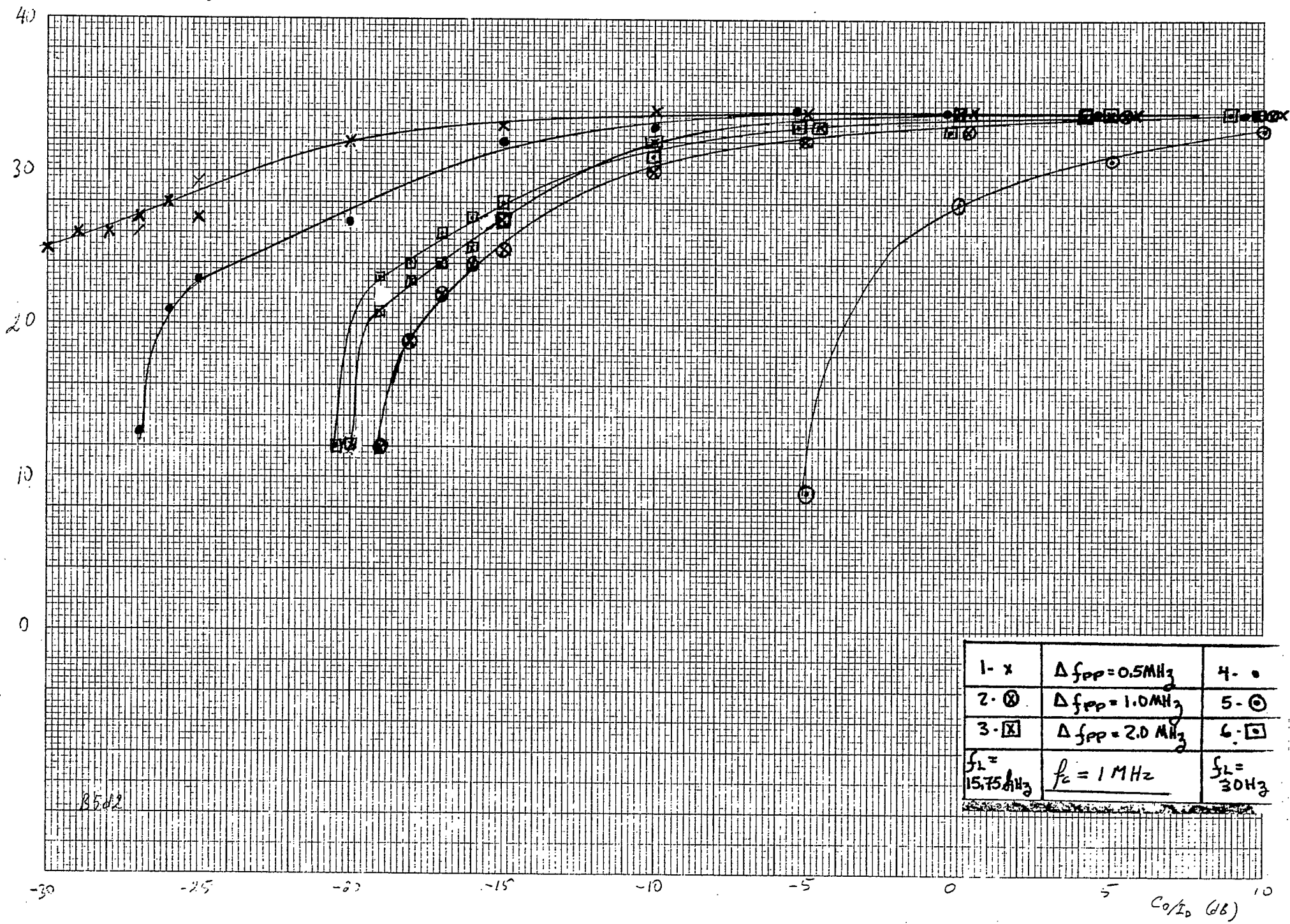
$S/N (dB) = 88 (dB) - T_n (dBm)$



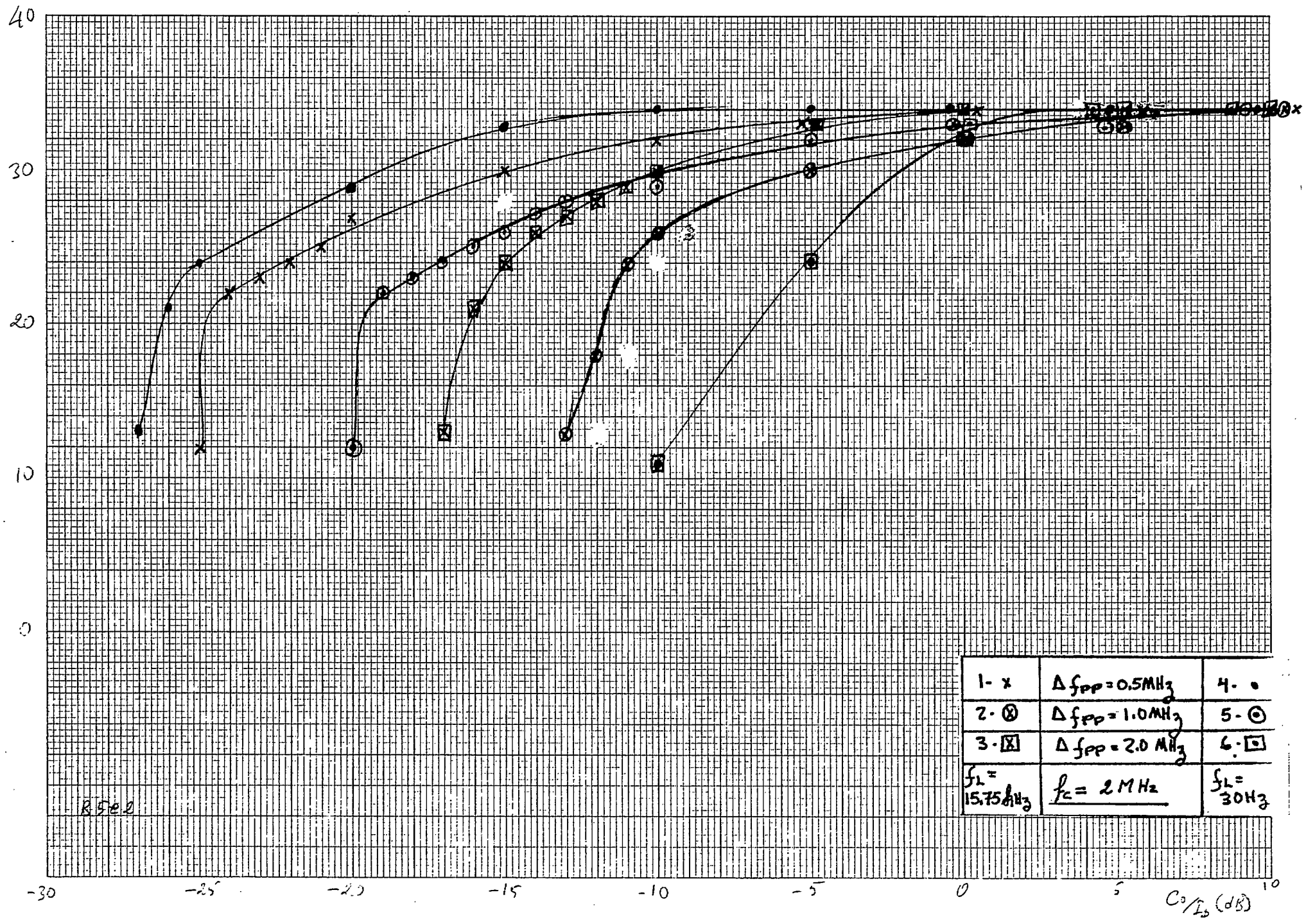
$$S/N(\text{dB}) = 88(\text{dB}) - T_h(\text{dB RN})$$



$$S/N (dB) = 38.46 - T_n (dBK(1))$$

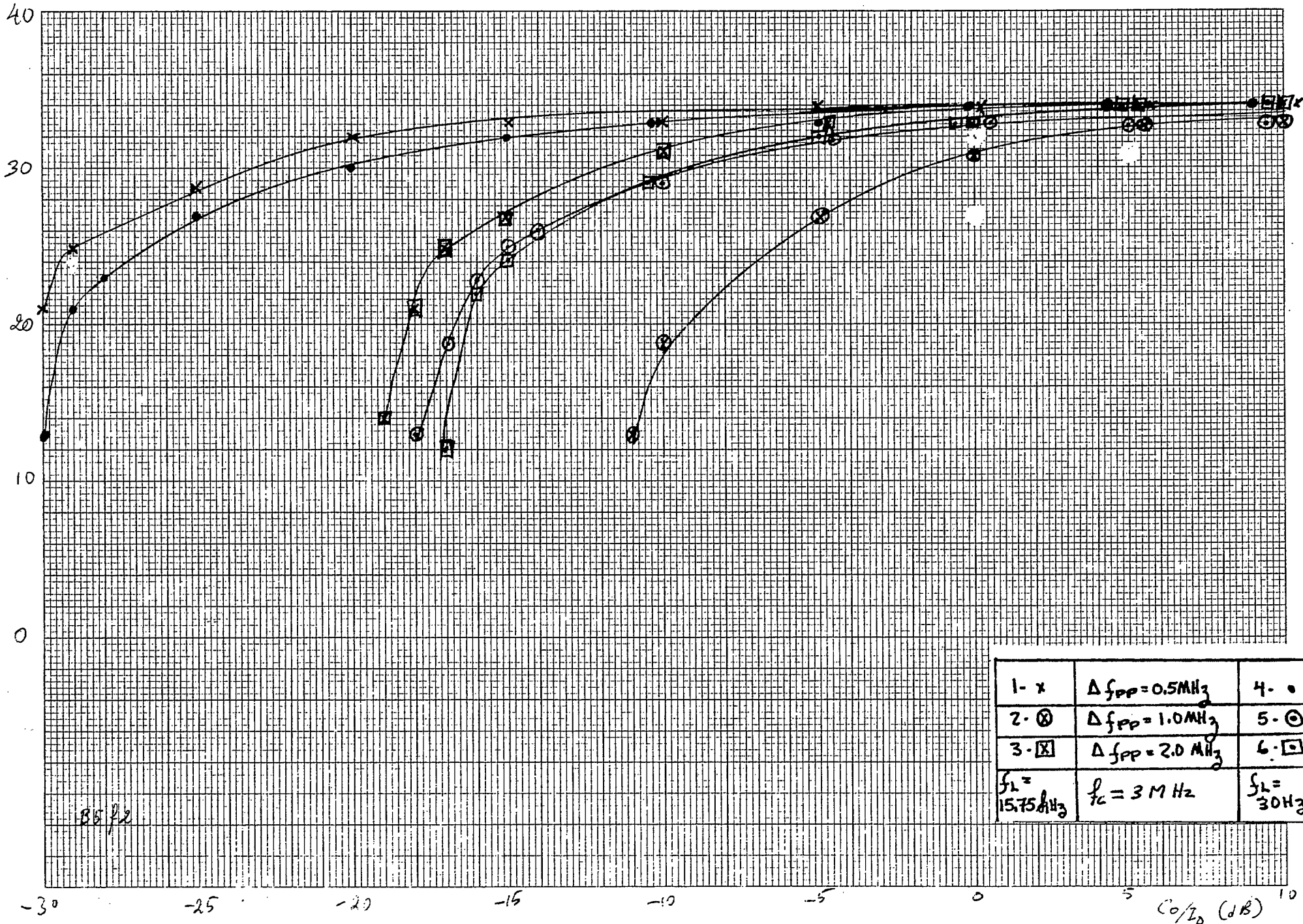


$S/N (dB) = 83 (dB) - T_L (dB/N)$



R500

$$S/N(\text{dB}) = 89(\text{dB}) - T_n(\text{dB RN})$$



B6

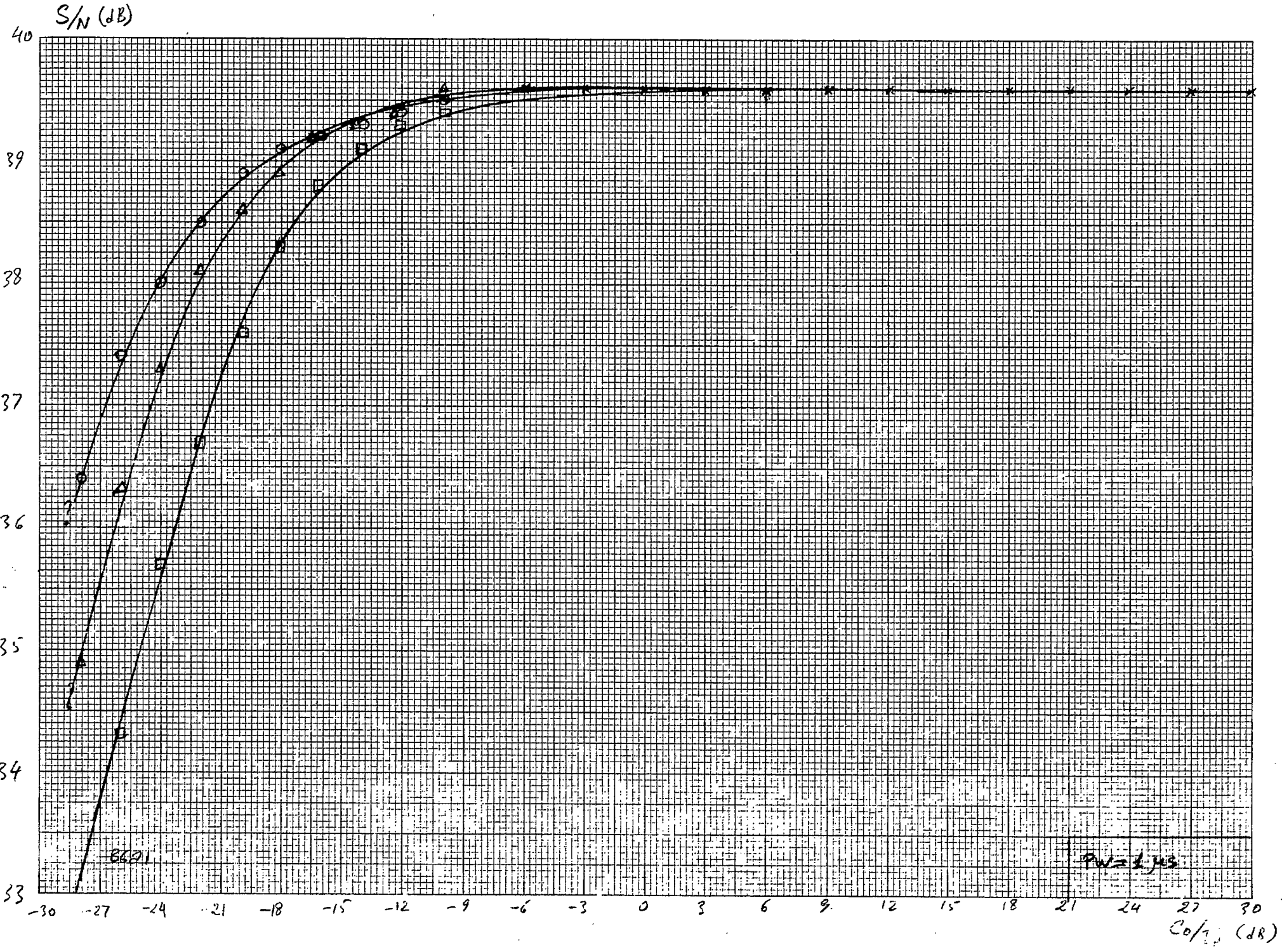
SYMBOL	PRF (kHz)
○	0.3
△	1.0
□	3.0
*	ALL ABOVE

? = ERRATIC READINGS

Radon-Like Interfering signal

S/N → ∞

$f_c = 10 \text{ kHz}$



PR F: $\circ = 0.3 \text{ KHz}$
 $\Delta = 1 \text{ KHz}$
 $\square = 2 \text{ KHz}$ X - ?

S/N (dB)

40

39

38

37

36

35

34

33

-30

-27

-24

-21

-18

-15

-12

-9

-6

-3

0

3

6

9

12

15

18

21

24

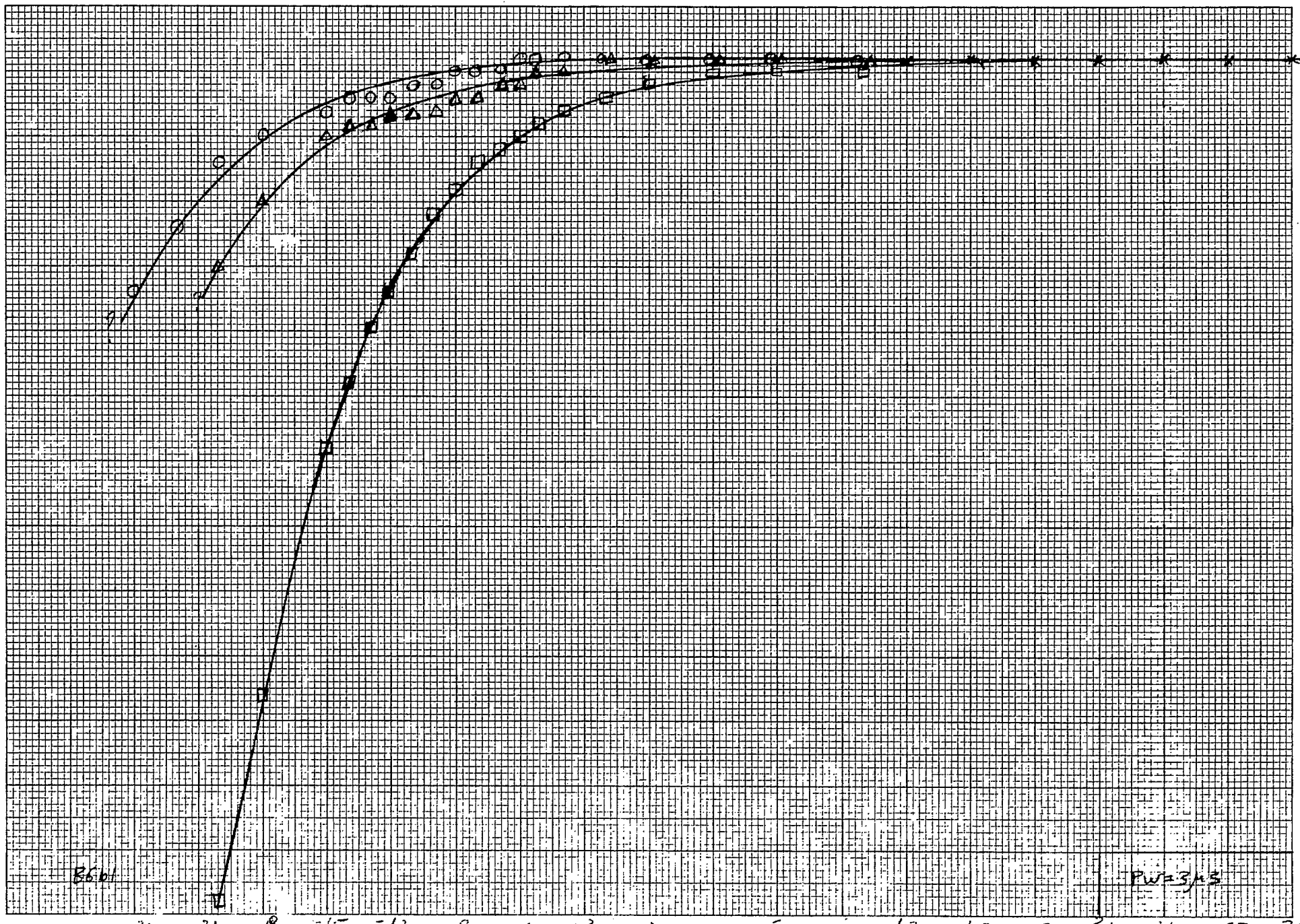
27

30

B6601

PW=3μs

C_0/I_p (dB)



PRR = $\circ = 3\text{KHz}$
 $\Delta = 1\text{KHz}$
 $\square = 3\text{KHz}$ } * = 20

3/N (dB)

40

39

38

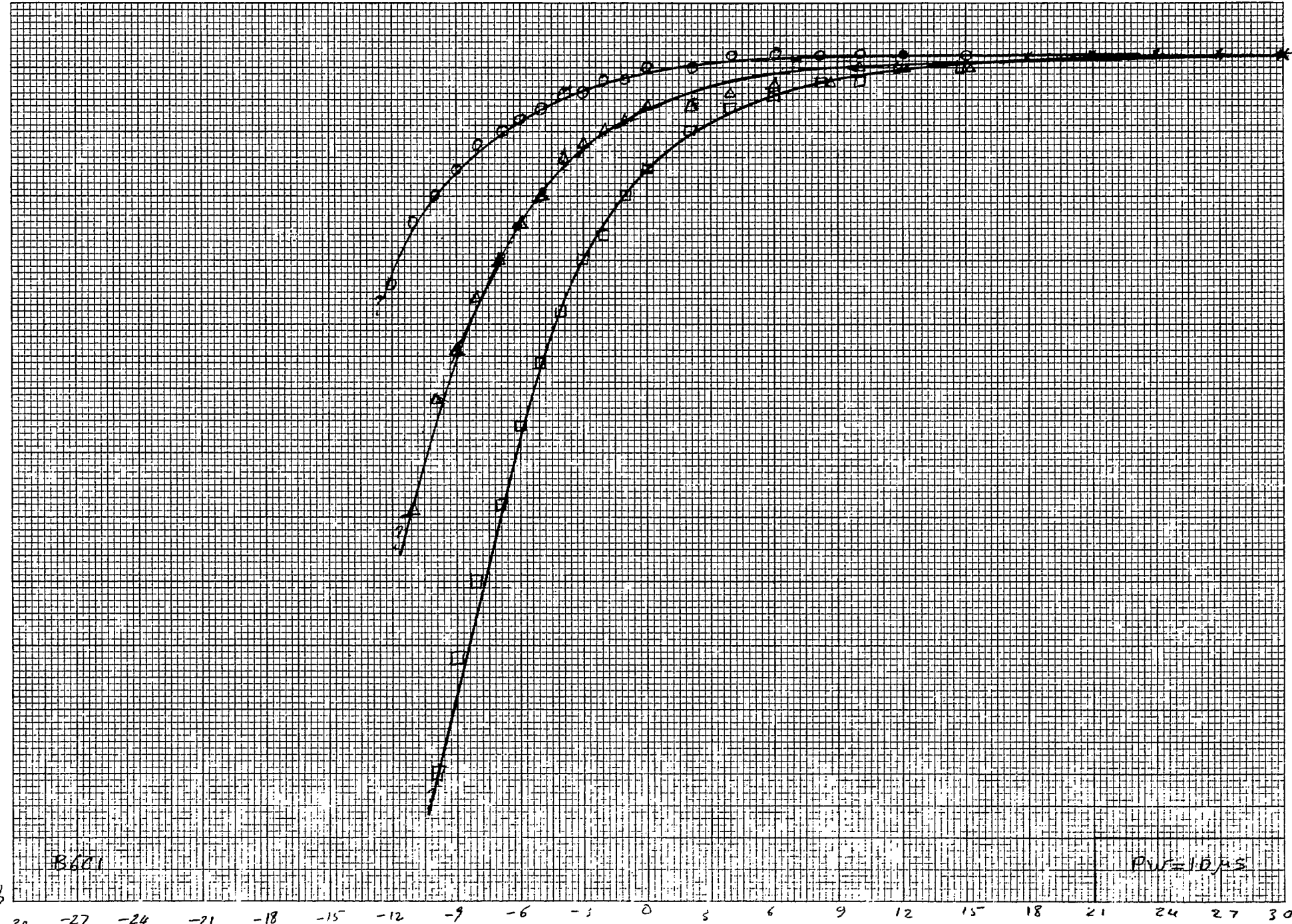
37

36

35

34

33



B601

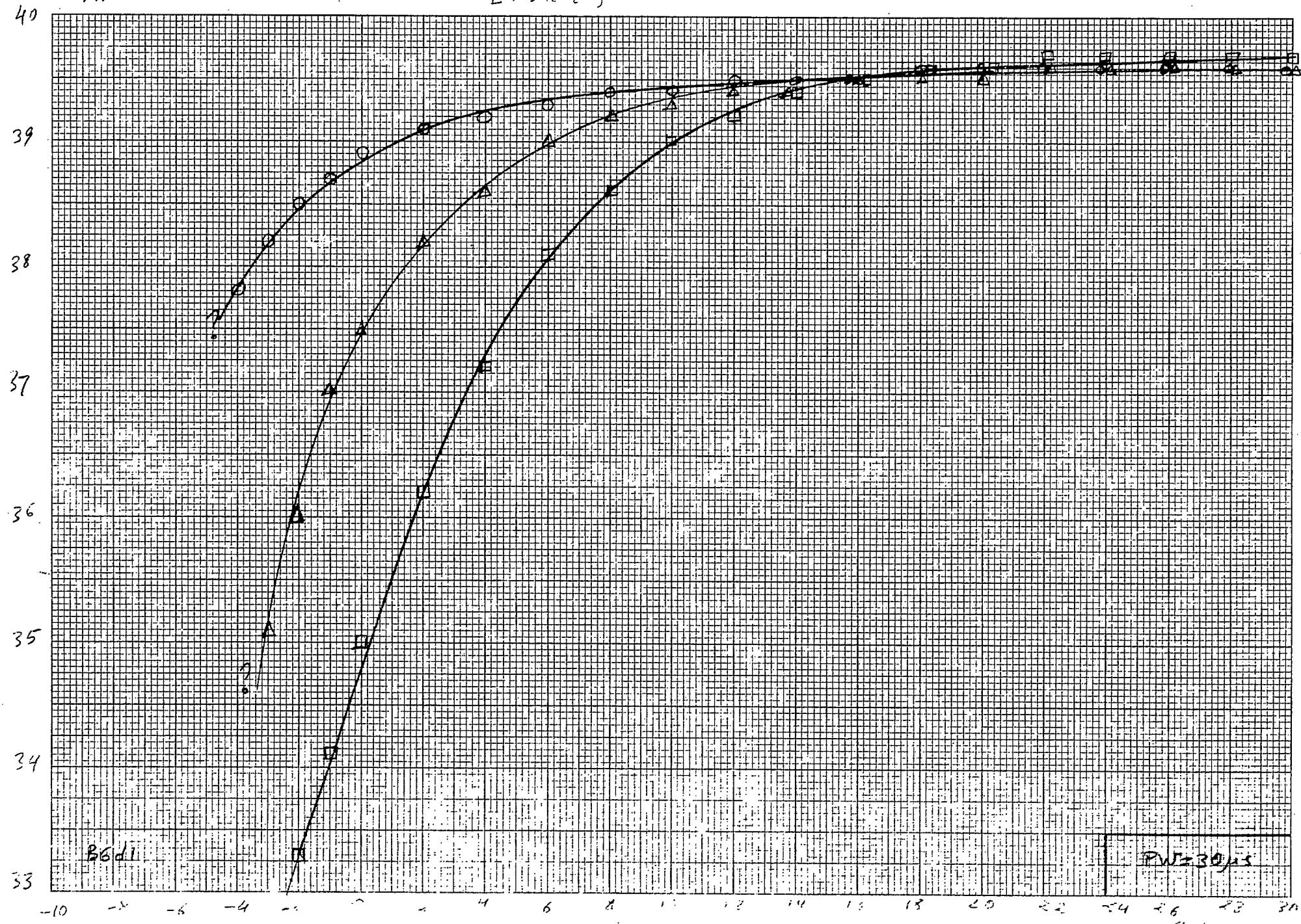
PWR = 10 MS

-30 -27 -24 -21 -18 -15 -12 -9 -6 -3 0 3 6 9 12 15 18 21 24 27 30

Co/Ip (dB)

S/N (dB)

PRF = $\begin{cases} \sigma = 3 \text{ KHz} \\ \Delta = 1 \text{ KHz} \\ \Omega = 3 \text{ KHz} \end{cases} \quad * = \text{all}$



BBd1

PULSE 30 μs

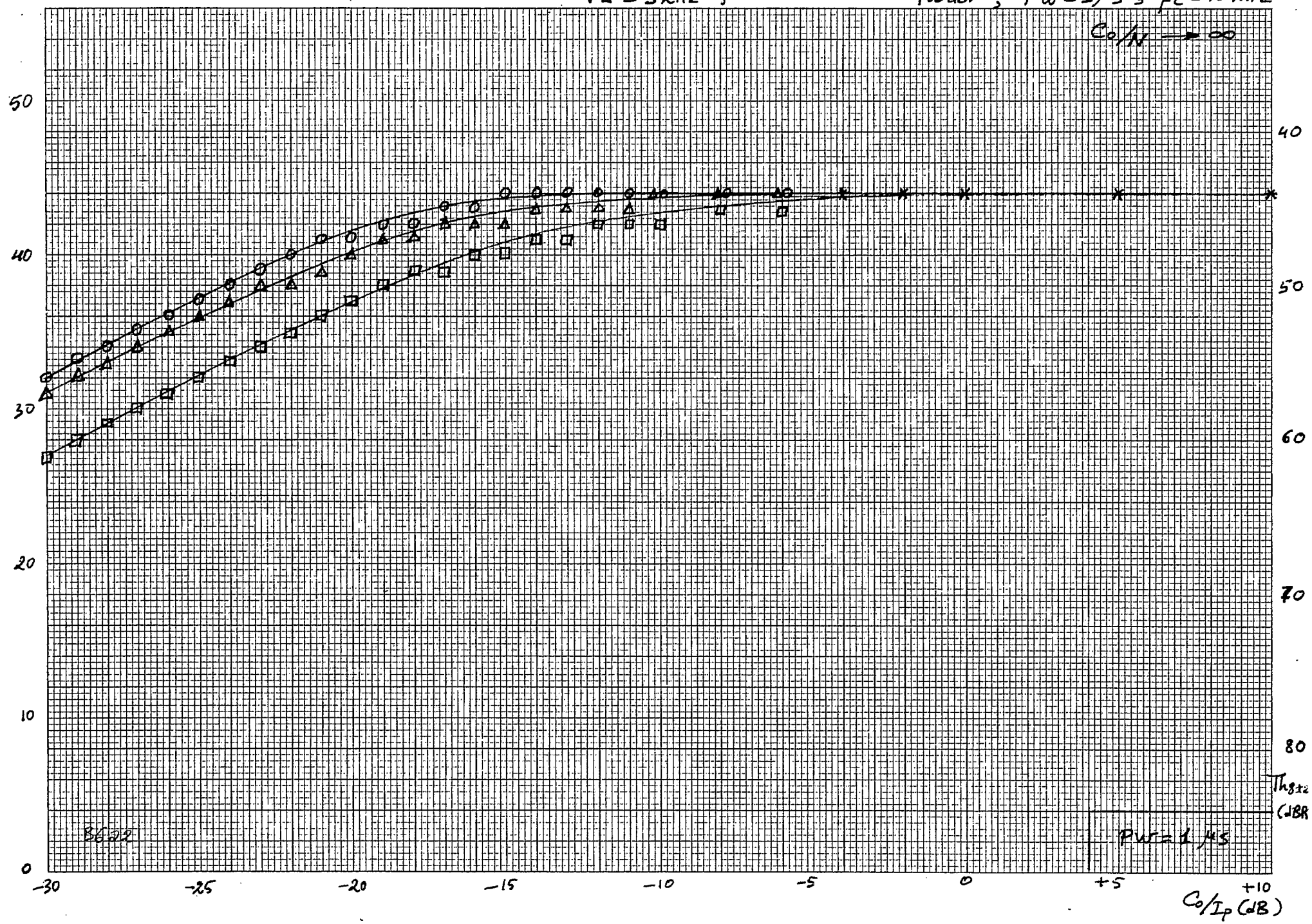
Go/ep (dB)

$$S/N (dB) = 88 (dB) - T_{h_{2+2\mu s}} (dB_{RN})$$

$$PRF = \begin{cases} \circ = 0.3 \text{ KHz} \\ \Delta = 1 \text{ KHz} \\ \square = 3 \text{ KHz} \end{cases} * = \text{all}$$

Radar ; $PW = 1 \mu s$; $f_c = 10 \text{ KHz}$

$C_0/N \rightarrow \infty$

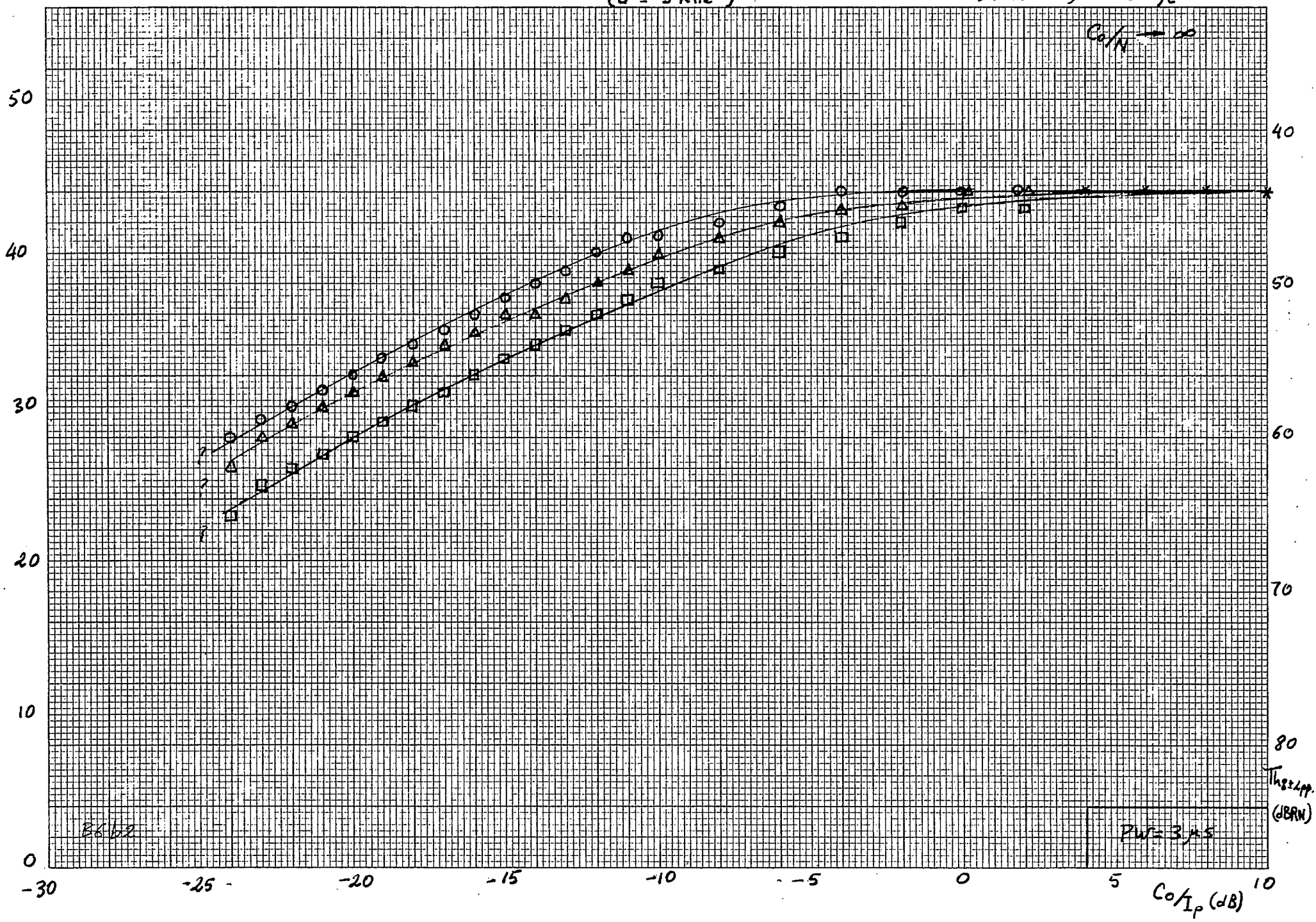


$$S/N \text{ (dB)} = 88 \text{ (dB)} - T_{\text{Thres}} \text{ (dB)} - T_{\text{Thres}} \text{ (dB)}$$

$$PRF = \begin{cases} \circ = .3 \text{ KHz} \\ \Delta = 1 \text{ KHz} \\ \square = 3 \text{ KHz} \end{cases} * = \text{all}$$

Radar ; PW = 3 μs ; f_c = 10 KHz

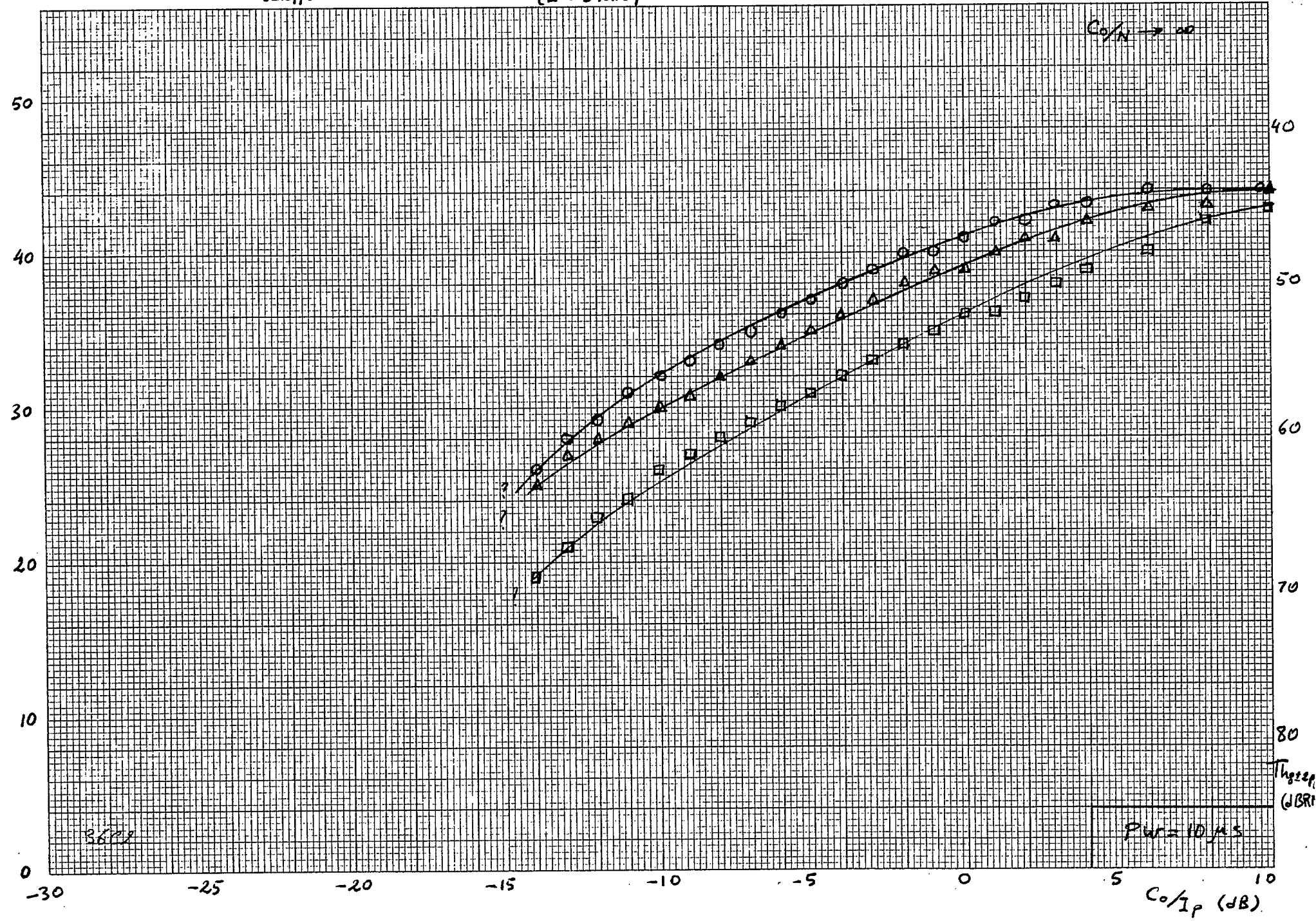
C₀/N = 10



$$S/N \text{ (dB)} = 88 \text{ (dB)} - T_{hg \pm 2pps} \text{ (dB RN)}$$

$$PRF = \begin{cases} \sigma = 3 \text{ KHz} \\ \Delta = 1 \text{ KHz} \\ \square = 3 \text{ KHz} \end{cases}$$

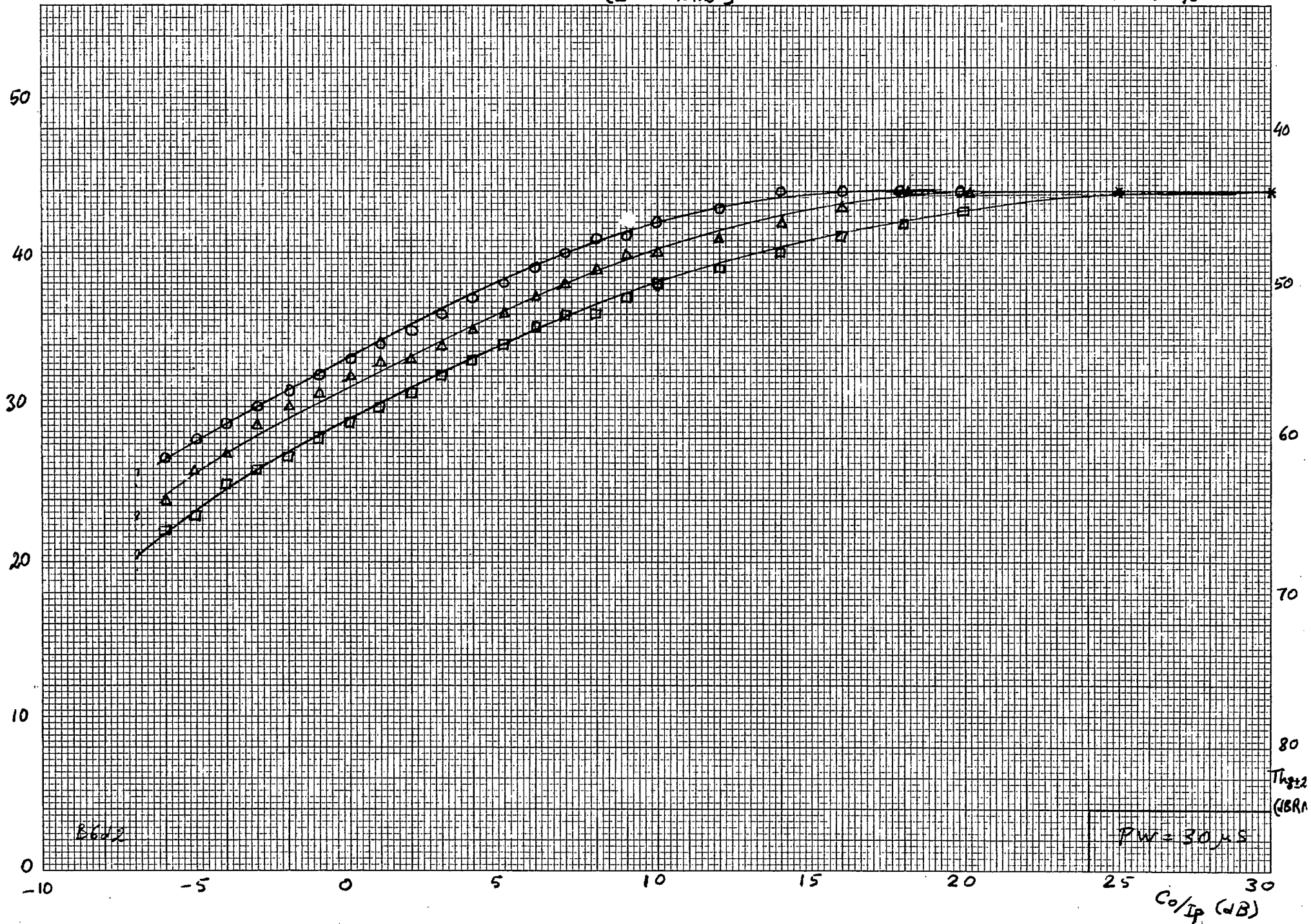
Radar ; $PW = 10 \mu s$; $f_c = 10 \text{ KHz}$



$S/N (dB) = 88 (dB) - T_{h8 \pm 2pps} (dB RN)$

PRF = $\left\{ \begin{array}{l} \circ = .3 \text{ KHz} \\ \Delta = 1 \text{ KHz} \\ \square = 3 \text{ KHz} \end{array} \right\} * = \text{all}$

Radar ; PW = 30 μ s ; $f_c = 10 \text{ KHz}$



B7

SYMBOL	PRF (kHz)
o	0.3
Δ	1.0
□	3.0
*	ALL ABOVE

? = ERRATIC READINGS

Redor-Like Interfering signal

$$\underline{C_0/N = 15 \text{ dB}}$$

$$\underline{f_c = 10 \text{ kHz}}$$

PRF = $\left. \begin{array}{l} \circ = 3 \text{ KHz} \\ \triangle = 1 \text{ KHz} * = 211 \\ \square = 3 \text{ KHz} \end{array} \right\}$

S/N (dB)

40

39

38

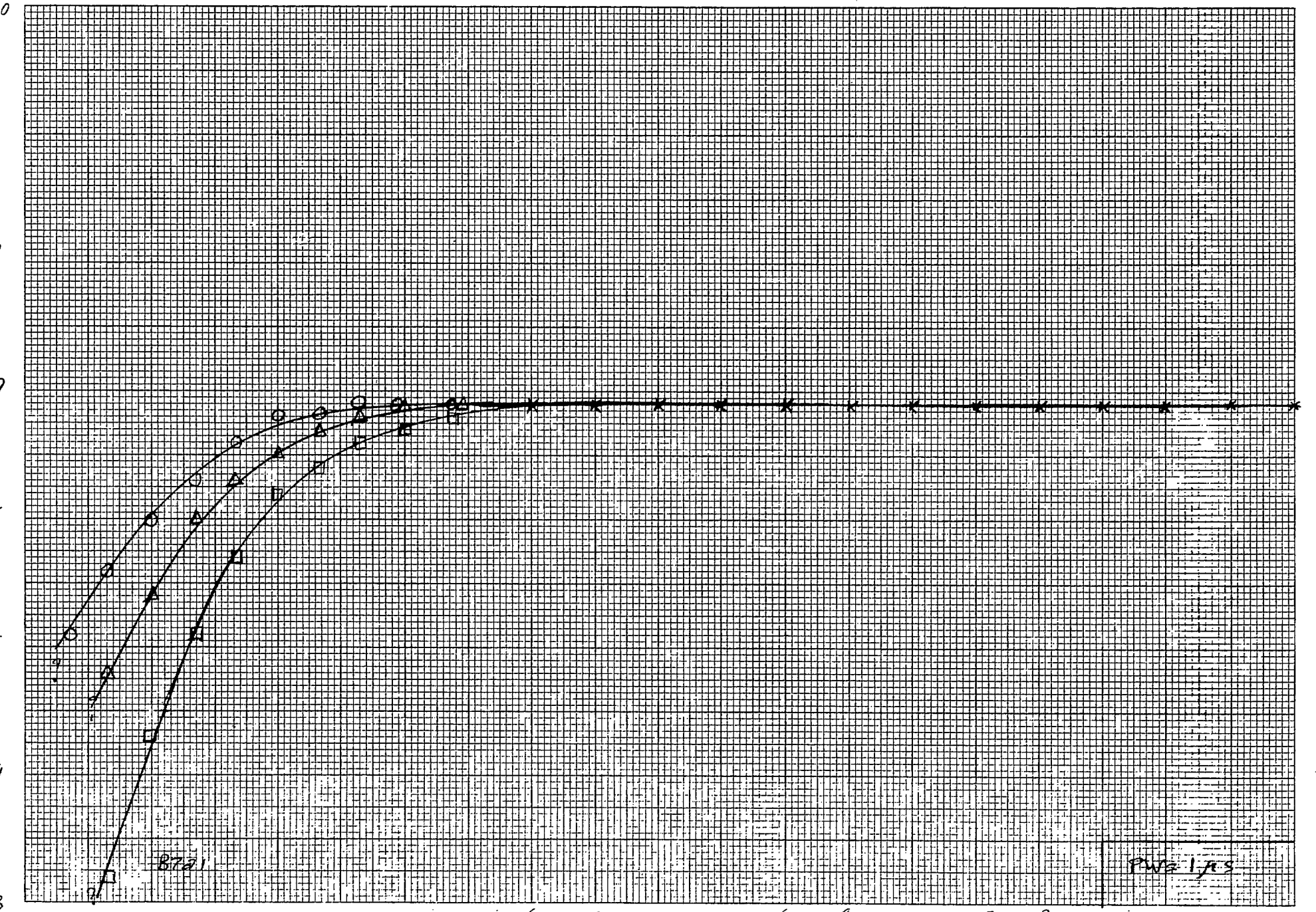
37

36

35

34

33



8721

PWR 1/CS

-30 -27 -24 -21 -18 -15 -12 -9 -6 -3 0 3 6 9 12 15 18 21 24 27 30
Co/Ip (dB)

PRF = $\circ = .3 \text{ KHz}$
 $\Delta = 1 \text{ KHz}$
 $\square = 3 \text{ KHz}$ } * = 2U.

S/N (dB)

40

39

38

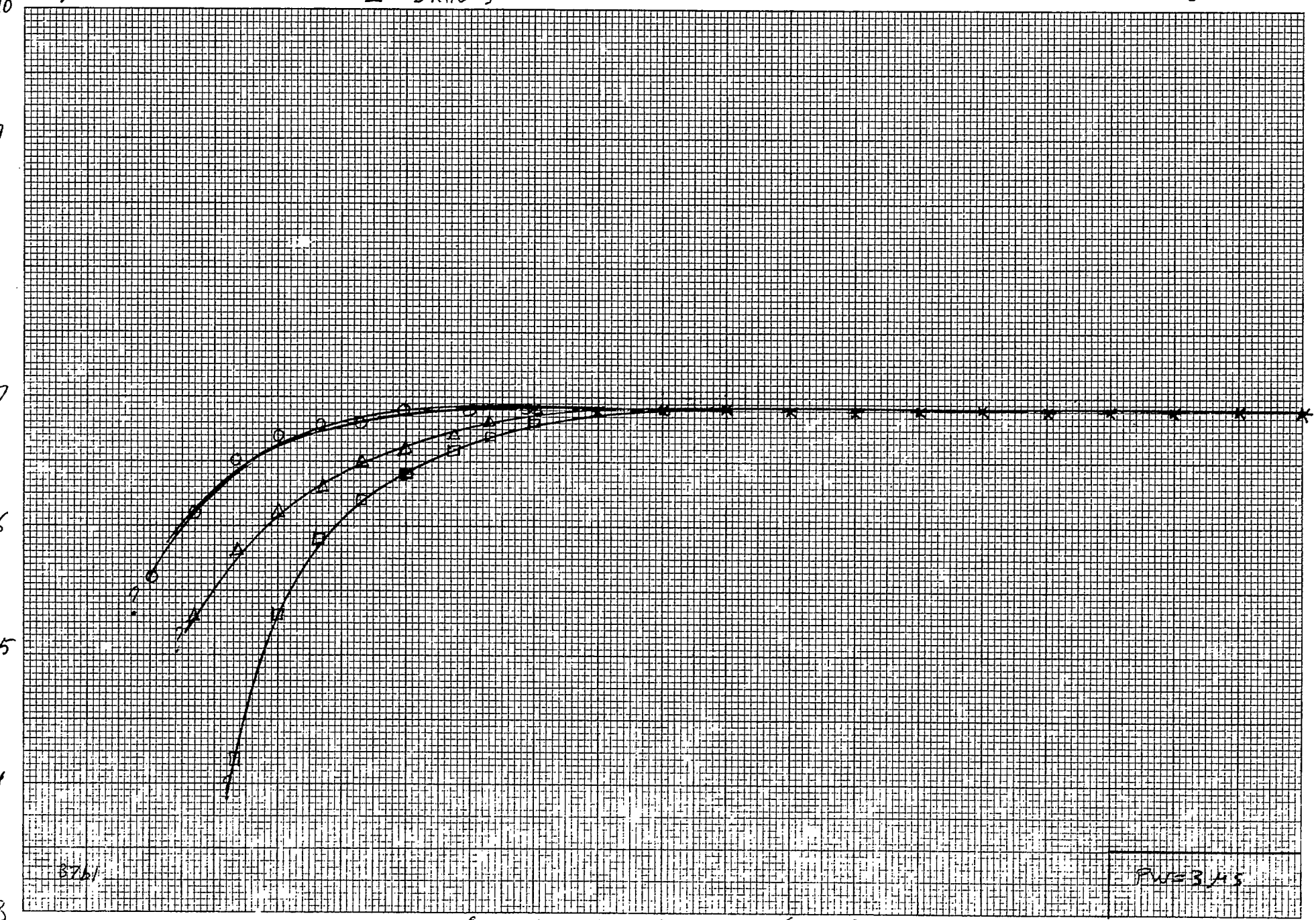
37

36

35

34

33



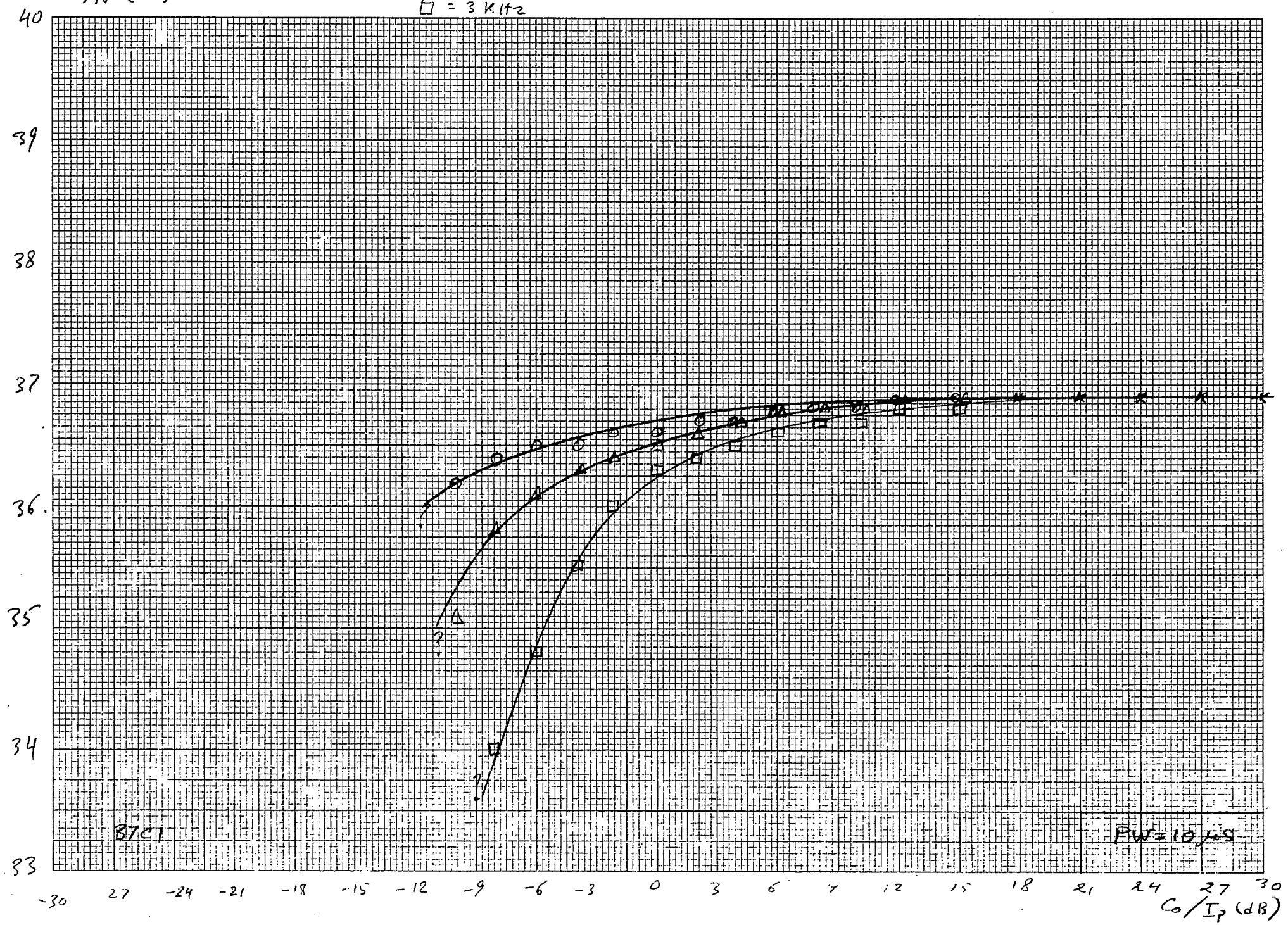
8761

FWE 3/15

-30 -27 -24 -21 -18 -15 -12 -9 -6 -3 0 3 6 9 12 15 18 21 24 27 30
Co/Ip (dB)

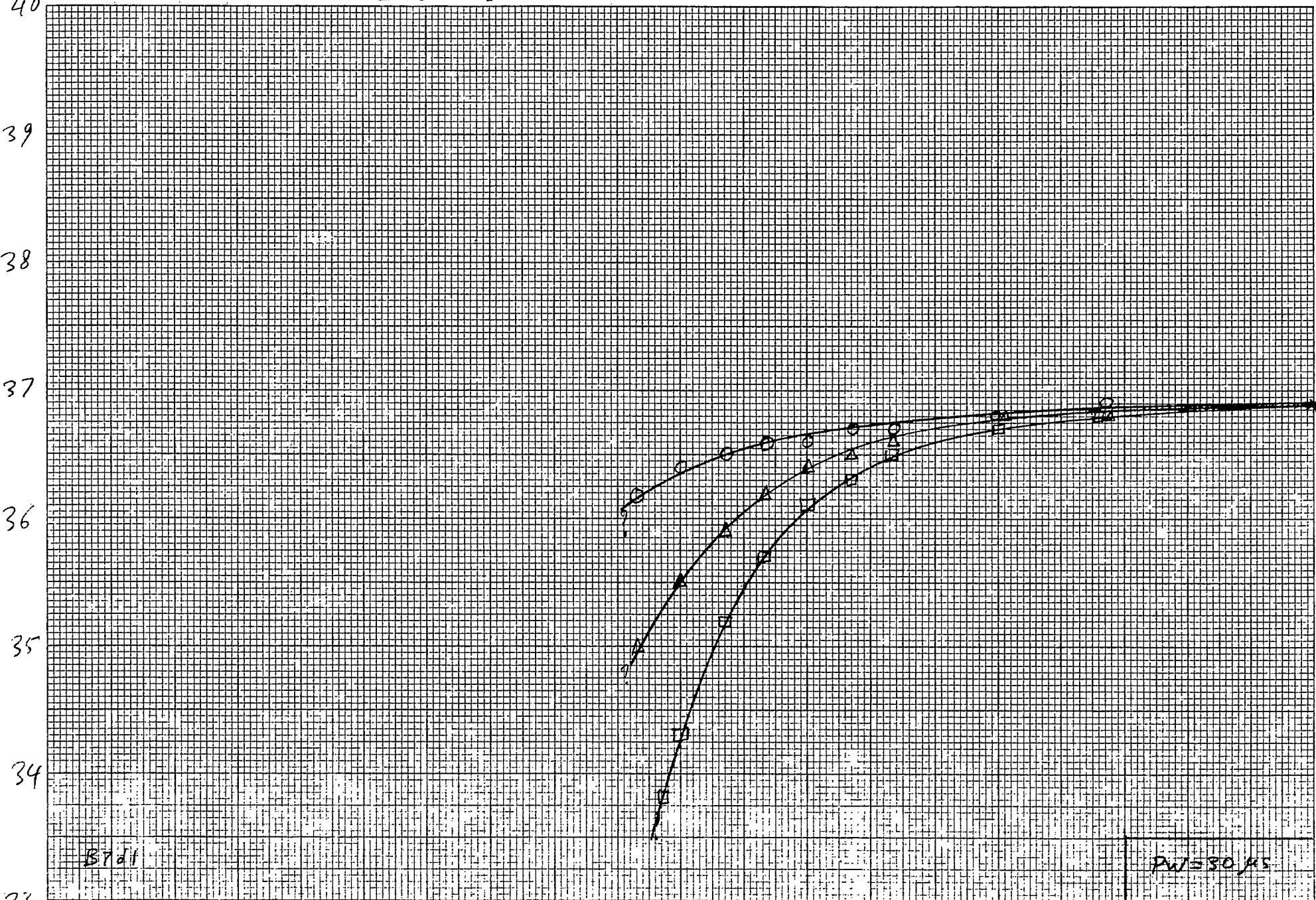
S/N (dB)

PRF = $\circ = .3 \text{ KHz}$
 $\Delta = 1 \text{ KHz}$
 $\square = 3 \text{ KHz}$



PRP = $\circ = 3\text{KHz}$
 $\Delta = 1\text{KHz}$ * = all.
 $\square = 3\text{KHz}$

S/N (dB)



Std

PW=30 μs

33 -30 -27 -24 -21 -18 -15 -12 -9 -6 -3 0 3 6 9 12 15 18 21 24 27 30

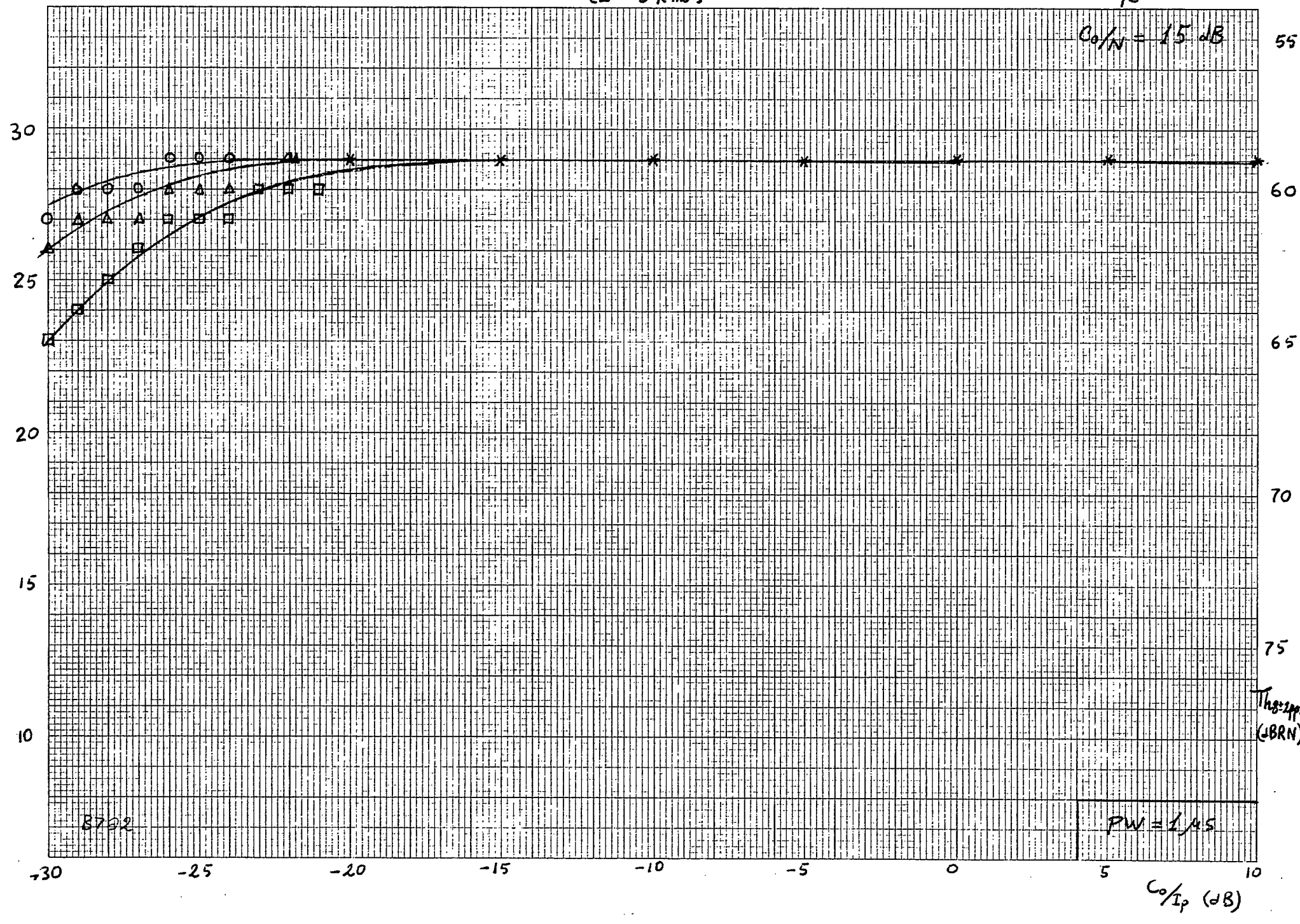
10/Ip (dB)

$S/N (dB) = 88 (dB) - T_{hg+2pps} (dB_{BRN})$

PRF = $\begin{cases} \circ = .3 \text{ KHz} \\ \Delta = 1 \text{ KHz} \\ \square = 3 \text{ KHz} \end{cases} * = \text{all}$

Radar ; PW = 1 μ s ; $f_c = 10 \text{ KHz}$

$C_0/N = 15 \text{ dB}$



8702

PW = 1 μ s

$C_0/I_p (dB)$

$$S/N \text{ (dB)} = 88 \text{ (dB)} - T_{h8\pm 2pp} \text{ (dB RN)}$$

$$PRF = \begin{cases} \circ = .3 \text{ KHz} \\ \Delta = 1 \text{ KHz} \\ \square = 3 \text{ KHz} \end{cases} * = \text{all}$$

Radar ; $PW = 3 \mu s$; $f_c = 10 \text{ KHz}$

$C_0/N = 15 \text{ dB}$

55

60

65

70

75

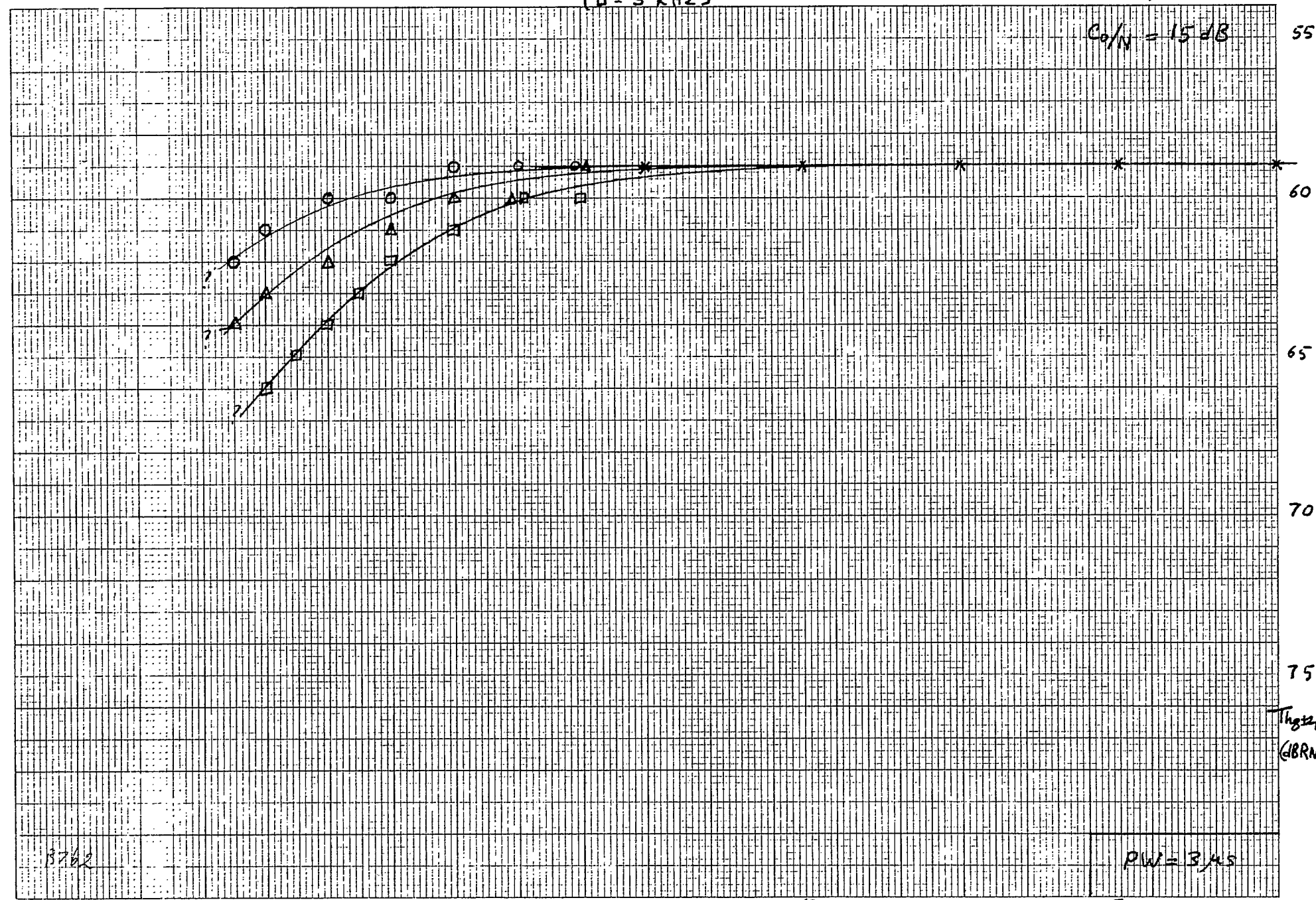
$T_{h8\pm 2pp}$
(dB RN)

$PW = 3 \mu s$

C_0/I_p (dB)

-30 -25 -20 -15 -10 -5 0 5 10

3782

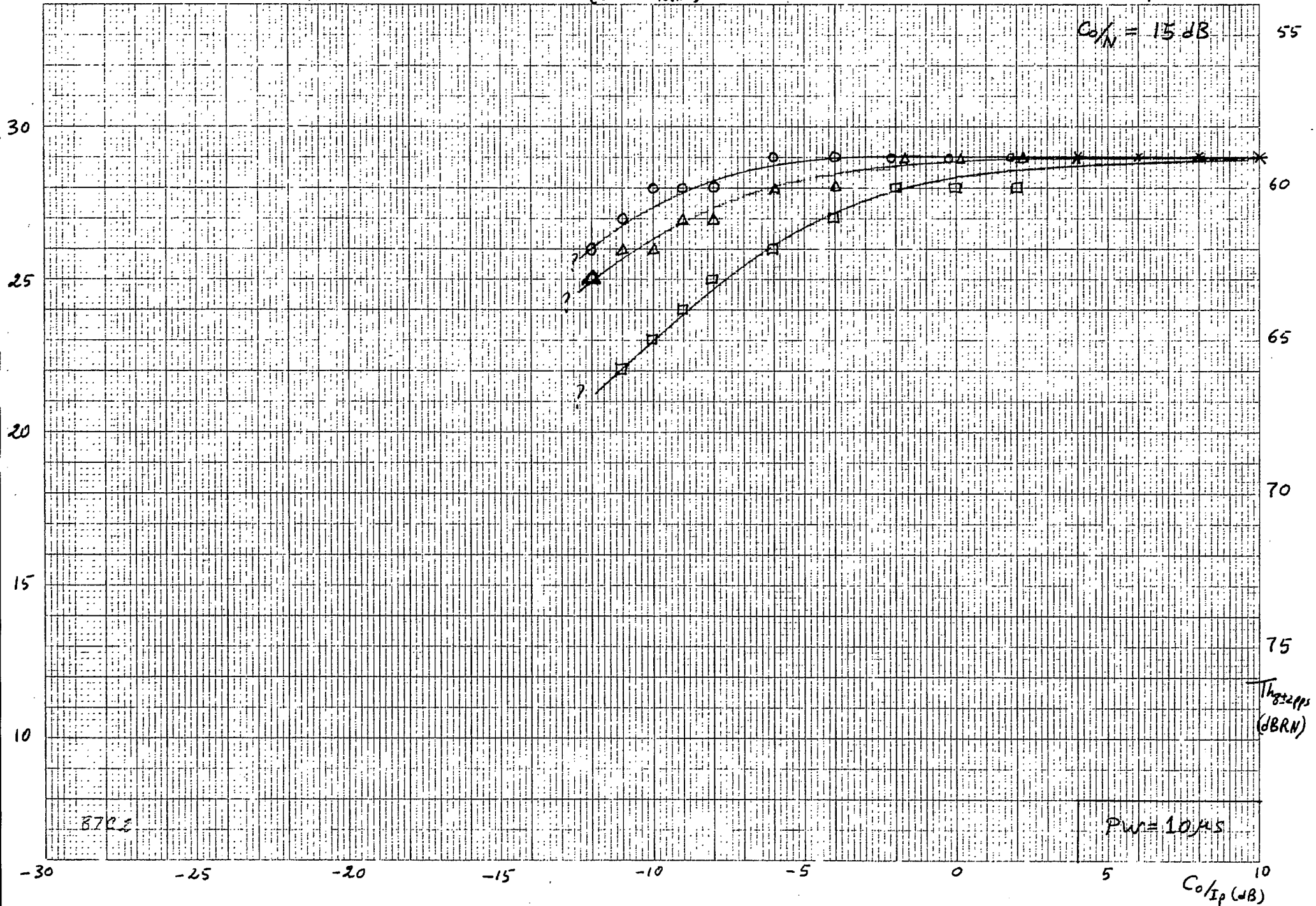


$$S/N \text{ (dB)} = 88 \text{ (dB)} - T_{h_{8 \pm 2 \text{pps}}} \text{ (dB RN)}$$

$$PRF = \begin{cases} \circ = .3 \text{ KHz} \\ \Delta = 1 \text{ KHz} \\ \square = 3 \text{ KHz} \end{cases} * = \text{all}$$

Radax ; PW = 10 μ s ; $f_c = 10 \text{ KHz}$

$C_0/I_N = 15 \text{ dB}$



8722

PW = 10 μ s

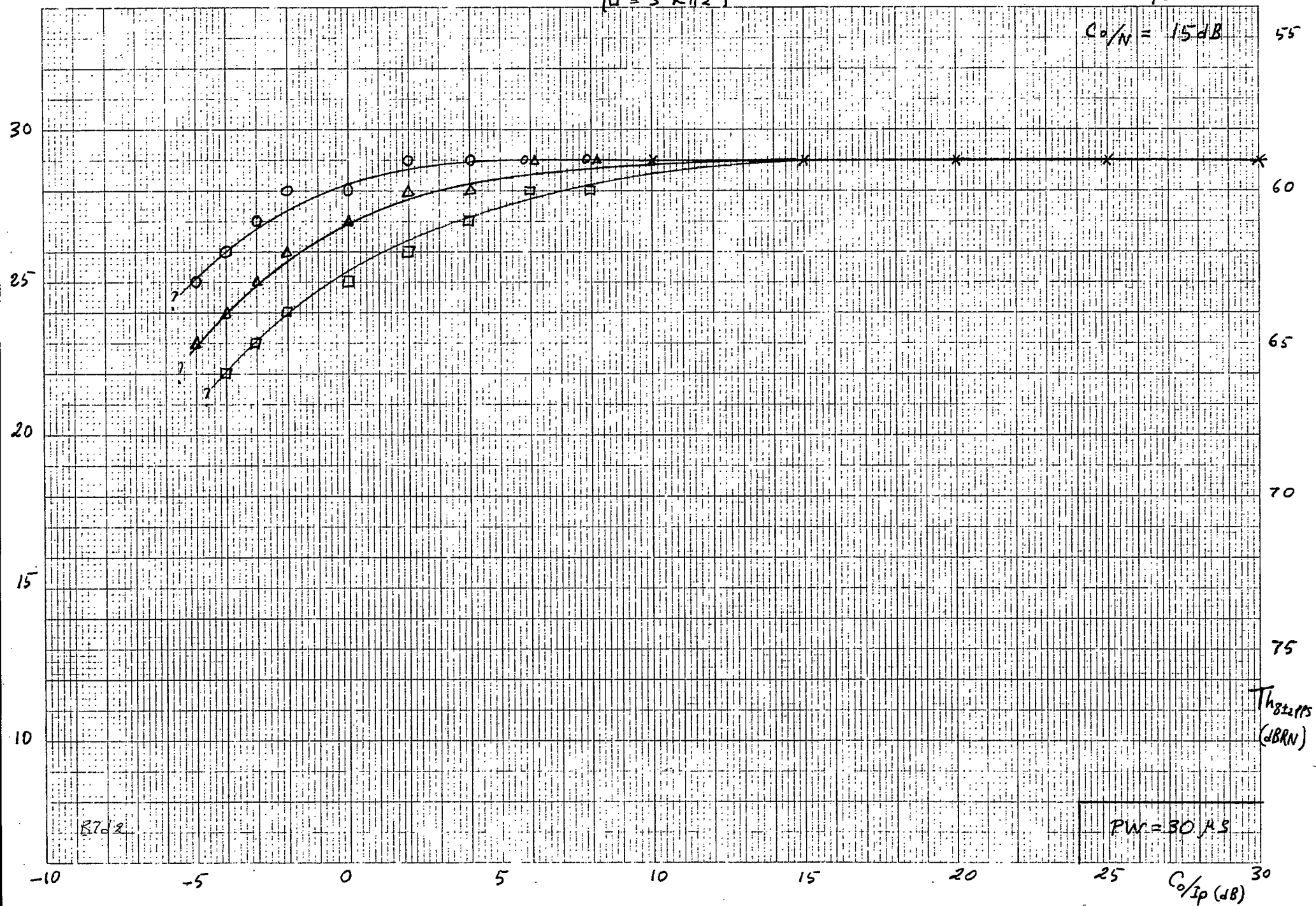
$T_{h_{8 \pm 2 \text{pps}}}$
(dB RN)

$$S/N \text{ (dB)} = 88 \text{ (dB)} - T_{h_{8\pm 2\text{pps}}} \text{ (dB RN)}$$

$$PRF = \begin{cases} f_0 = 0.3 \text{ KHz} \\ \Delta = 1 \text{ KHz} \\ \Gamma = 3 \text{ KHz} \end{cases} * = \text{all}$$

Radar ; PW = 30 μ s ; $f_c = 10 \text{ KHz}$

$C_0/N = 15 \text{ dB}$



C1

4 INTERFERING CARRIERS

SCPC / FM

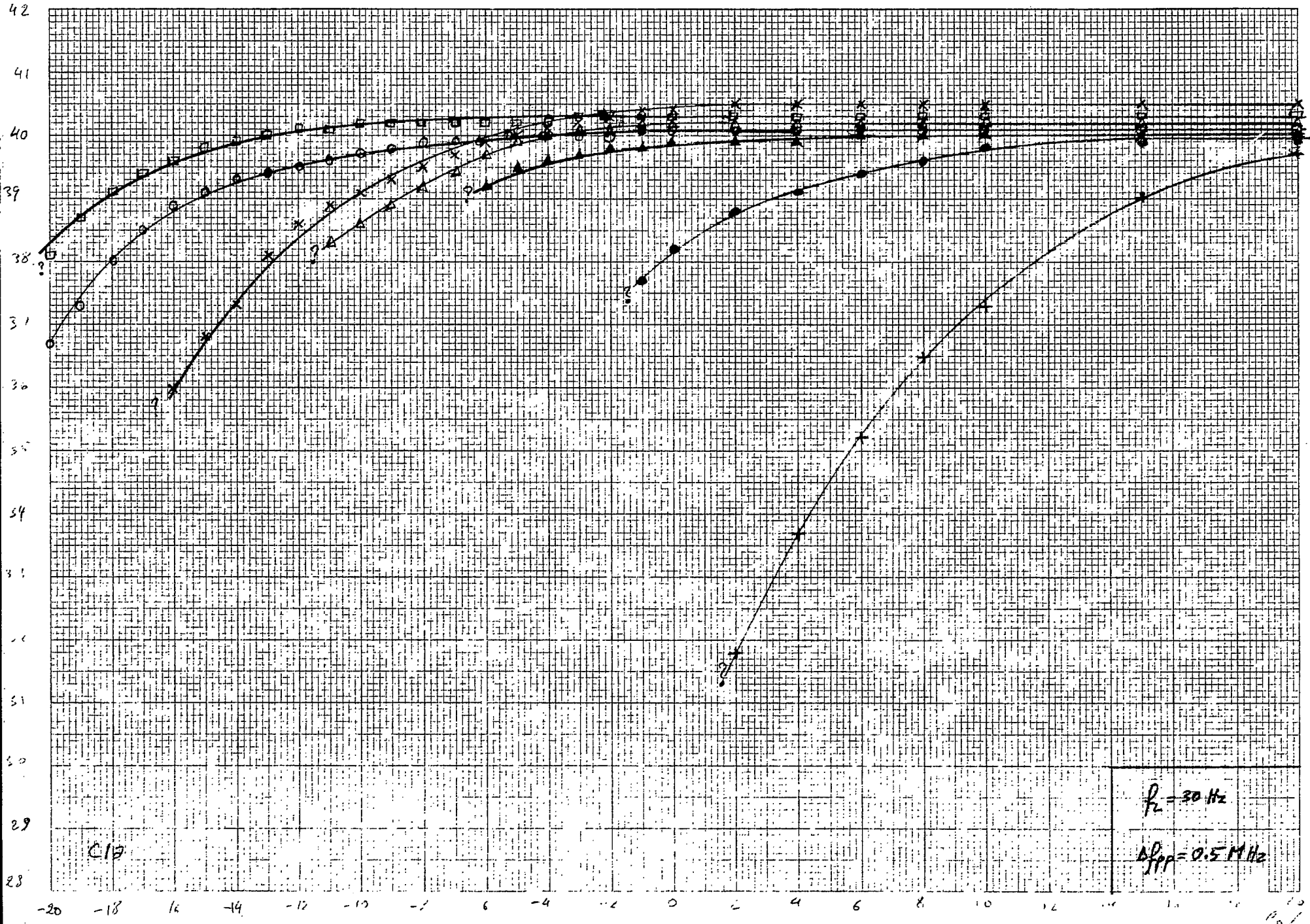
S/N_{LOADED} VS C₀/I₀

C/N = ∞

SYMBOL	f _c
X	3.6 MHz
□	3.0 MHz
Δ	2.0 MHz
○	1.0 MHz
▲	500 kHz
●	100 kHz
+	10 kHz

? = ERRATIC READING

S/N (dB)



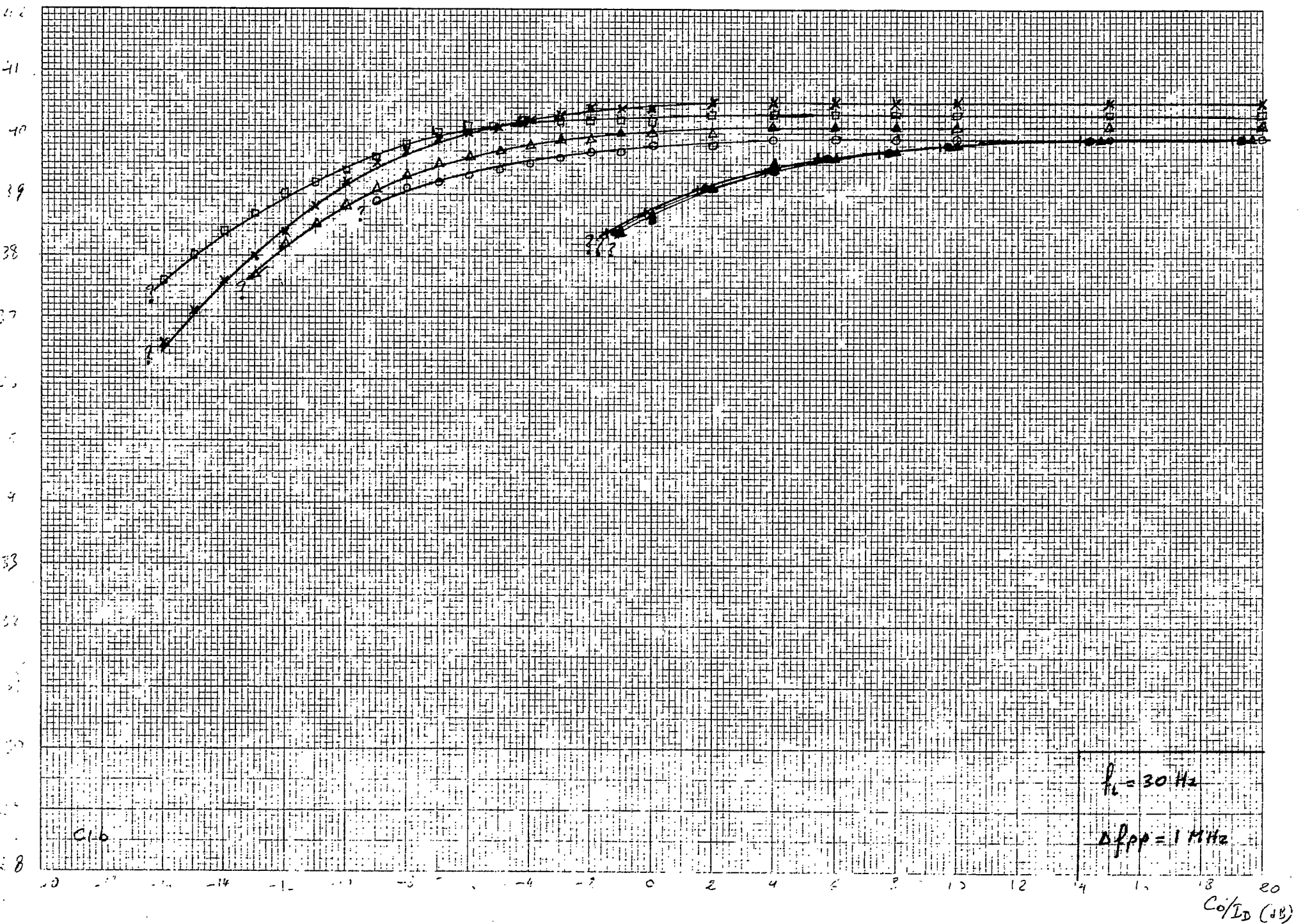
C17

$f_L = 30 \text{ Hz}$

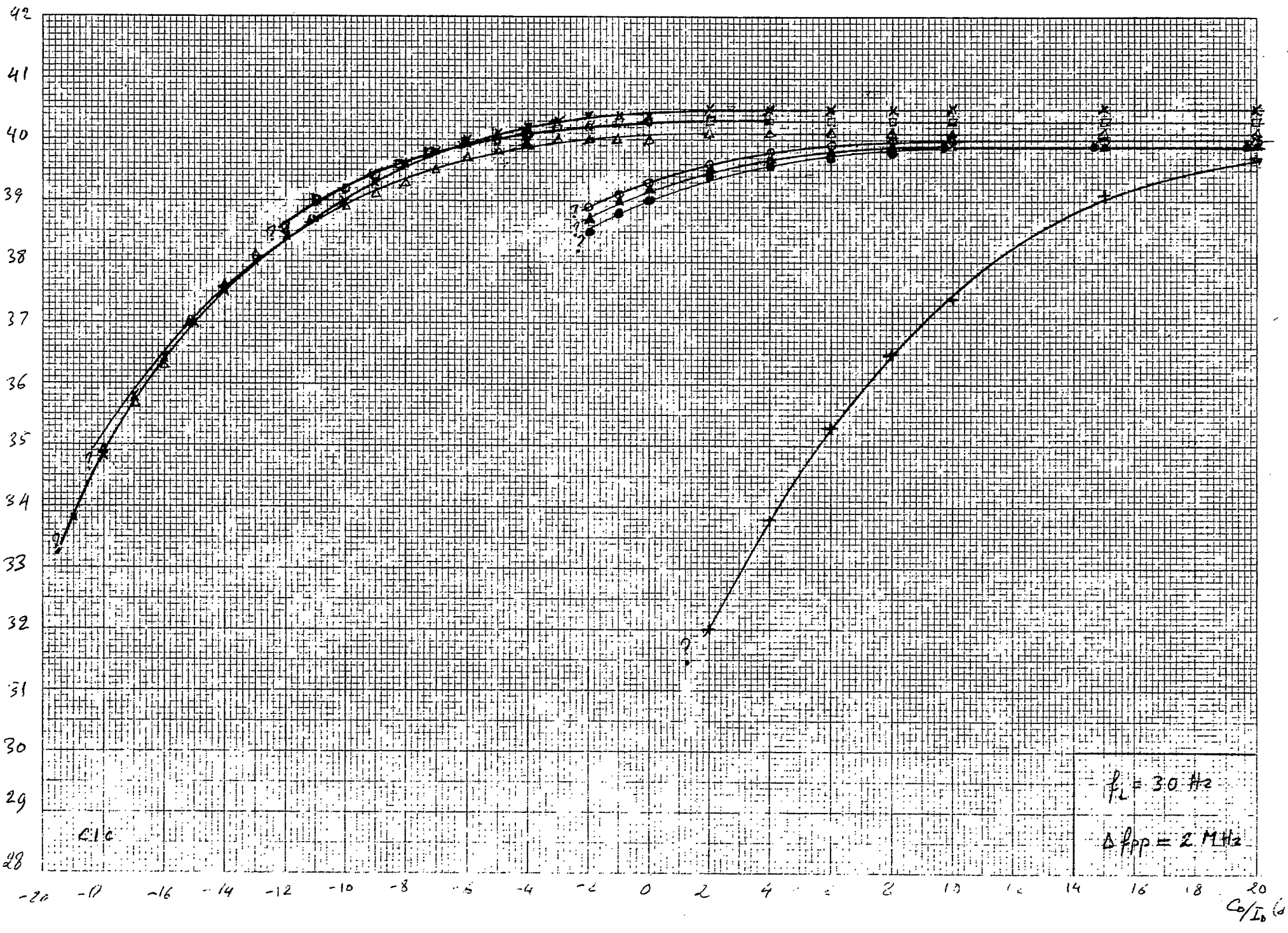
$BPF = 0.5 \text{ MHz}$

P_n/P_0 (dB)

S/N (dB)



S/N (dB)

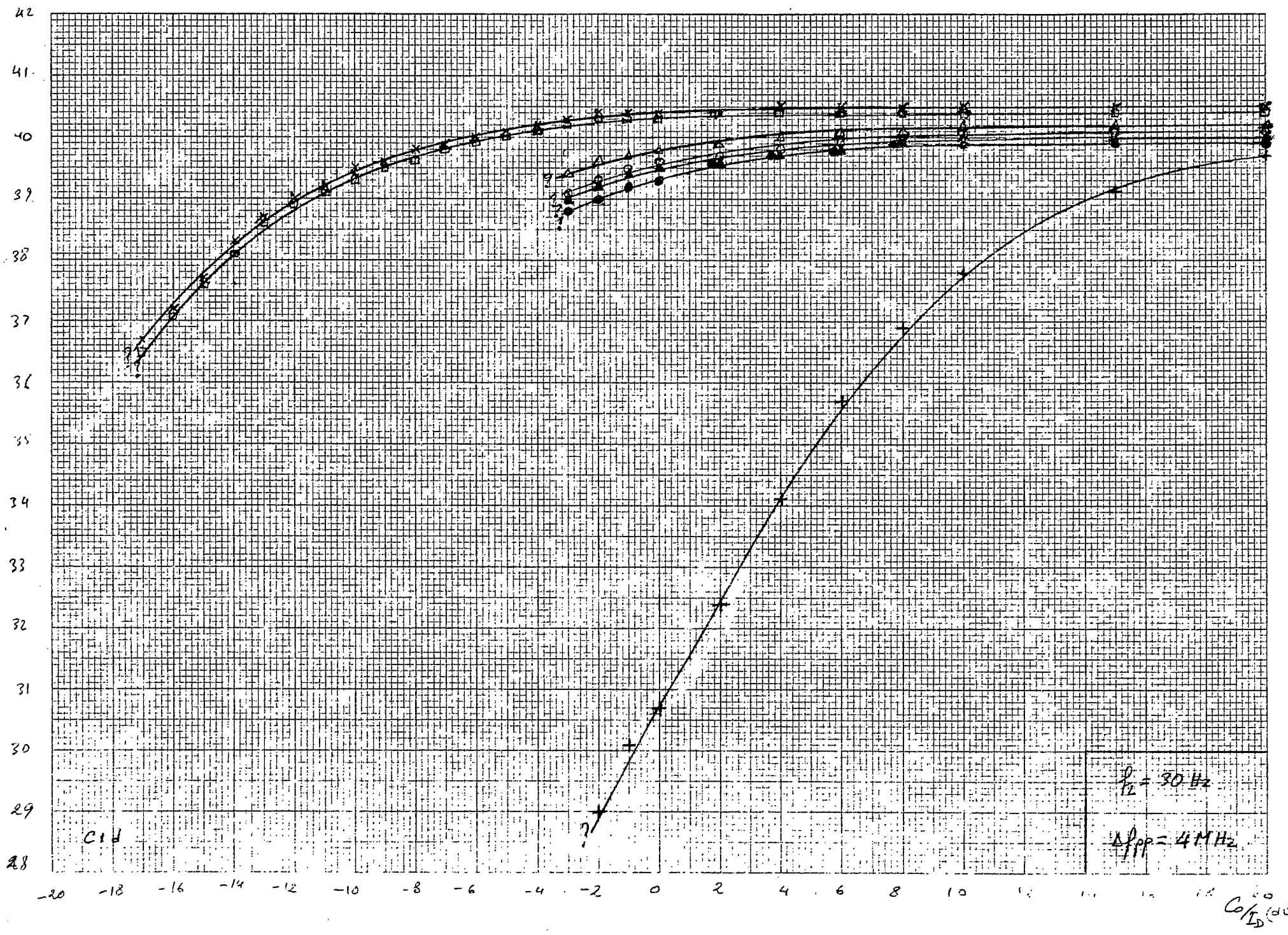


C1C

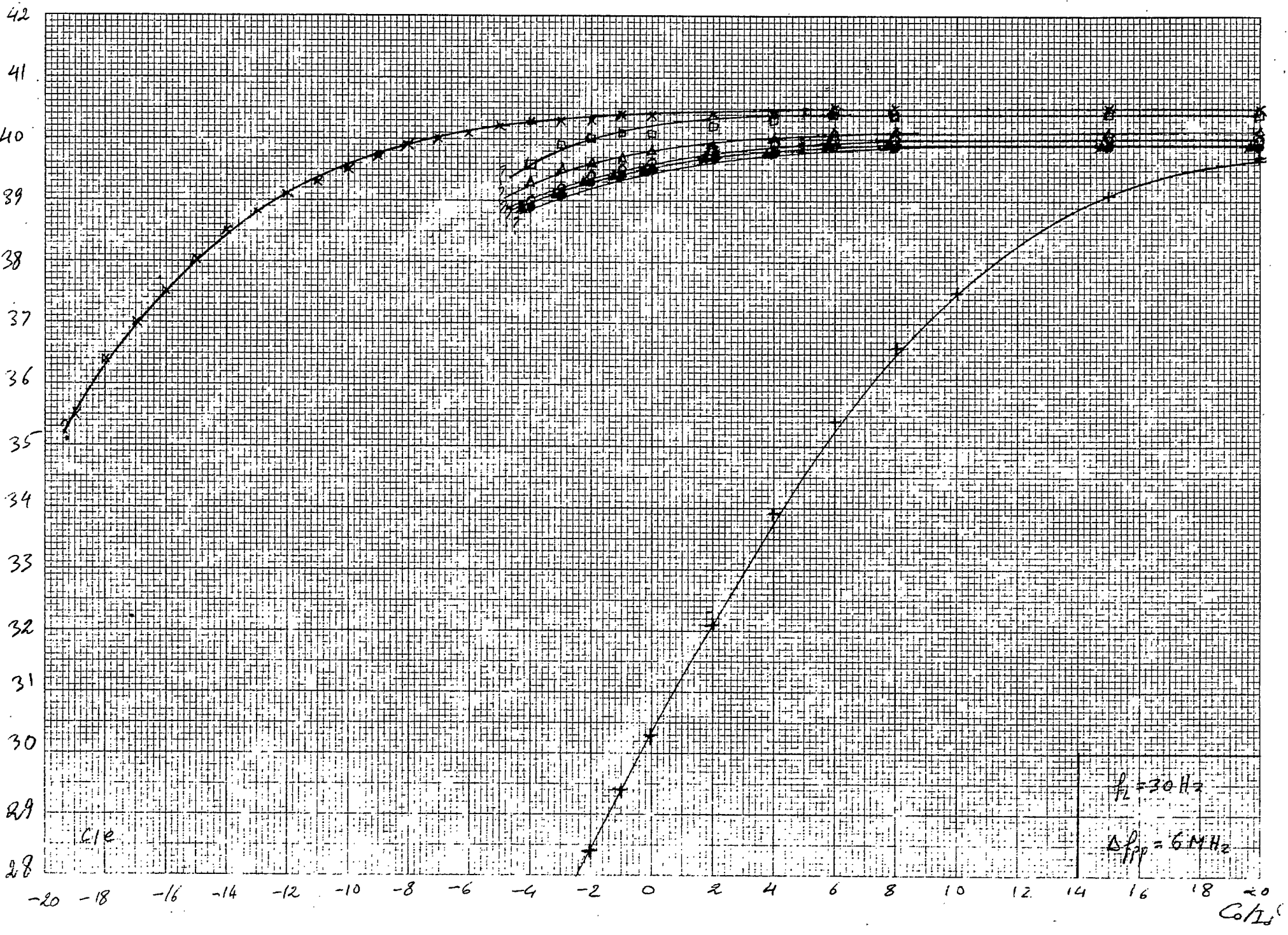
$f_L = 30 \text{ Hz}$
 $\Delta f_{pp} = 2 \text{ MHz}$

Co/Is (dB)

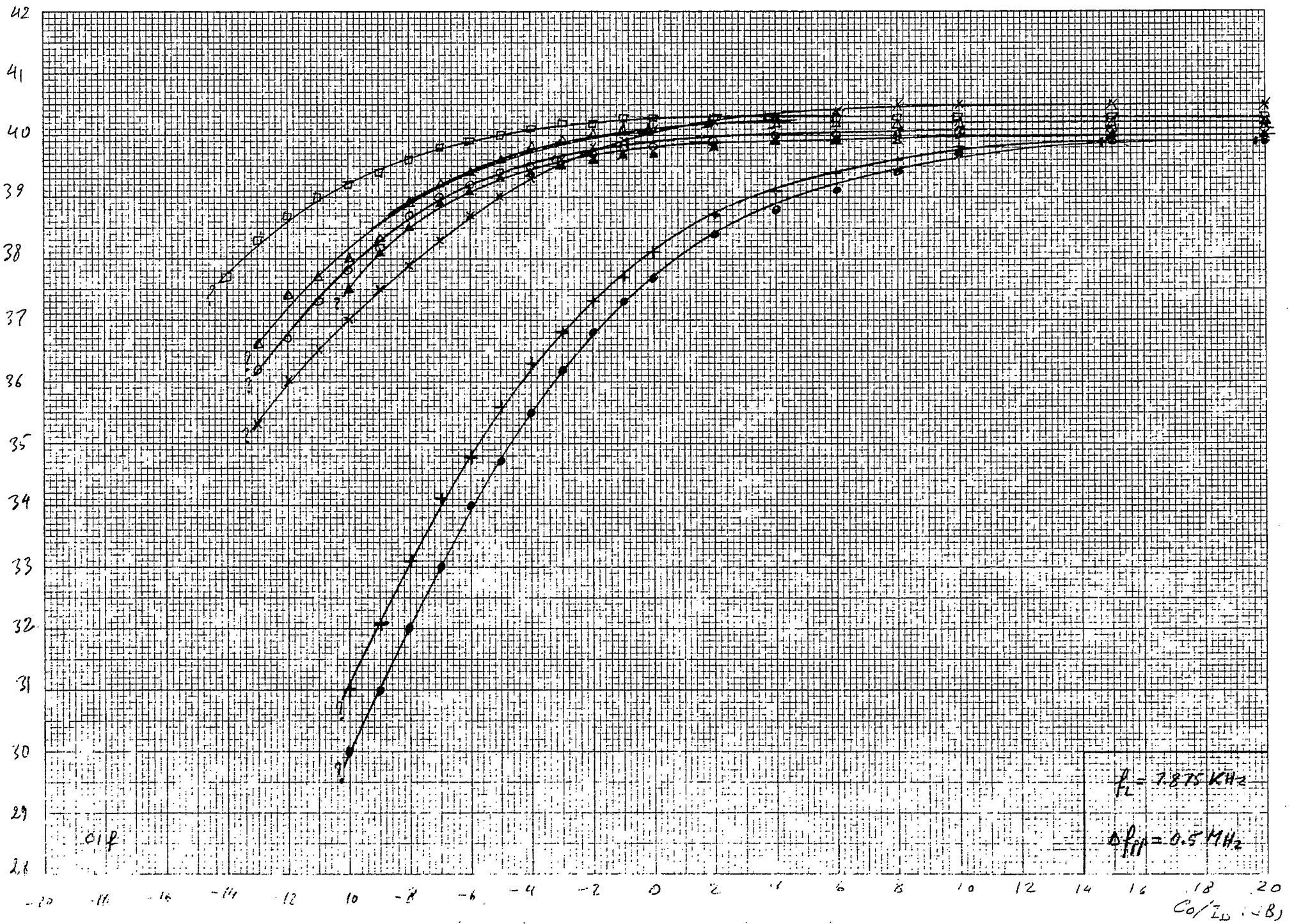
S/N (dB)



S/N (dB)



S/N (dB)



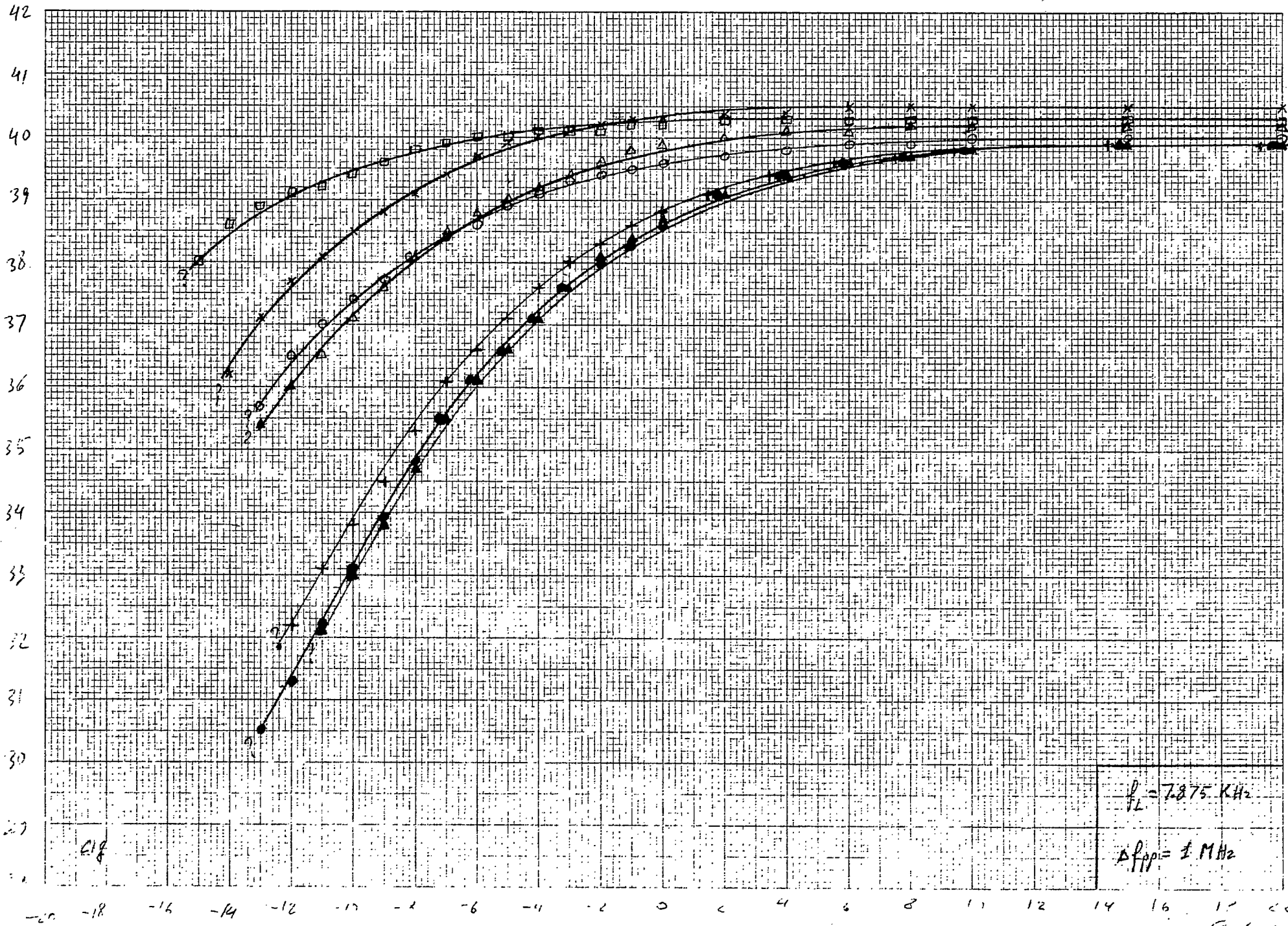
012

$f_c = 7.875 \text{ KHz}$

$\Delta f_{IF} = 0.5 \text{ MHz}$

Co/LD (dB)

S/N (dB)



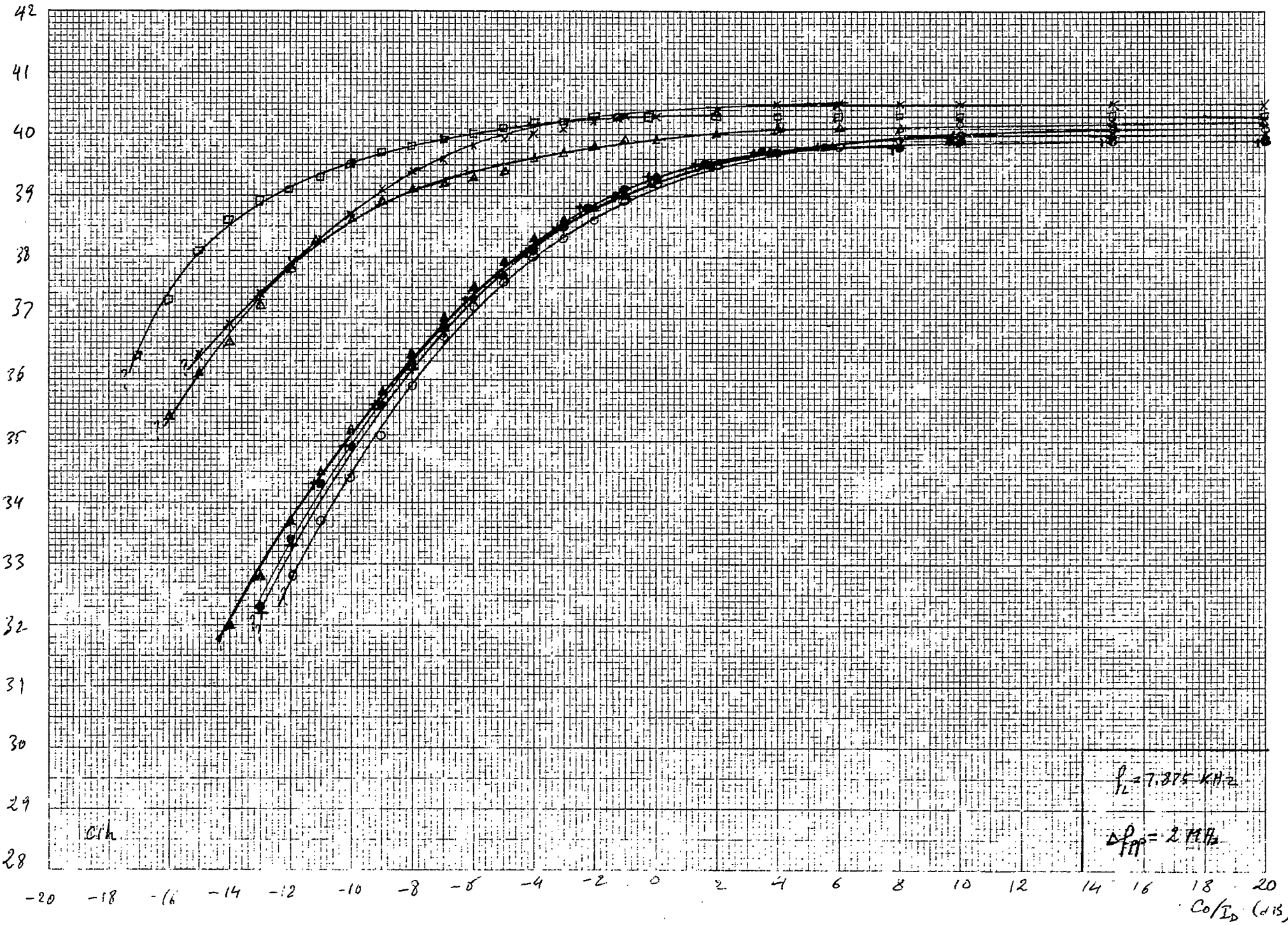
$f_L = 7.875 \text{ KHz}$

$\Delta f_{pp} = 1 \text{ MHz}$

218

C0/Id (dB)

S/N (dB)

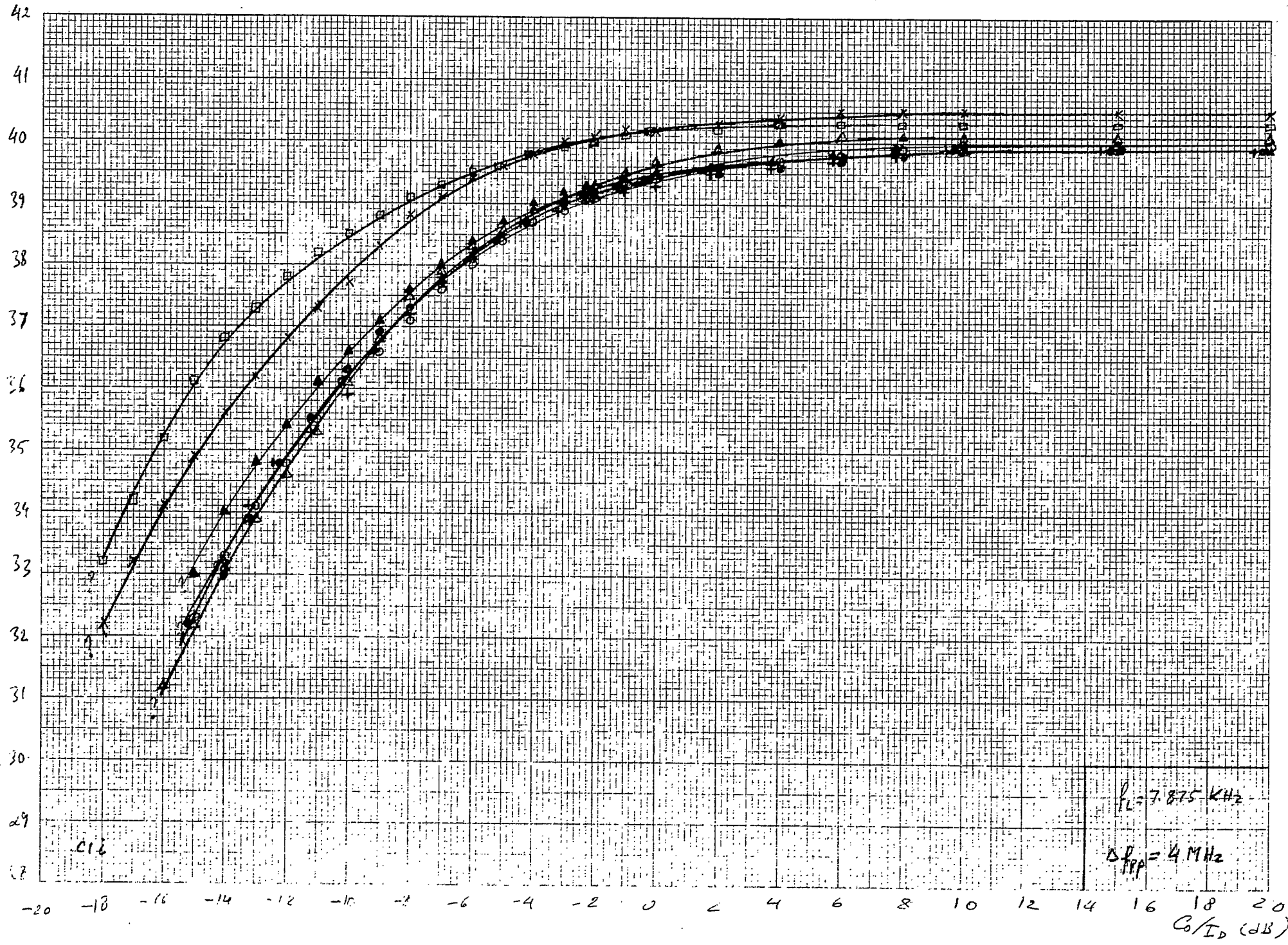


C1h

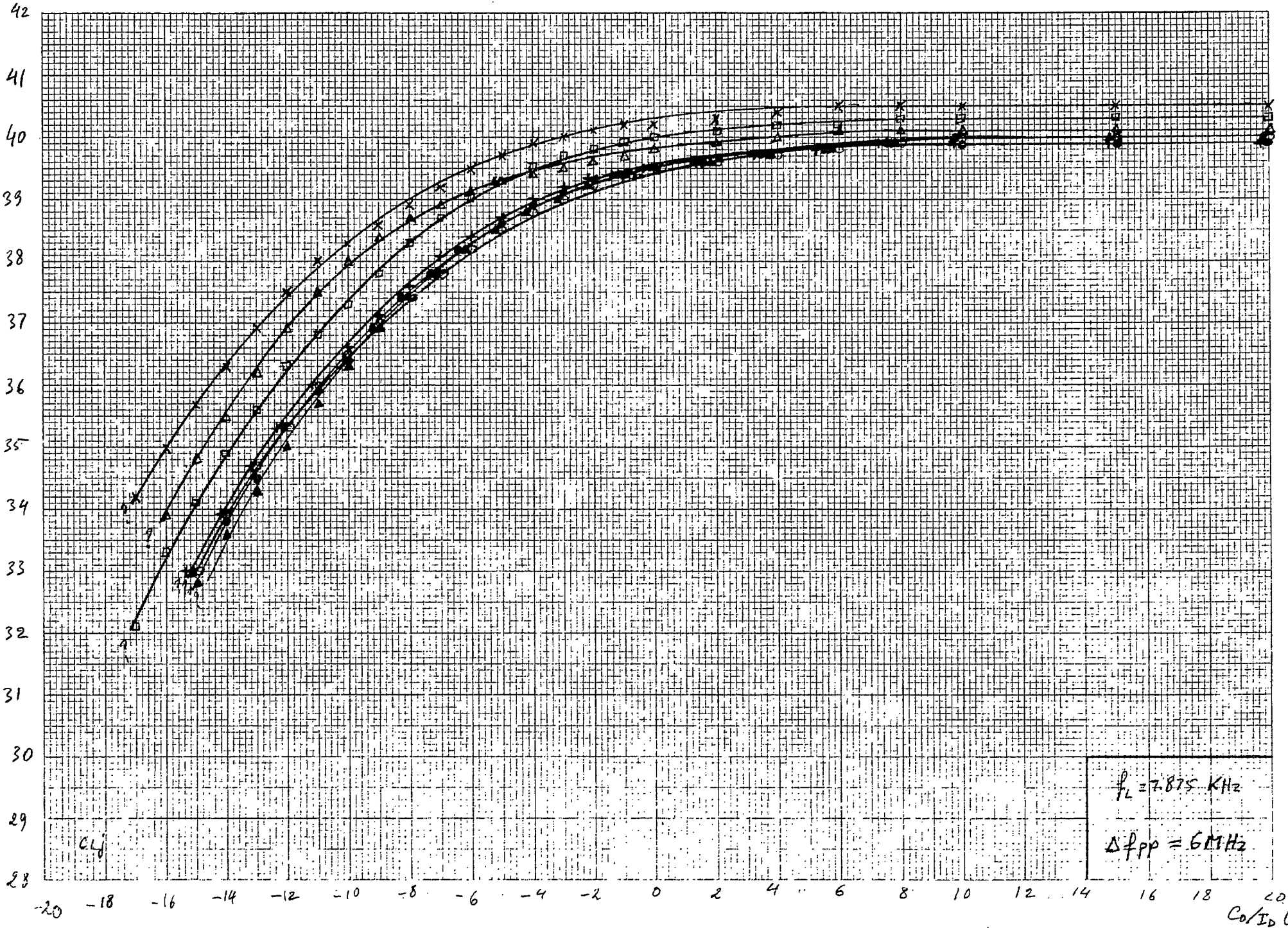
$f_L = 7.375 \text{ kHz}$

$\Delta P_{TP} = 2.7 \text{ dB}$

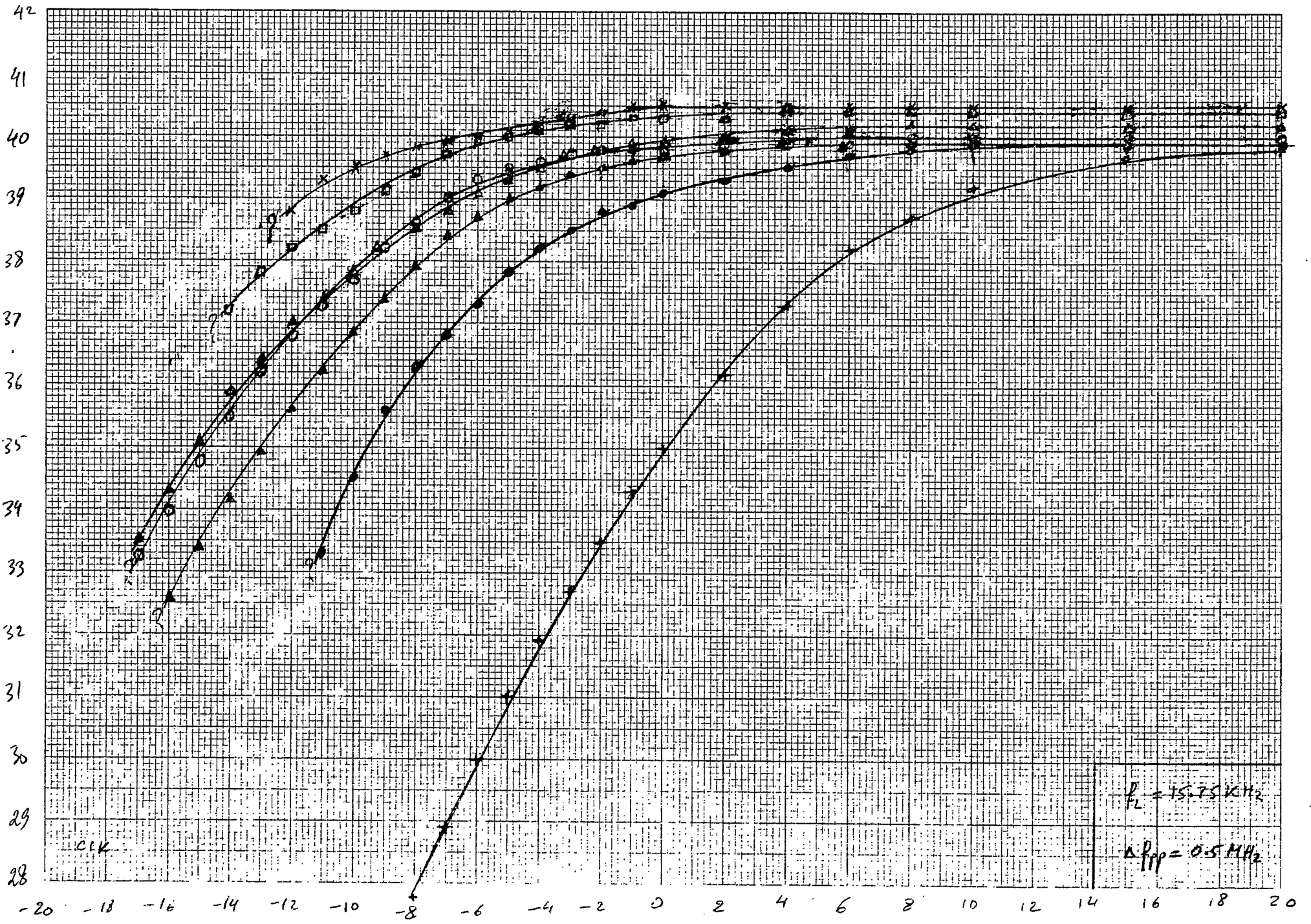
S/N (dB)



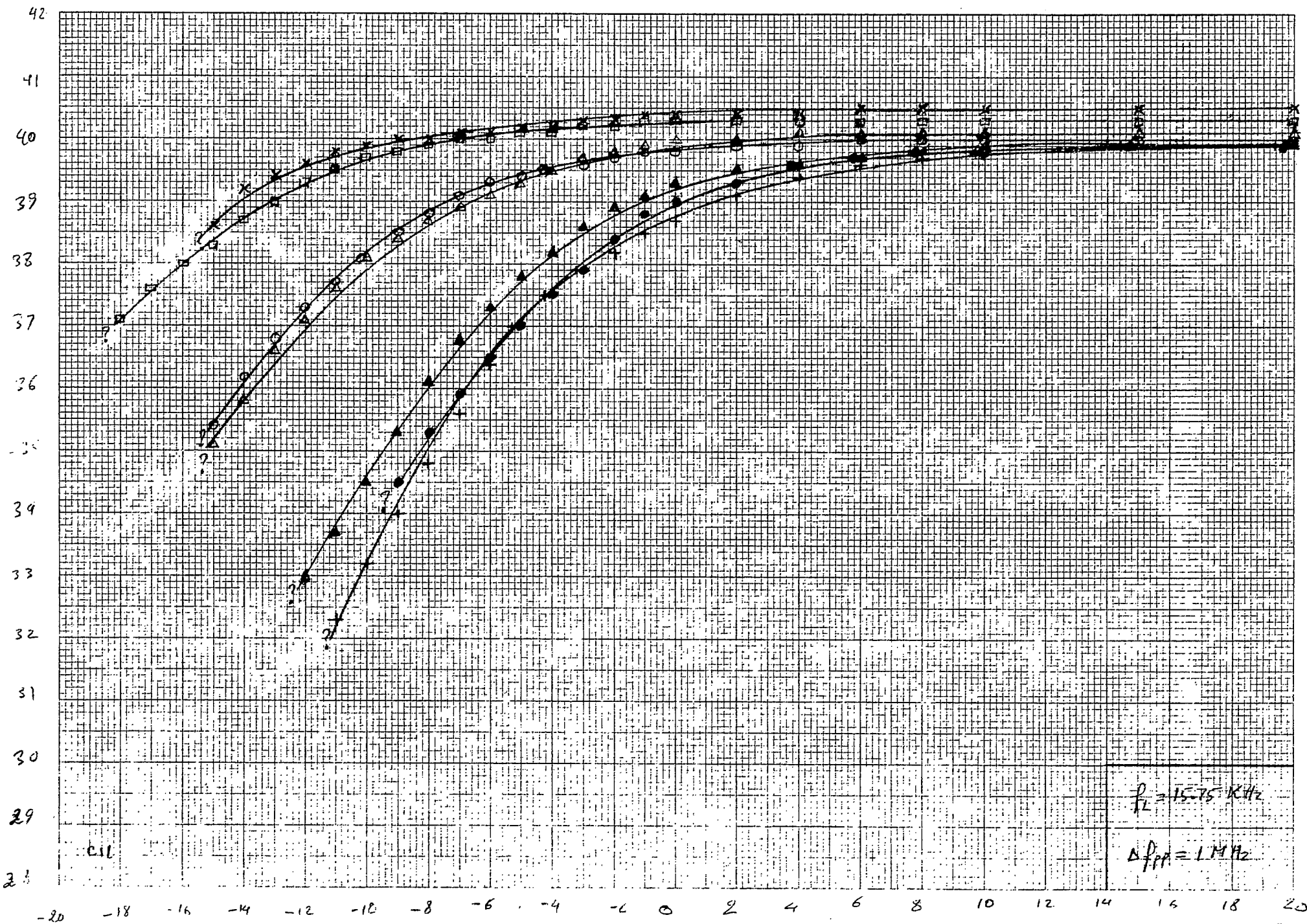
S/N (dB)



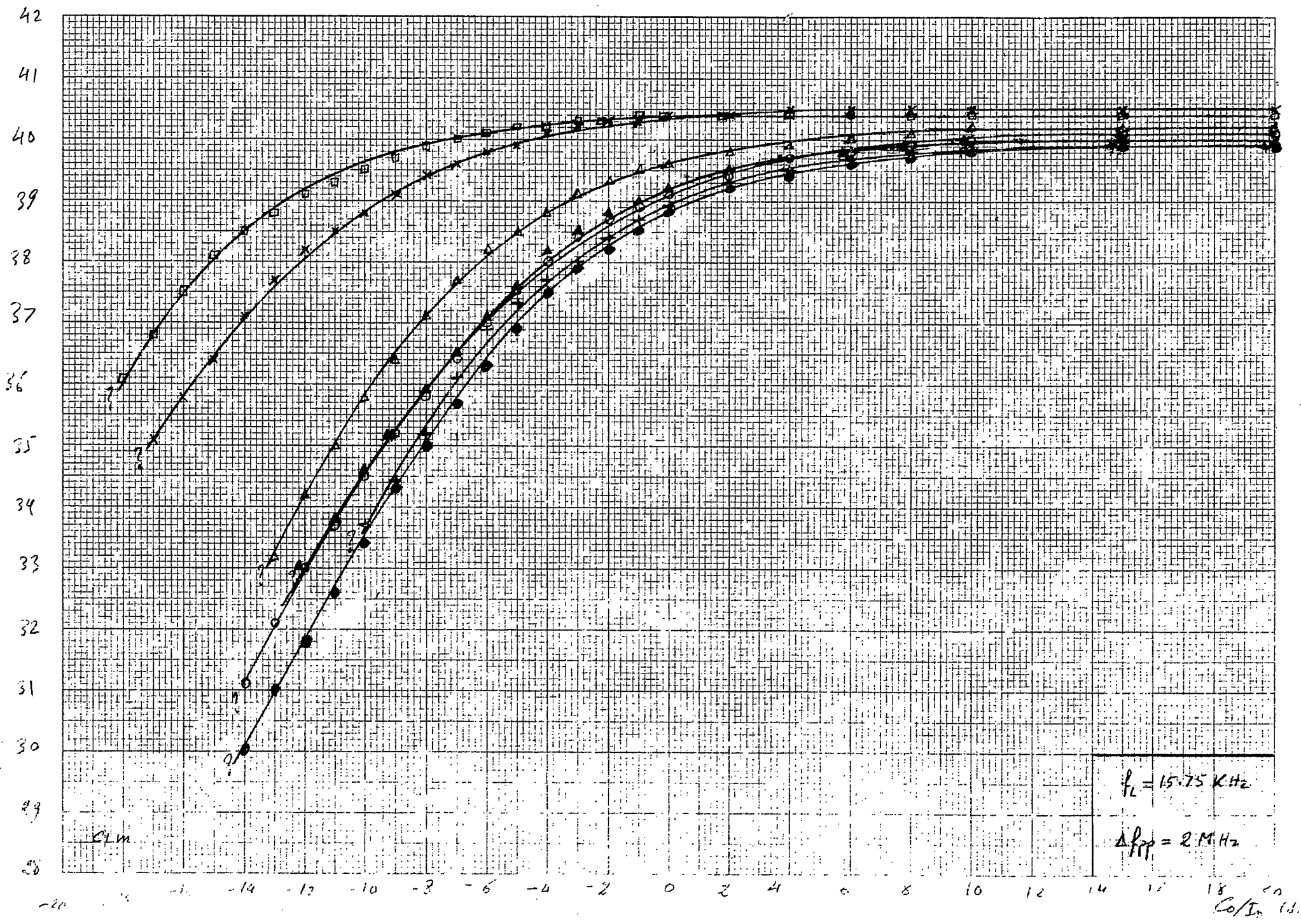
S/N (dB)



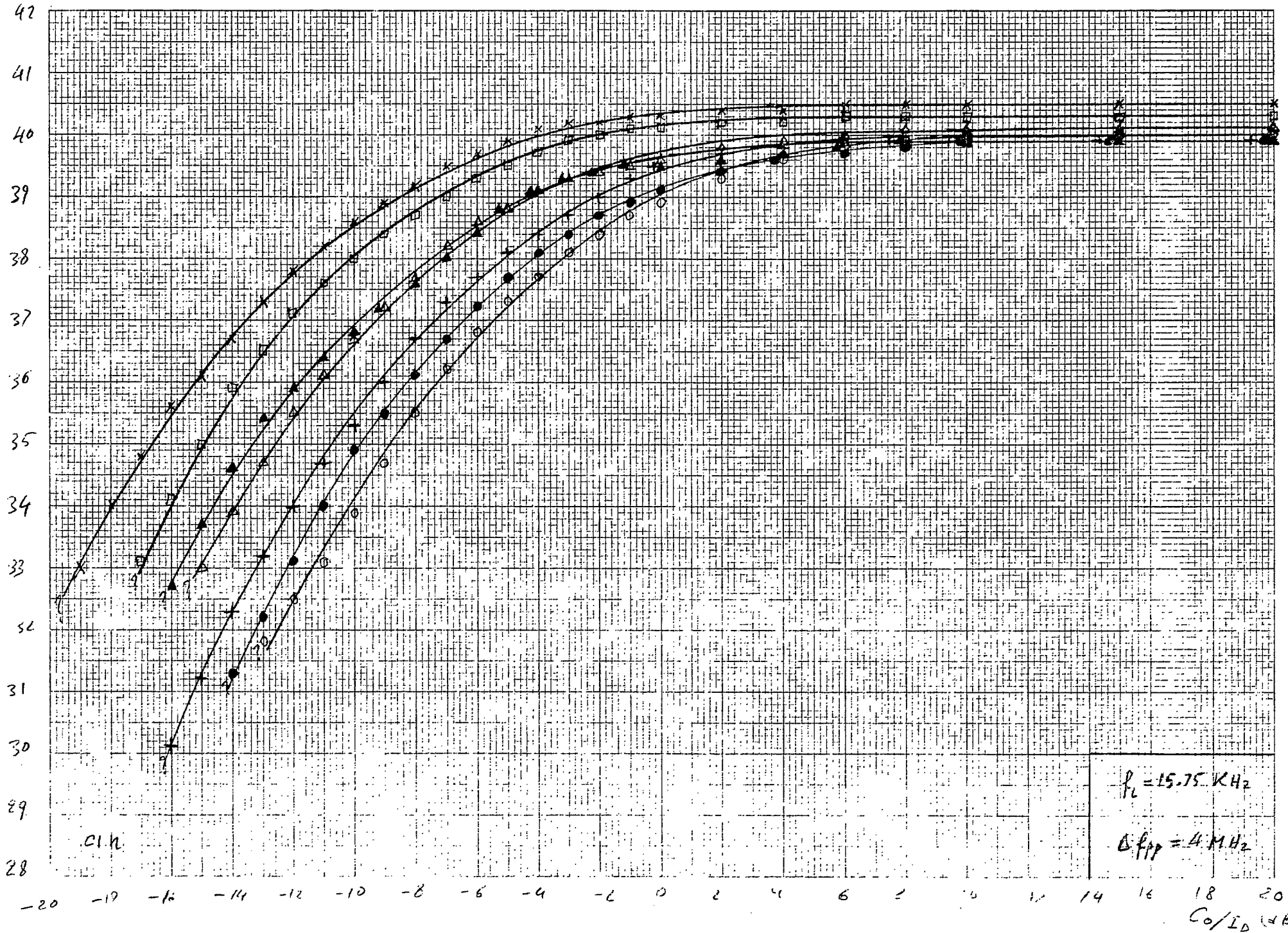
S/N (dB)



S/N (dB)



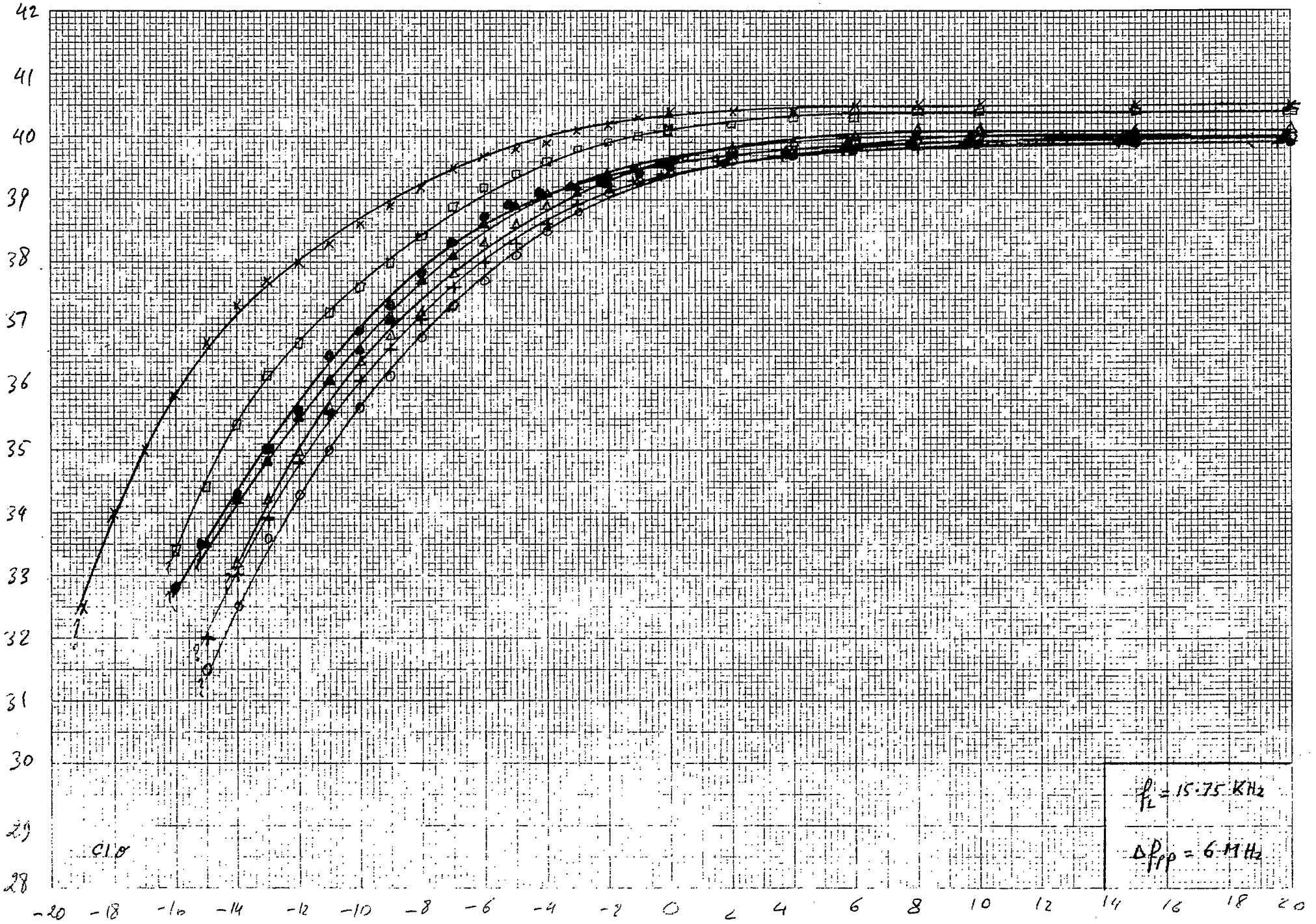
S/N (dB)



C/I

$f_L = 15.75 \text{ KHz}$
 $\Delta f_{PP} = 4 \text{ MHz}$

S/N (dB)



C2

4 INTERFERING

CARRIERS

IMPULSE NOISE THRESHOLD vs

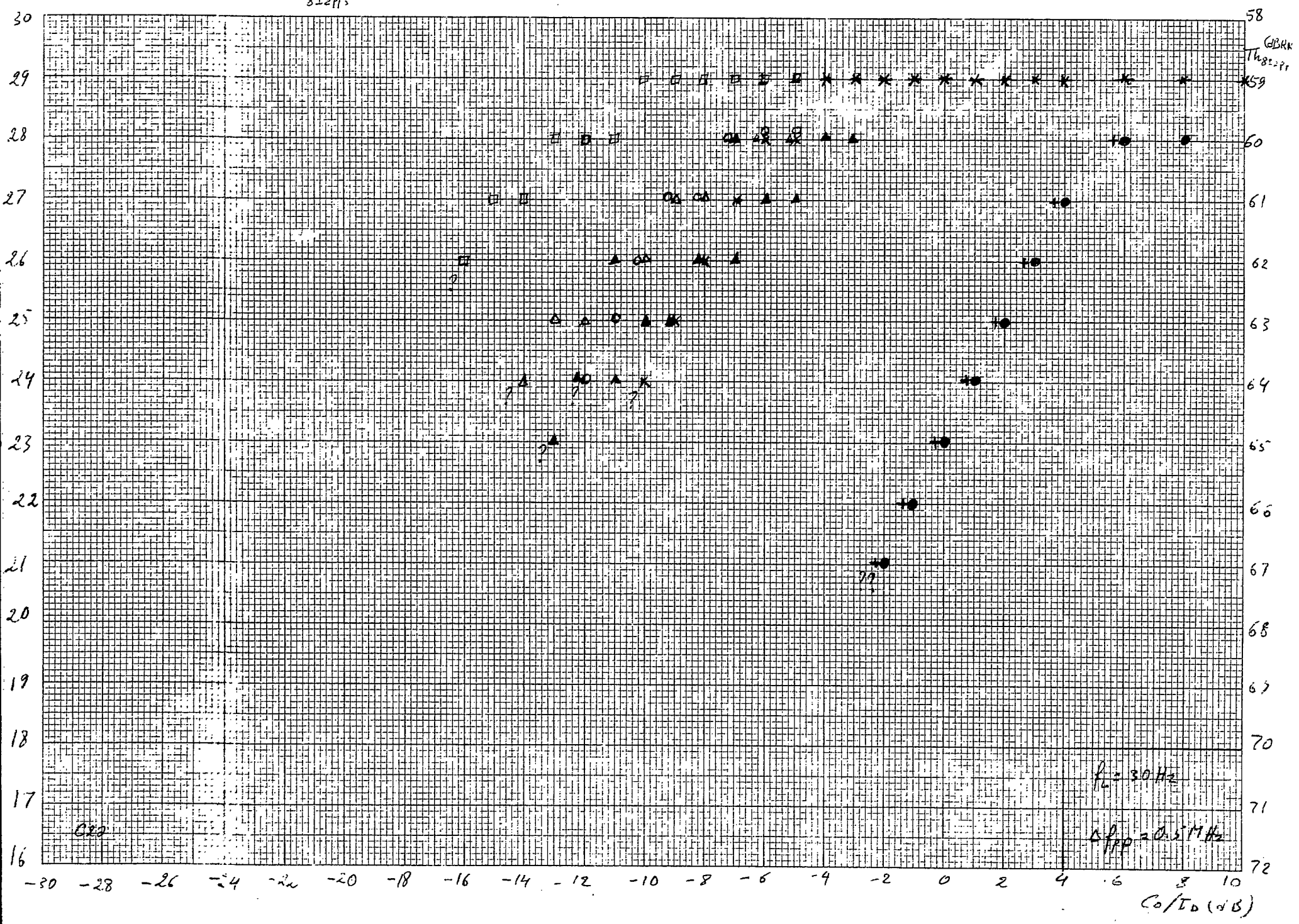
C_0/I_D

WITH $C/N = 15$ dB

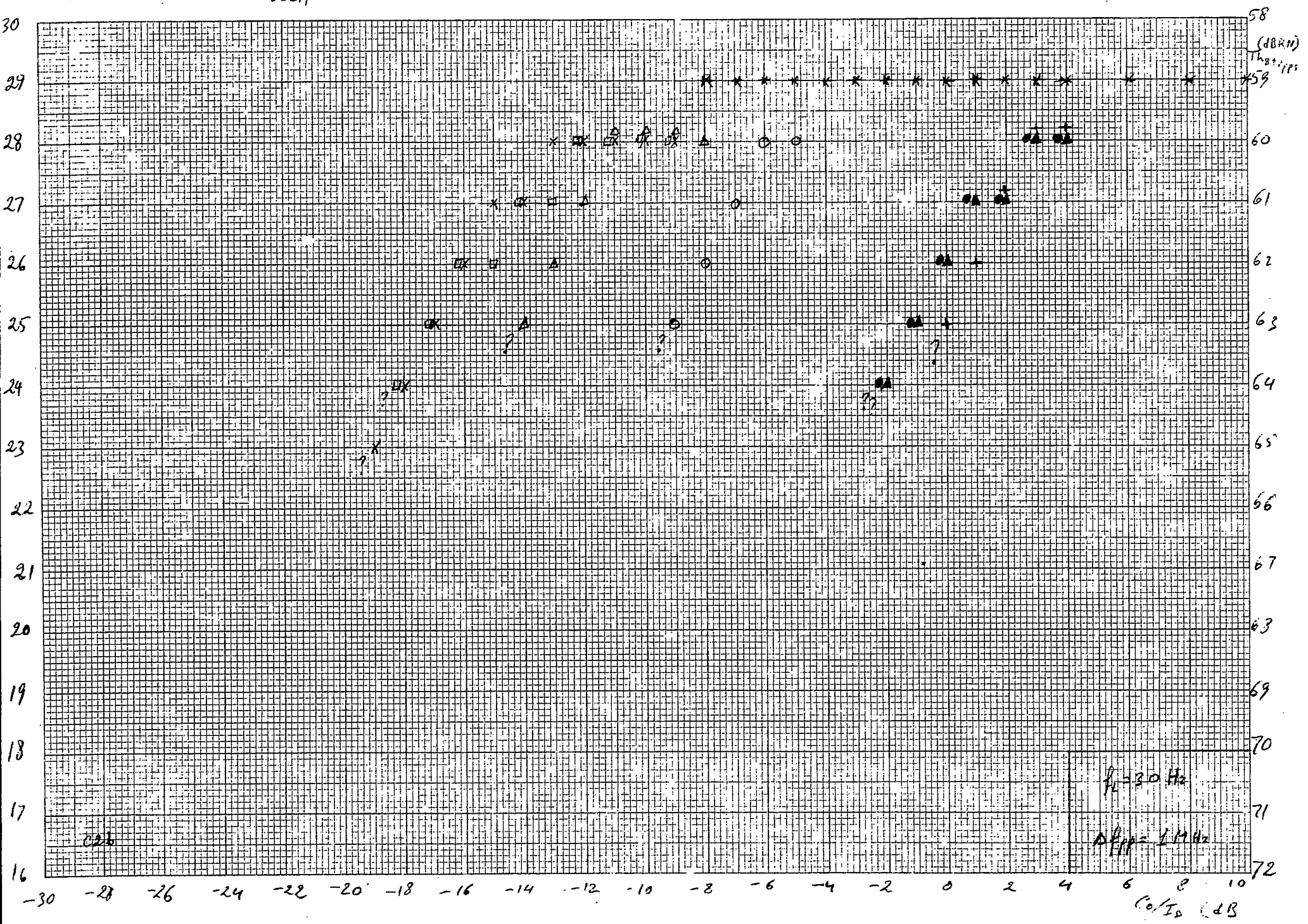
SYMBOL	f_c
X	3.6 MHz
□	3.0 MHz
Δ	2.0 MHz
○	1.0 MHz
▲	500 kHz
●	100 kHz
+	10 kHz

? = ERRATIC BEHAVIOR

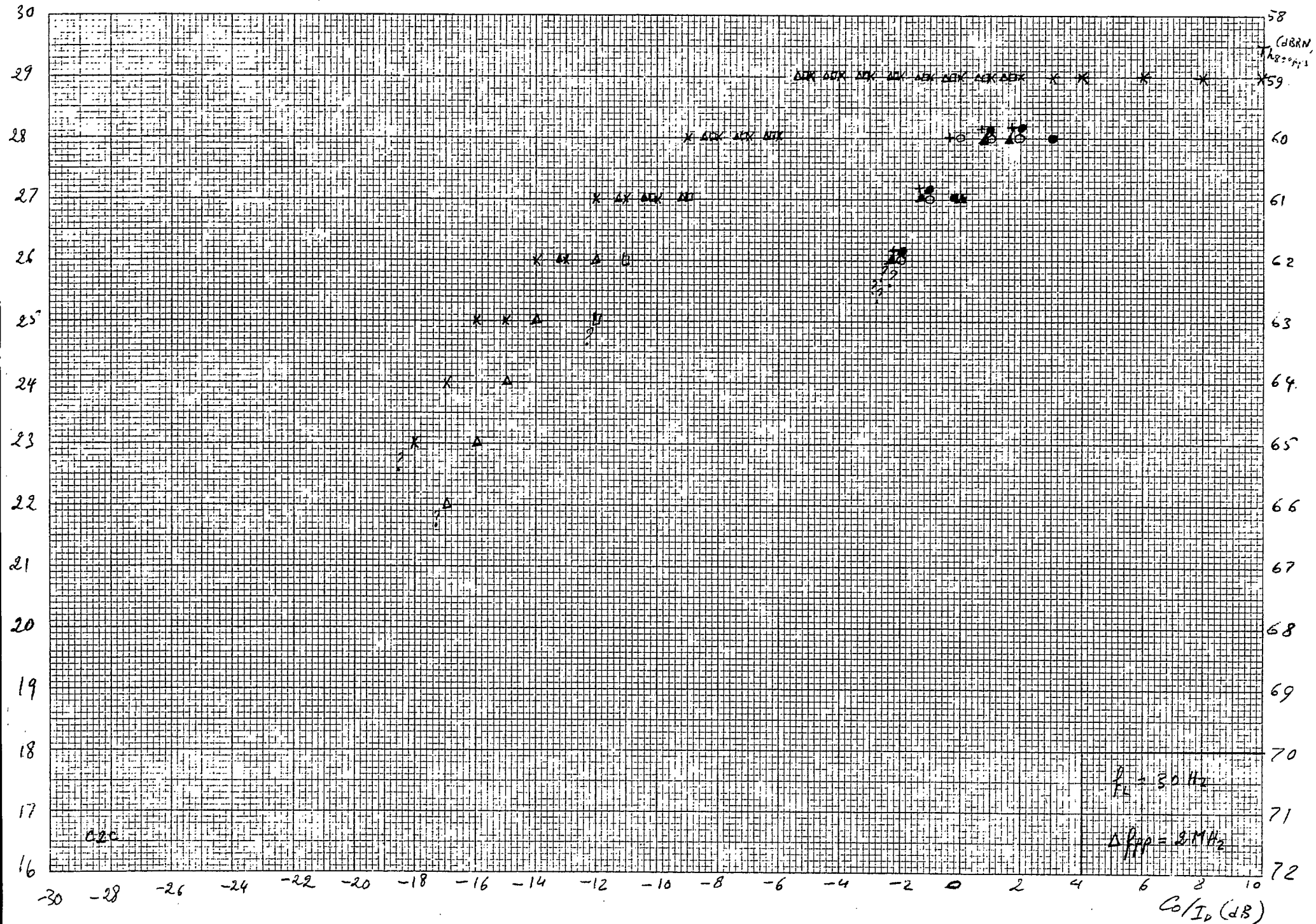
$$S/N(\text{dB}) = 20 \log_{10} \left(\frac{P}{kT_{\text{eff}} B} \right) - T_{\text{h}}(8 \pm 2 \mu\text{s}) (\text{dB RN})$$



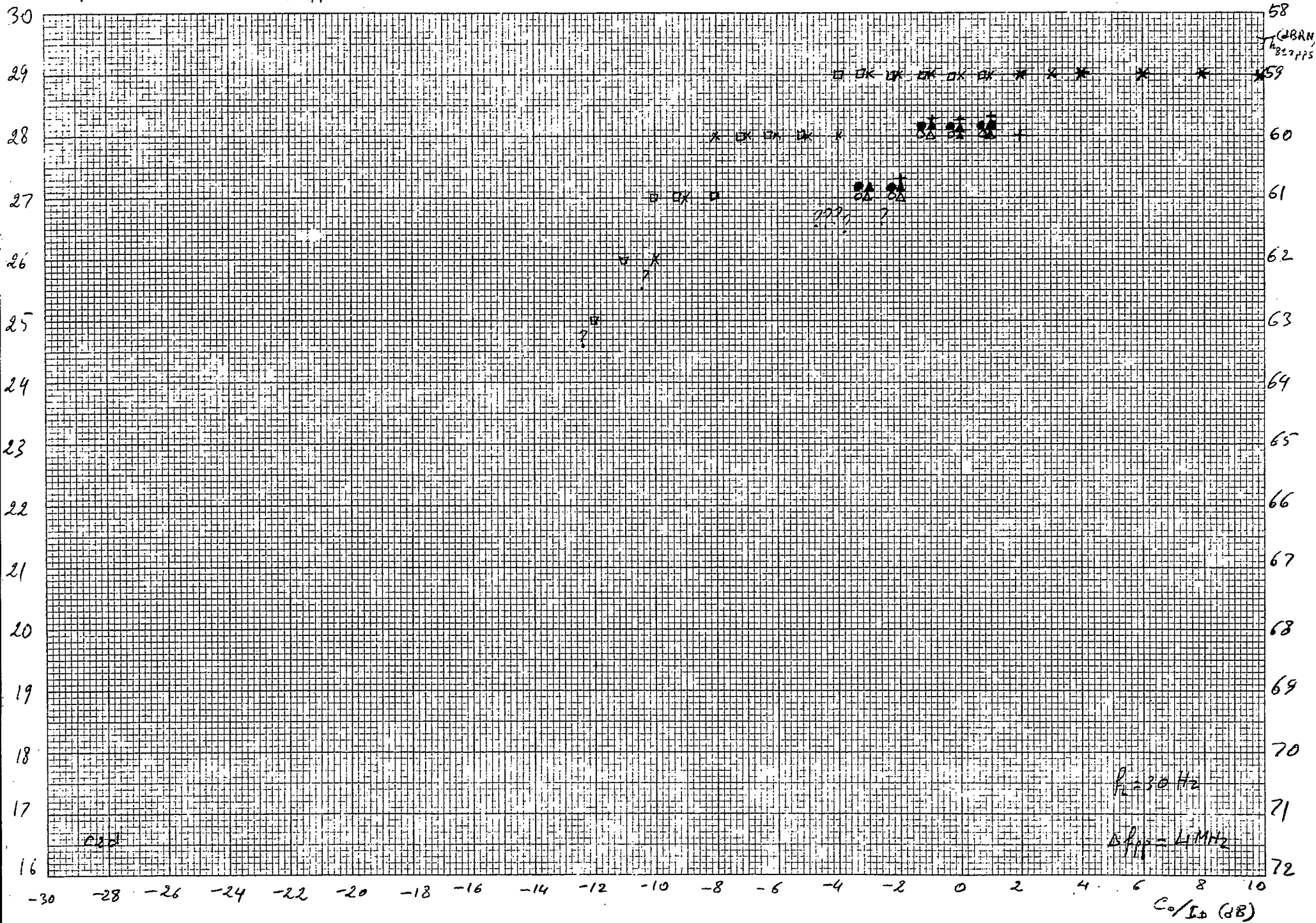
$$S/N \text{ (dB)} = 32 \text{ (dB)} - T_{h8 \pm 2fp3} \text{ (dBKN)}$$



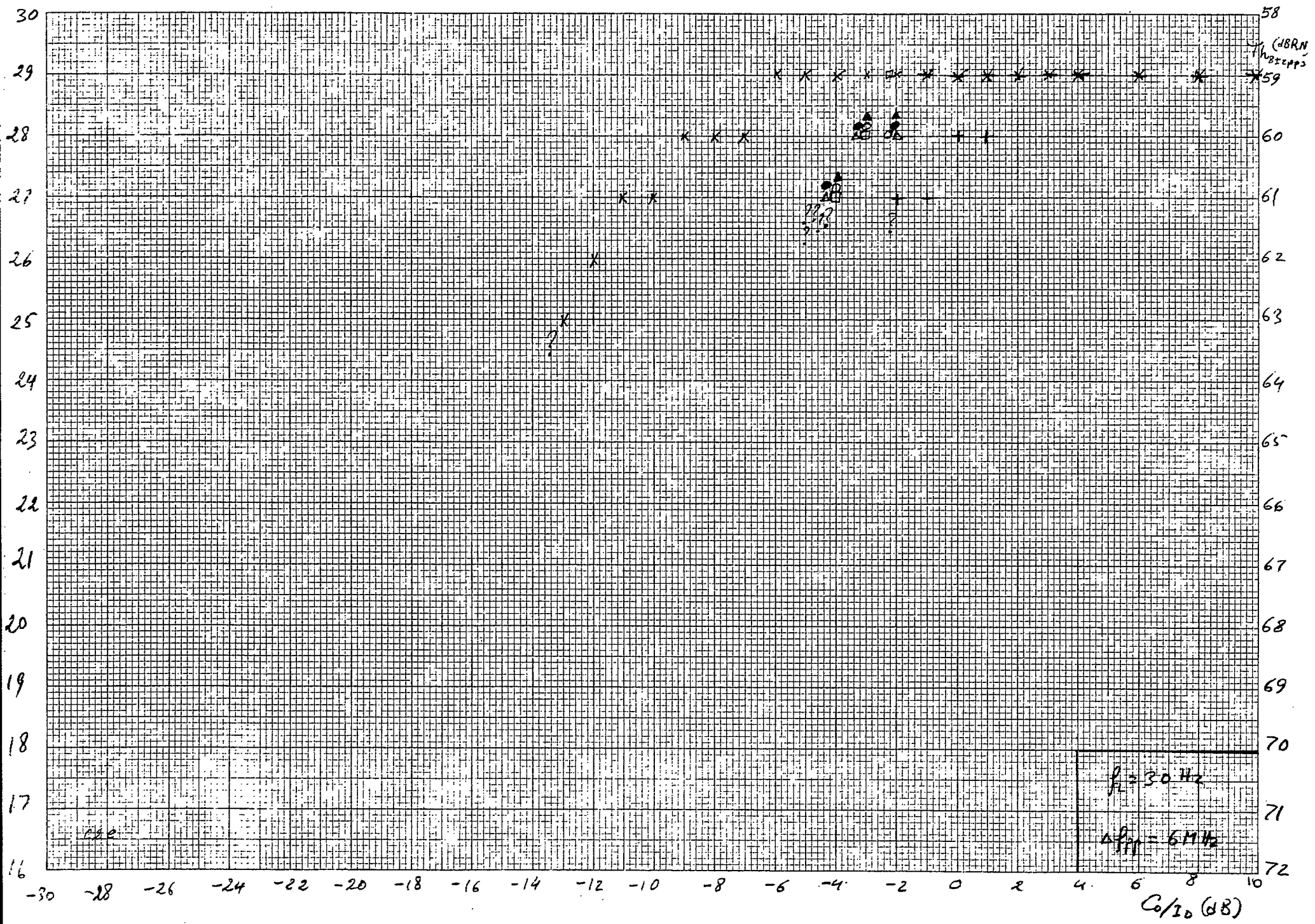
$$S/N(\text{dB}) = 33(\text{dB}) - T_{8\pm 2\text{pps}}(\text{dB RN})$$



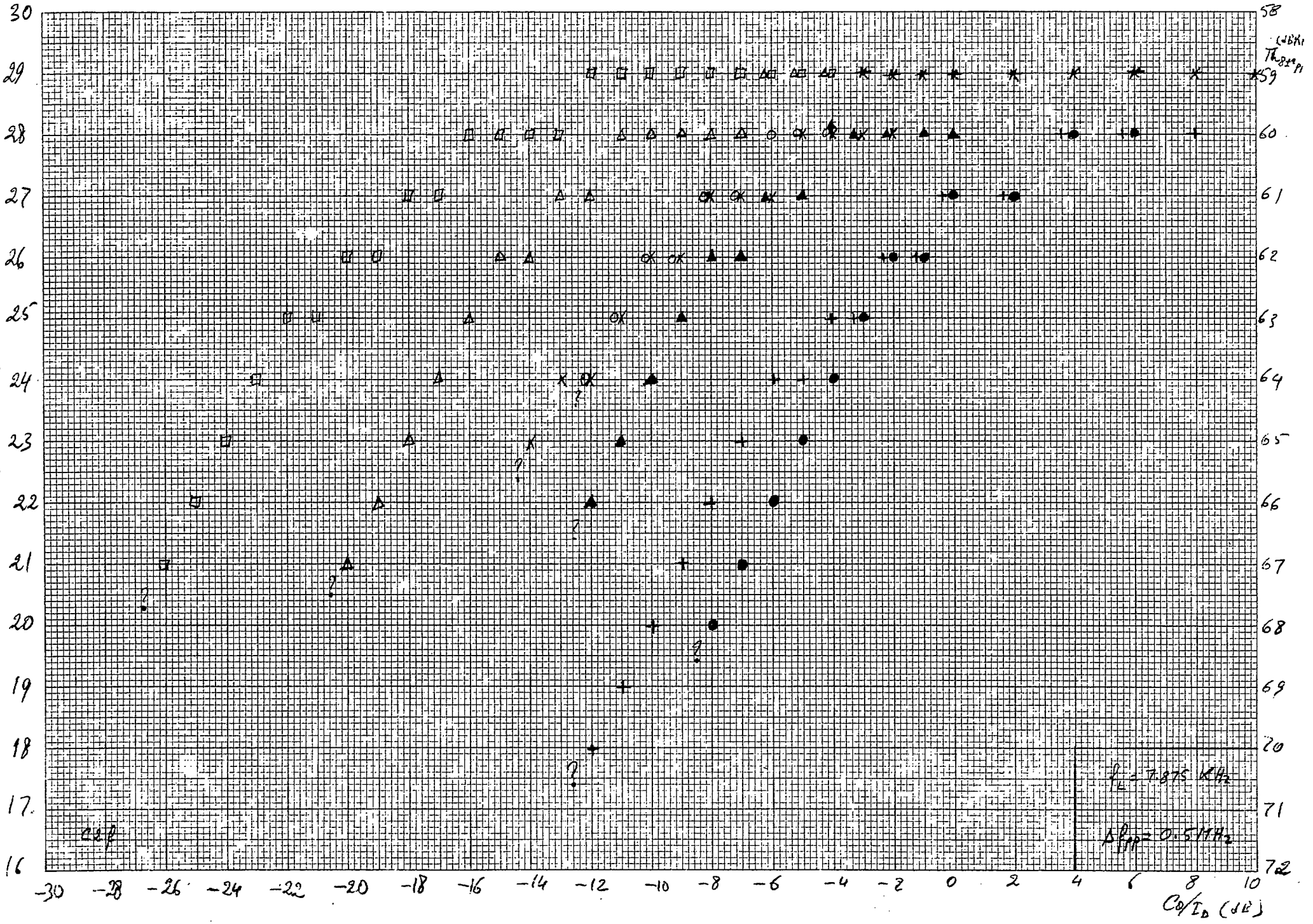
$$S/N \text{ (dB)} = 88 \text{ (dB)} - T_{h_{8-2pp}} \text{ (dBKN)}$$



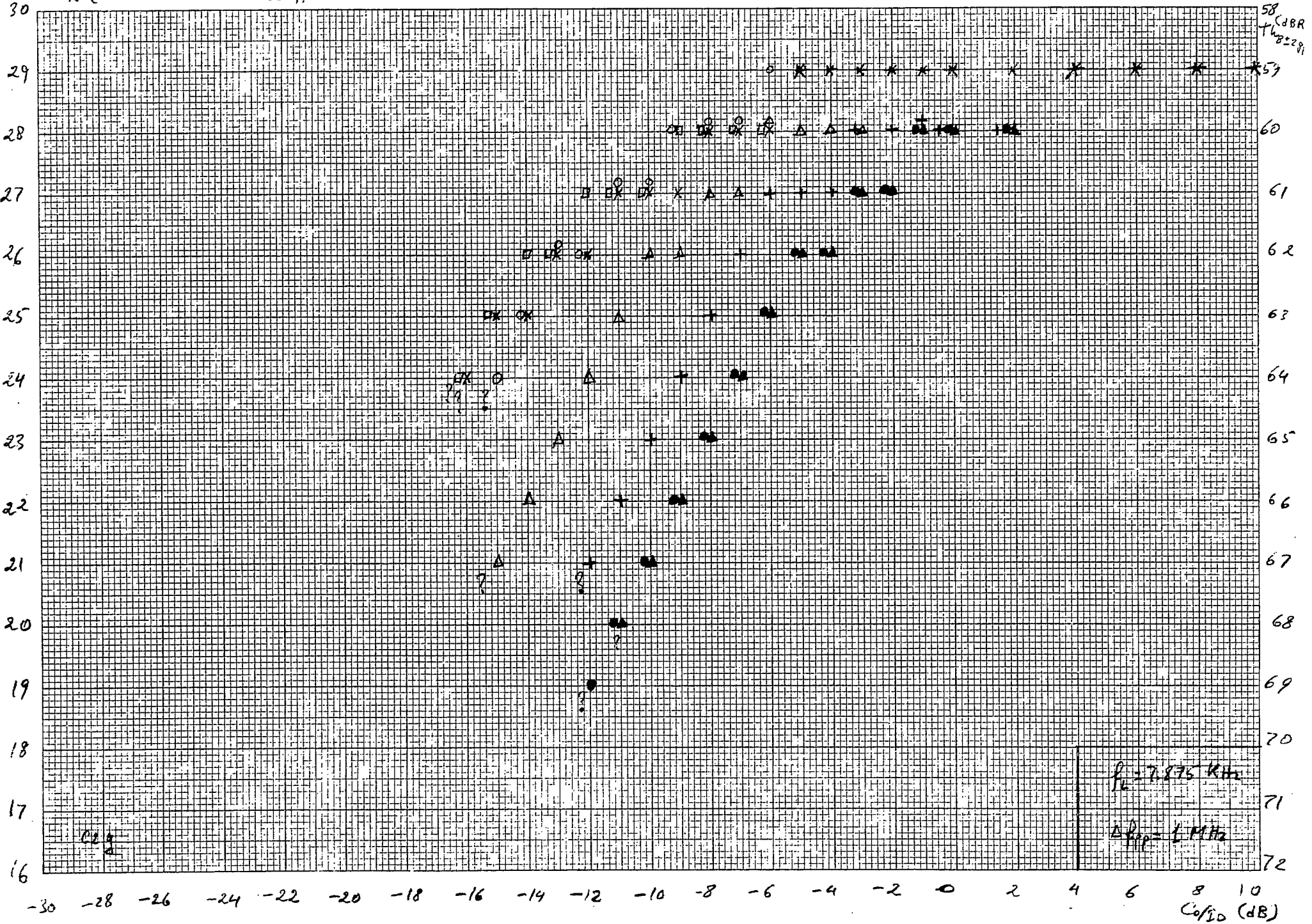
$$S/N (dB) = 82 (dB) - T_{h \pm 2ppp} (dB RN)$$



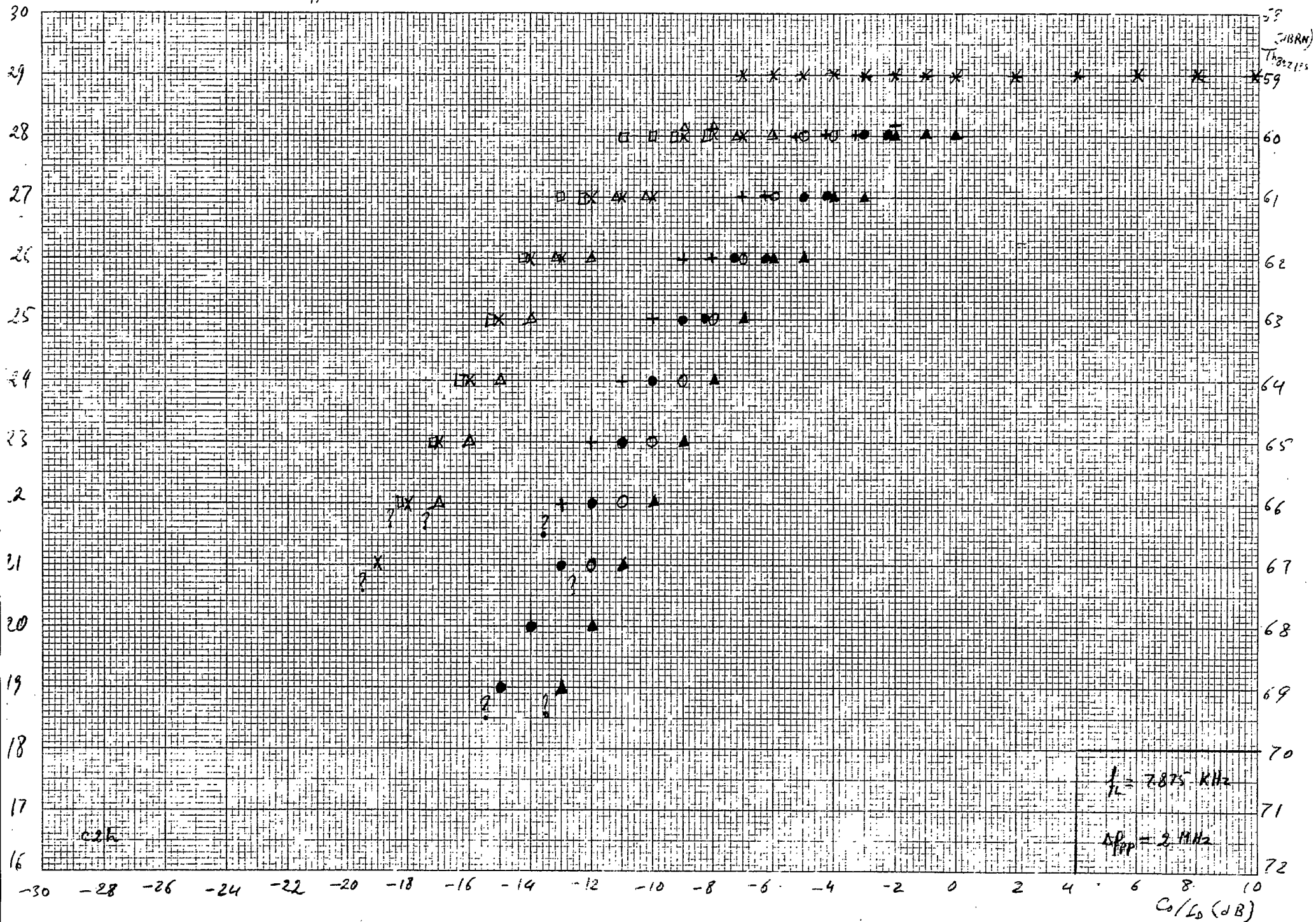
$$S/N (dB) = 83 dB - T_{h3} (dB) \quad (\text{BRN})$$



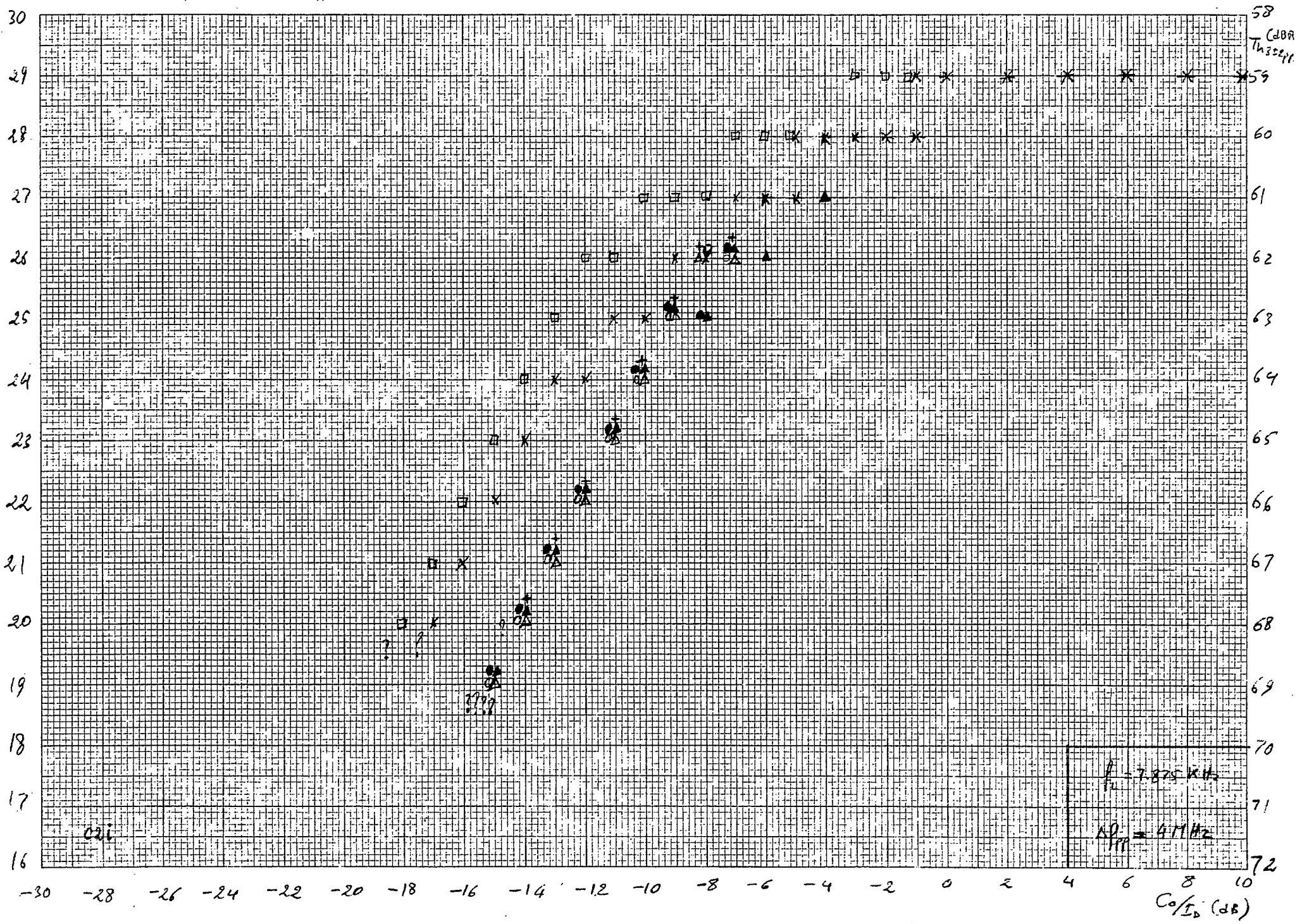
$$S/N (dB) = 88(dB) - T_{h_{8 \pm 5pp}} (dB) \quad (HRRN)$$



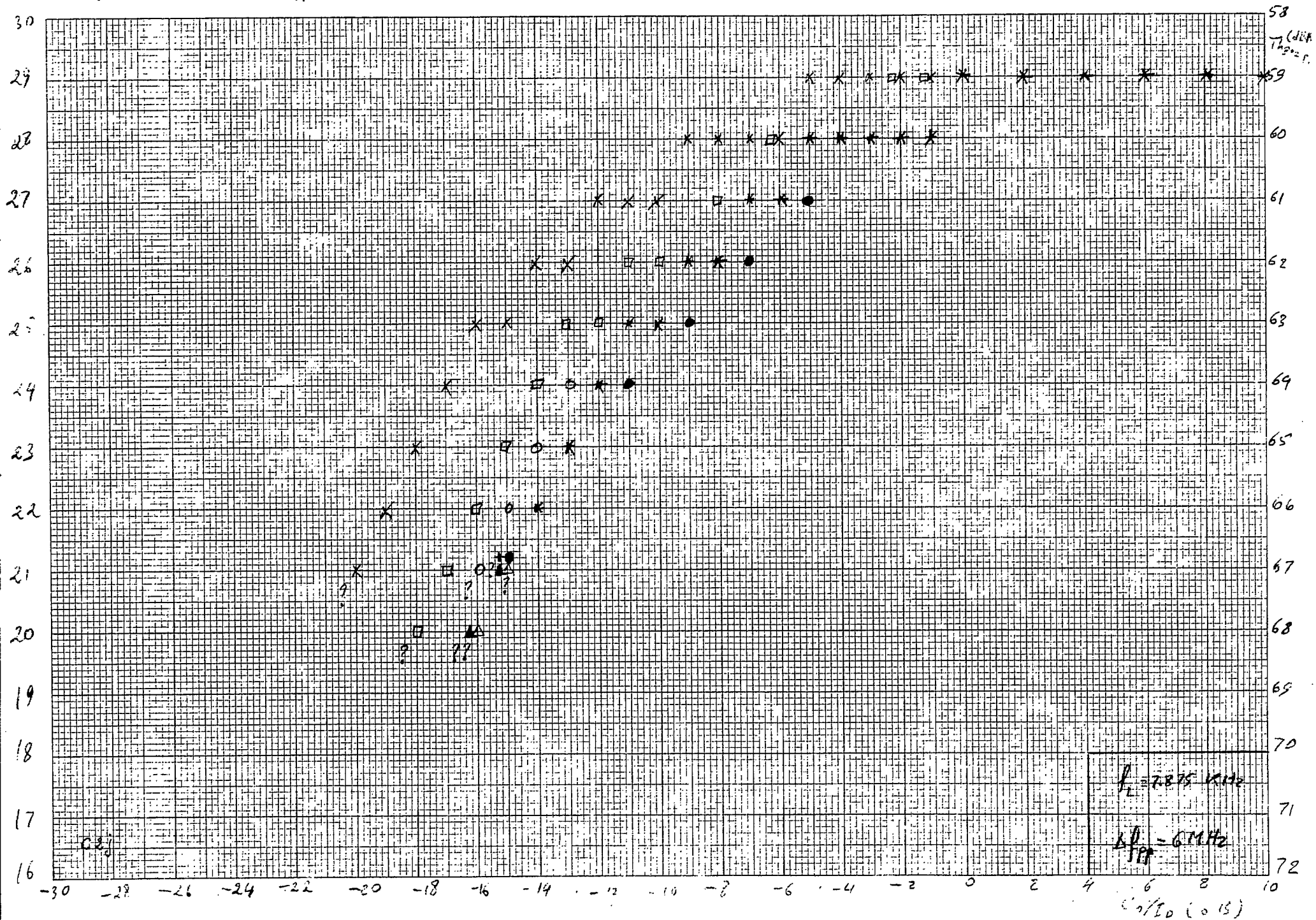
$$S/N \text{ (dB)} = 88 \text{ (dB)} - T_{h_{82.2 \text{ pp}} \text{ (dB RN)}}$$



$S/N (dB) = 83 (dB) - T_{h_{5\pm 2\sigma}}(KRN)$

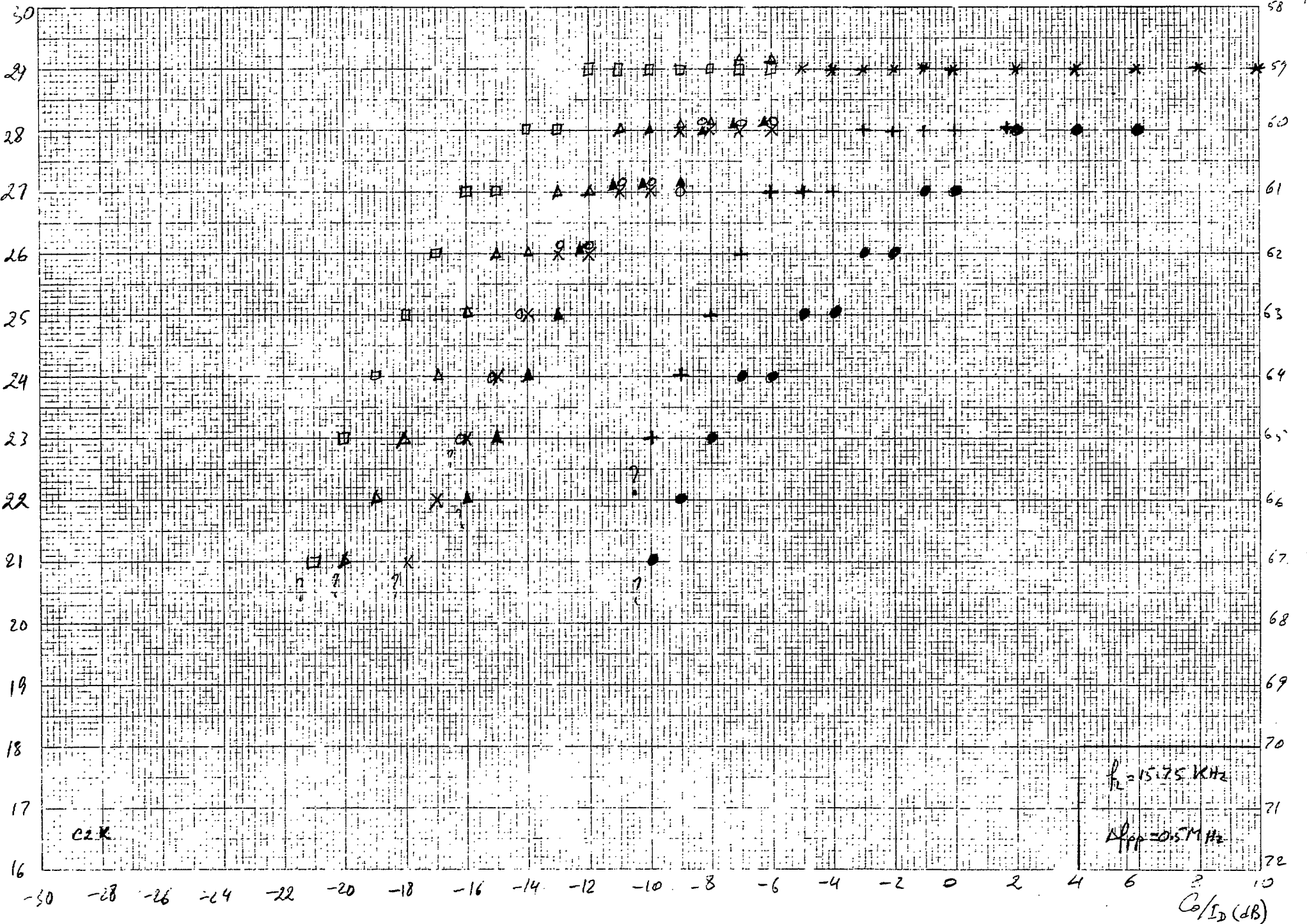


$$S/N (dB) = 33 (dB) - T_{h3} = 2 \text{ ppm} \quad (\text{dB RN})$$

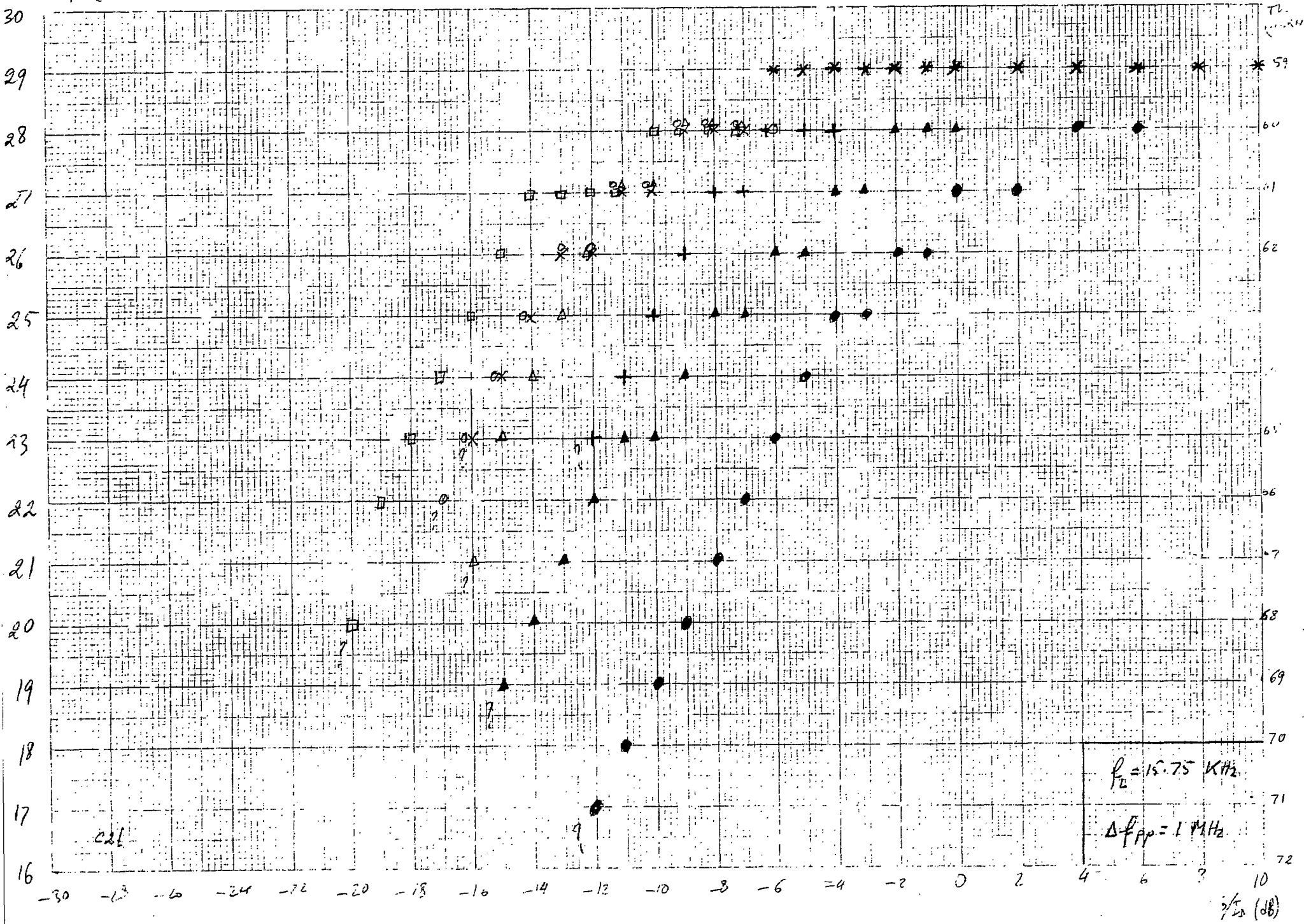


$$S/N (dB) = 23(1.5) - T_{h_{8 \pm 2 PPs}} (JBRN)$$

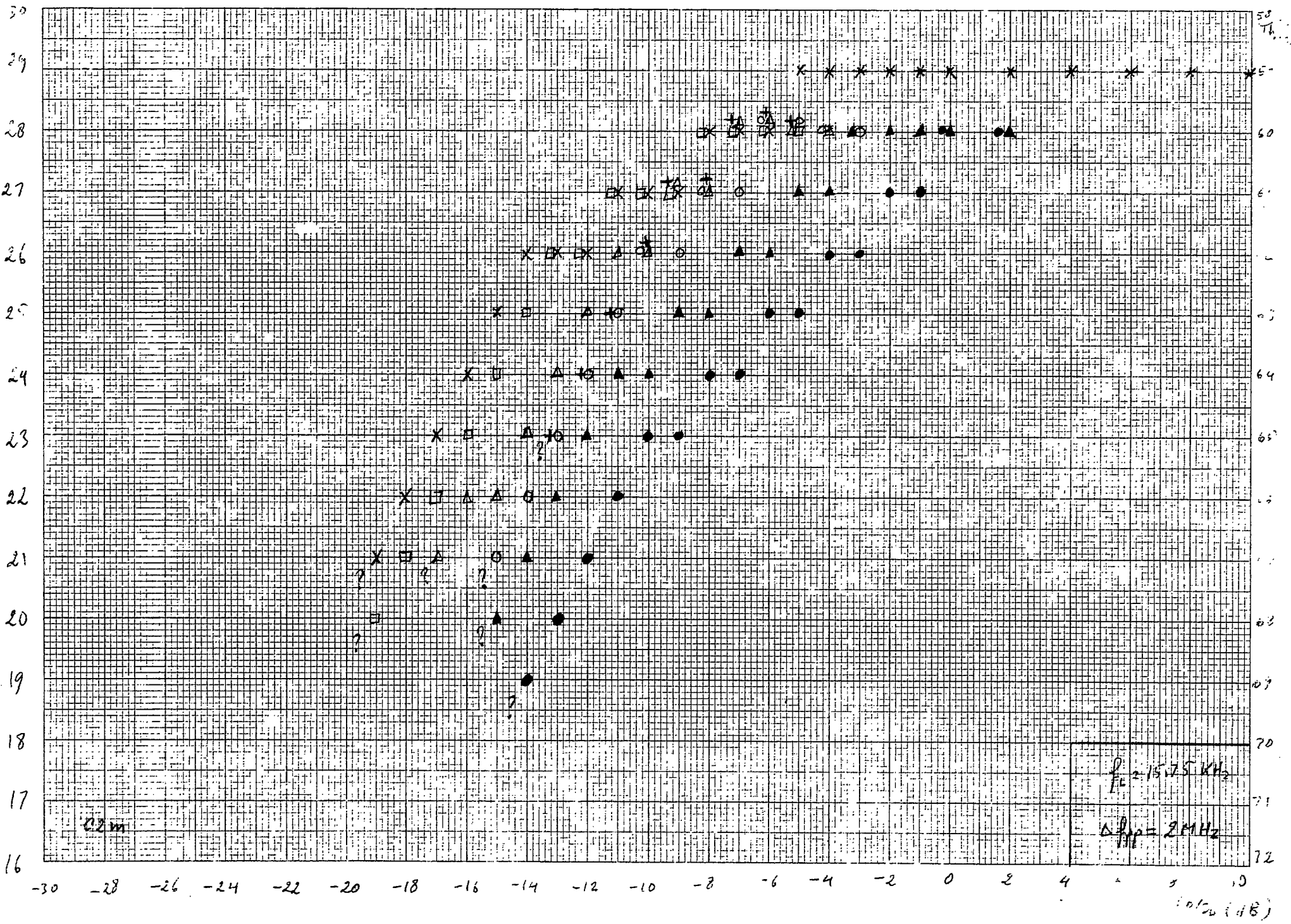
T_h JBRN
8.27



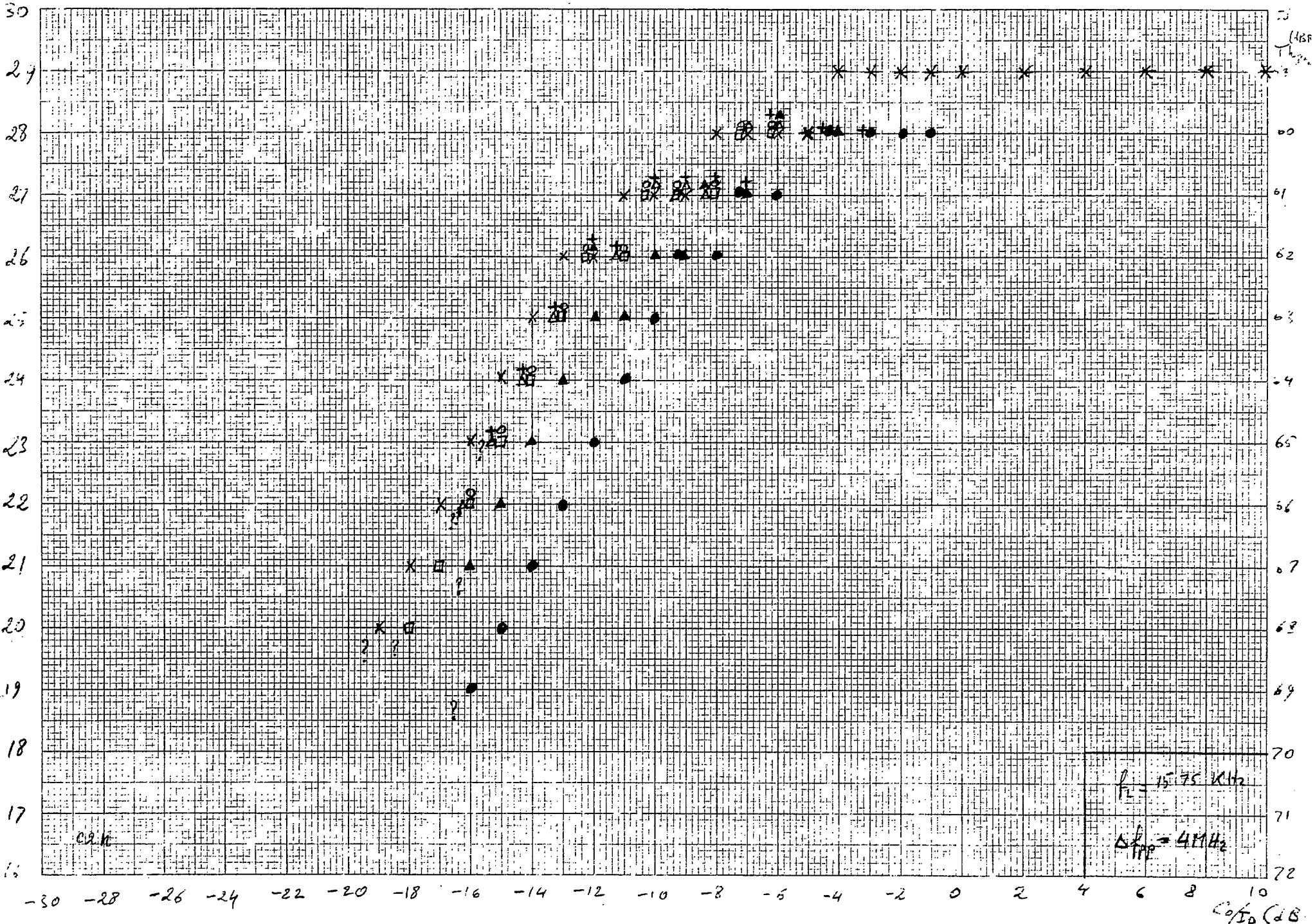
$$S/N \text{ (dB)} = 33 \text{ dB} - T_{h3} = 11.5 \text{ (BRAN)}$$



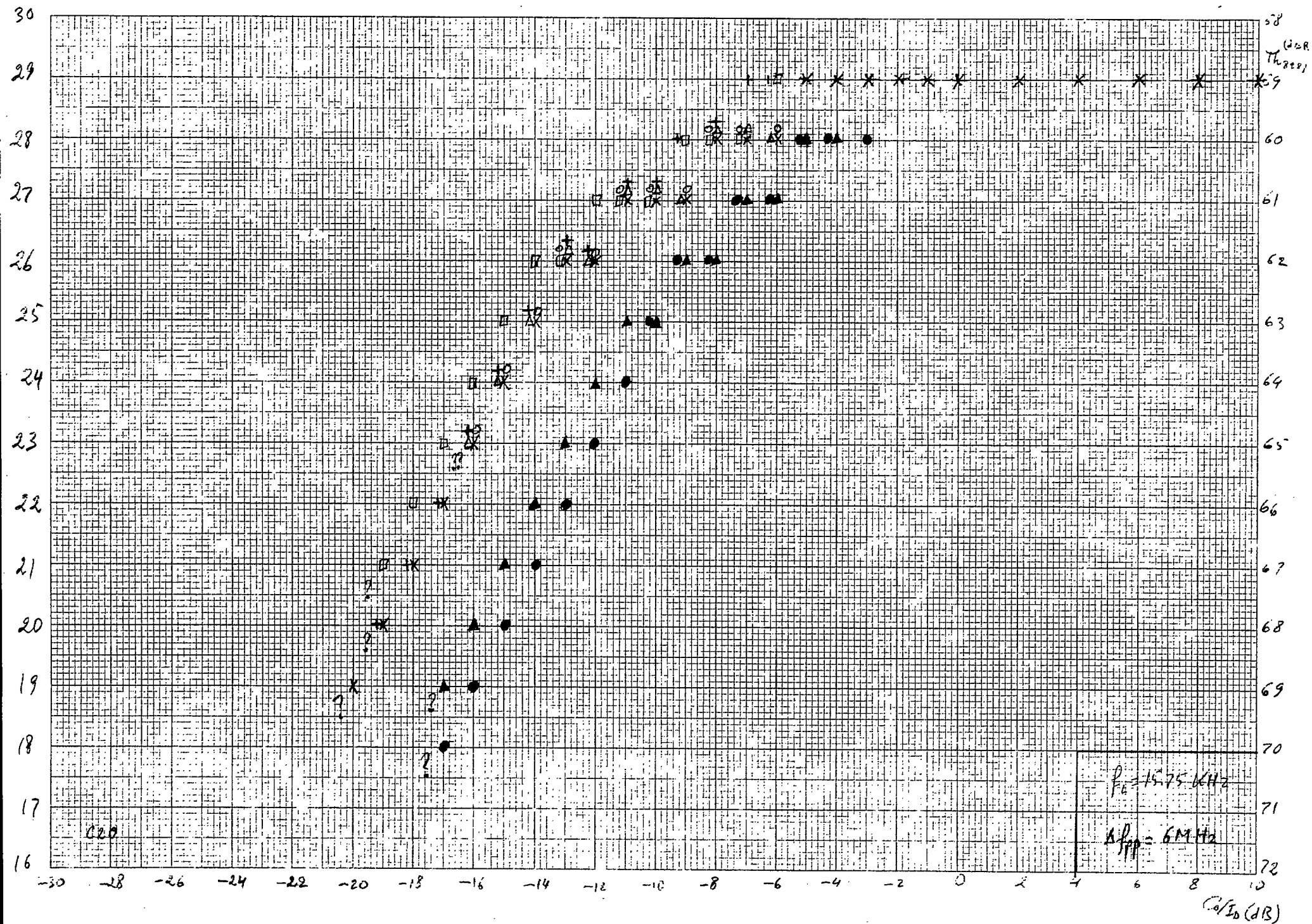
$$S/N(\text{dB}) = 35(\text{dB}) - T_{h3} = 2\text{pps} (\text{dBPN})$$



$$S/N (dB) = 33 (dB) - Th_2 = 2 \text{ pps (JBRN)}$$



$$S/N(\text{dB}) = 33 \text{ dB} - T_{\text{eff}}(\text{dB})$$



DI

4 INTERFERING

CARRIERS

SCPC / PSK

Offset Degradation - defined as the C_0/I_D at which the BER is three times greater than that produced by thermal noise alone at the level $C_0/N = 15$ dB in
33.22 kHz (Channel bandwidth)

$BER_{\text{noise only}} = 3.3 \times 10^{-6}$; $BER_{I+N} = 10 \times 10^{-6} \pm 10\%$

f_c		22.5 kHz	100 kHz	500 kHz	1 MHz	2 MHz	3 MHz	3.6 MHz
Δf_{pp}	f_L							
0.5 MHz	30 Hz	+8.5 dB	+8.5 dB	+1 dB	+1.5 dB	+1 dB	-2.5 dB	-1 dB
	7.875 kHz	+7 dB	+7 dB	+1 dB	0 dB	-5 dB	-4.2 dB	-1 dB
	15.75 kHz	+4 dB	+5.2 dB	-1 dB	-2 dB	-4 dB	-7 dB	-4 dB
1 MHz	30 Hz	+6.5 dB	+6.5 dB	+7 dB	0 dB	0 dB	-3 dB	
	7.875 kHz	+5 dB	+5 dB	+3 dB	-1 dB	-0.5 dB	-3 dB	-4 dB
	15.75 kHz	+3 dB	+3 dB	0 dB	-3 dB	-2 dB	-6.2 dB	-5 dB
2 MHz	30 Hz	+5 dB	+5 dB	+5 dB	+6 dB	-1 dB	+1 dB	-3.5 dB
	7.875 kHz	+2 dB	+2 dB	+2 dB	+1.5 dB	-2 dB	-4 dB	-4 dB
	15.75 kHz	+0.5 dB	+1 dB	+1 dB	-1 dB	-3 dB	-7 dB	-5 dB
4 MHz	30 Hz	+4 dB	+4 dB	+4 dB	+4 dB	+4 dB	+1 dB	+3 dB
	2.875 kHz	+1 dB	+1.5 dB	0 dB	-1 dB	+0.5 dB	-4.5 dB	-3 dB
	15.75 kHz	-0.5 dB	-0.5 dB	-0.5 dB	-0.5 dB	-0.5 dB	-5 dB	-5 dB
6 MHz	30 Hz	+3 dB	+3 dB	+3.5 dB	+3 dB	+3 dB	+3 dB	-5.5 dB
	7.875 kHz	-0.5 dB	0 dB	+2 dB	0 dB	-1 dB	-1 dB	-6 dB
	15.75 kHz	-1.5 dB	-1.5 dB	-2 dB	-2 dB	-4 dB	-2 dB	-5 dB

SI

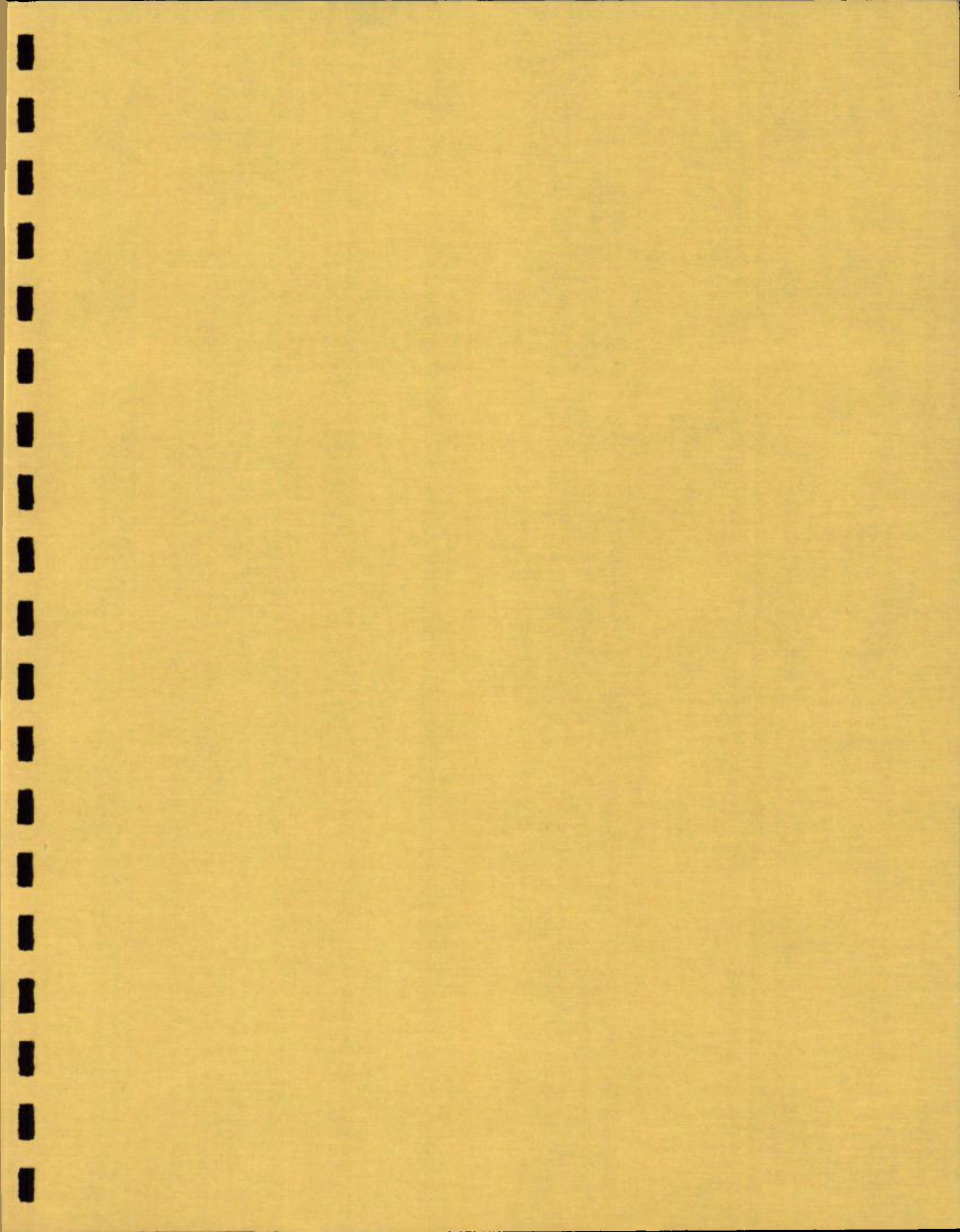
LIVE TV SIGNAL

INTERFERENCE

$$f_L = 30 \text{ Hz} ; \Delta f_{FP} = 1 \text{ MHz}$$

$$C_0/I_0 = 0 \text{ dB} ; C_0/N \rightarrow \infty$$

$$f_c = \left\{ \begin{array}{l} \text{a)} +2 \text{ MHz} \\ \text{b)} +500 \text{ KHz} \\ \text{c)} -1 \text{ MHz} \end{array} \right\} \begin{array}{l} \approx 20 \text{ sec} \\ \text{count time} \end{array}$$
$$\left. \begin{array}{l} \text{d)} -1 \text{ MHz} \end{array} \right\} \approx 3 \text{ min} \\ \text{count time.}$$

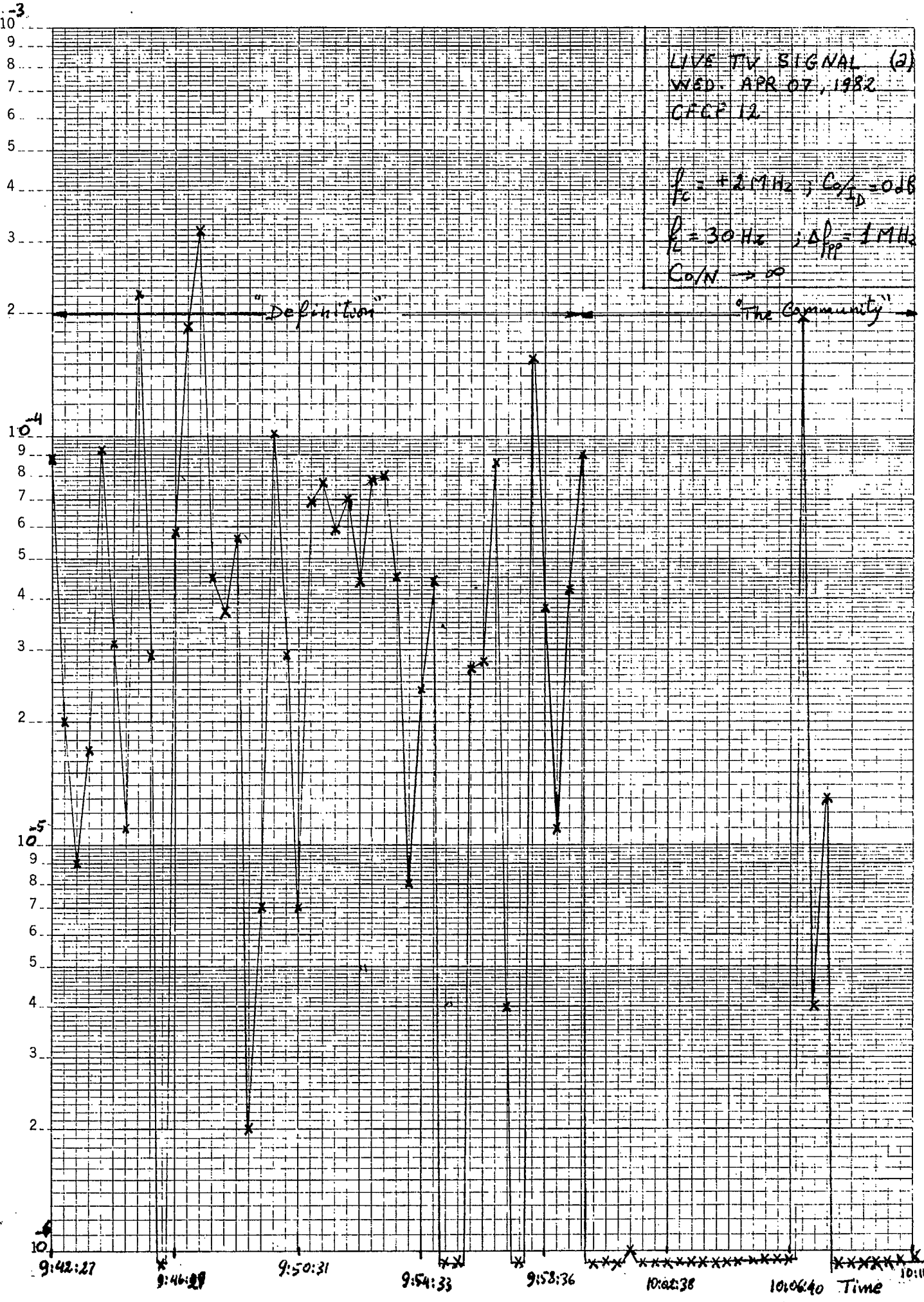


LIVE TV SIGNAL (2)
 WED. APR 07, 1982
 CFCF 12

$f_c = \pm 2 \text{ MHz}$; $C_{0.1} = 0 \text{ dB}$

$f_L = 30 \text{ Hz}$; $\Delta f_{\text{IFF}} = 1 \text{ MHz}$

$C_{0/N} \rightarrow \infty$



40 5490
 KEUFFEL & ESSER CO. MADE IN U.S.A.
 UNITS X 70

BER

10^{-3}

9

8

7

6

5

4

3

2

1

0

10^{-4}

9

8

7

6

5

4

3

2

1

0

10^{-5}

9

8

7

6

5

4

3

2

1

0

10^{-6}

10

LIVE TV SIGNAL (b)
Same as (a) except $f_c = +500\text{KHz}$

"Rhode"

45 5490

SE... GART... C...
KEUFFEL & ESSER CO. MADE IN U.S.A.

10:32:14

10:36:16

10:40:18

10:44:21

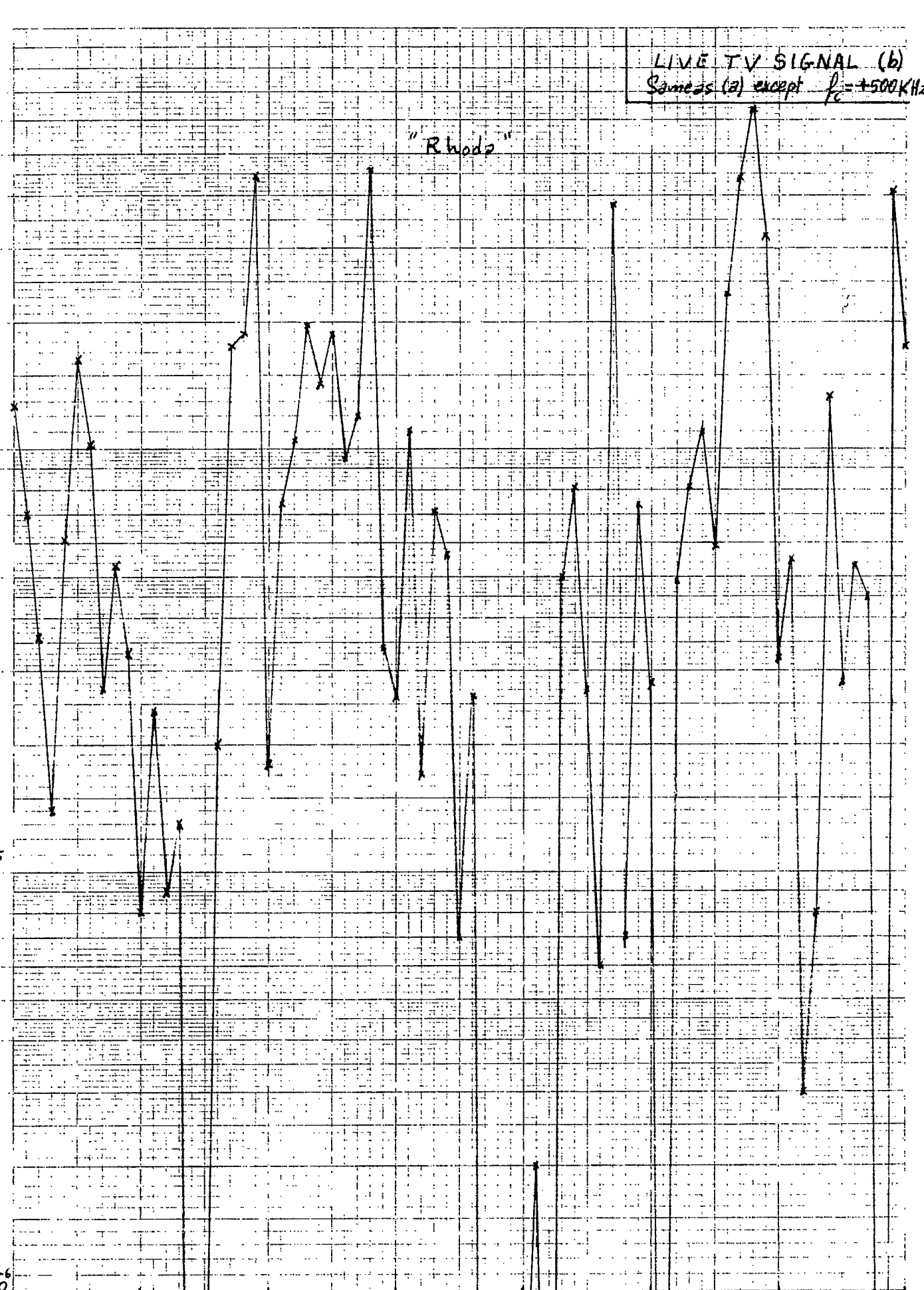
10:48:23

10:52:25

10:56:28

11:00:30

Time

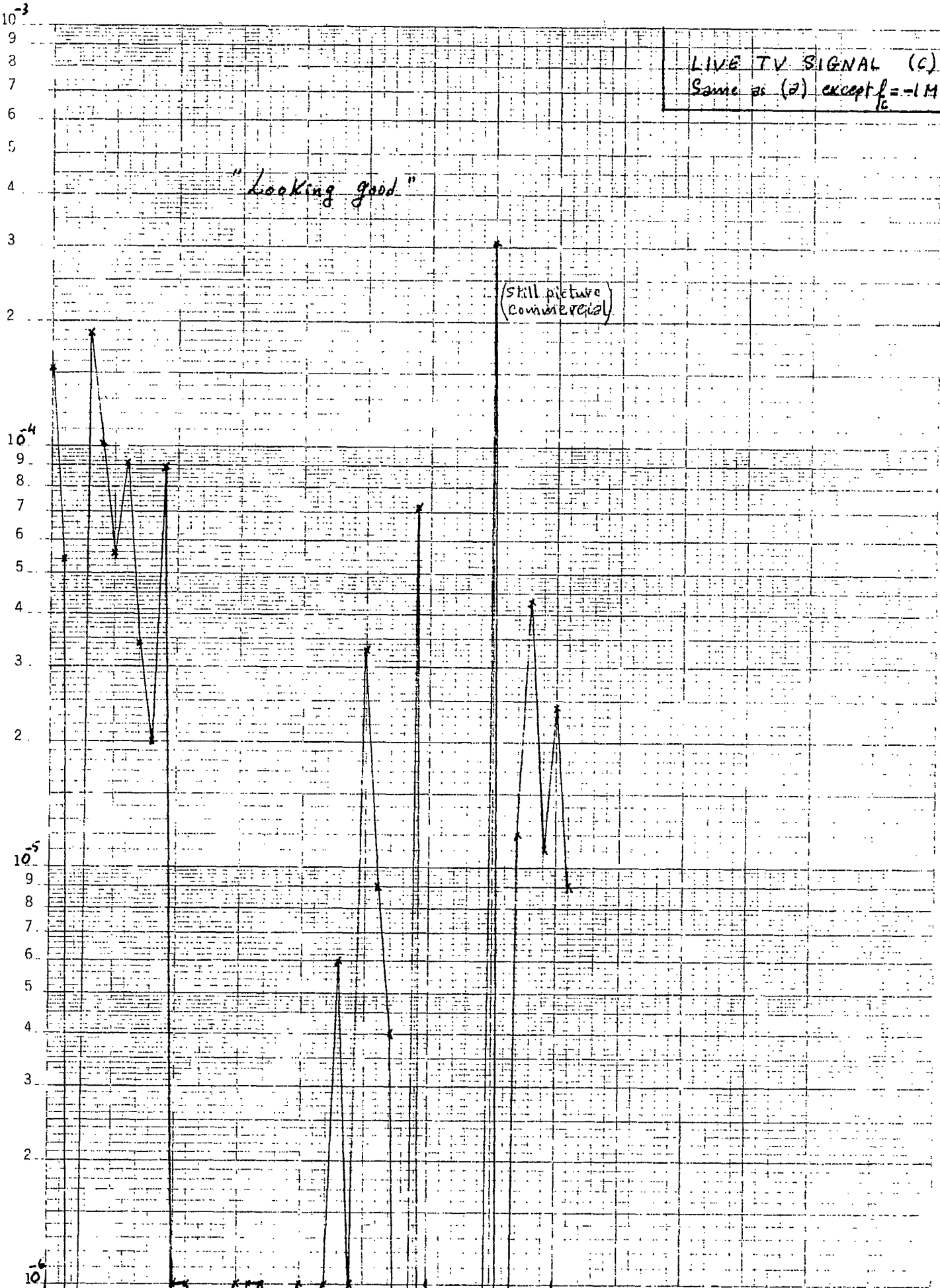


BER

LIVE TV SIGNAL (C)
Same as (2) except $f_c = -1 \text{ MHz}$

"Looking good"

(Still picture
Commercial)



46 5490

REUFFEL & ESSER CO. DIVISION

Time

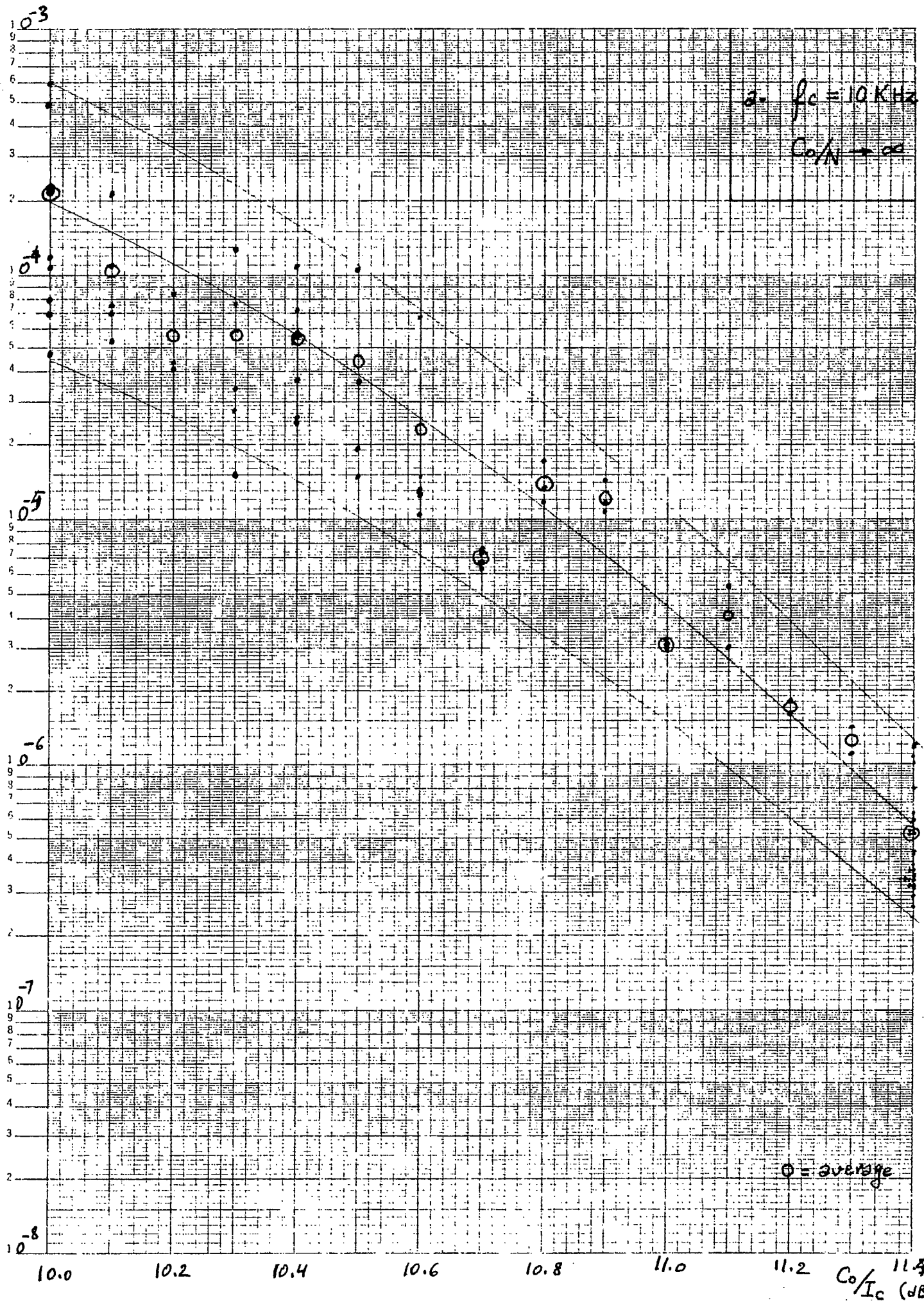
S2

UNMODULATED, UNDISPERSED

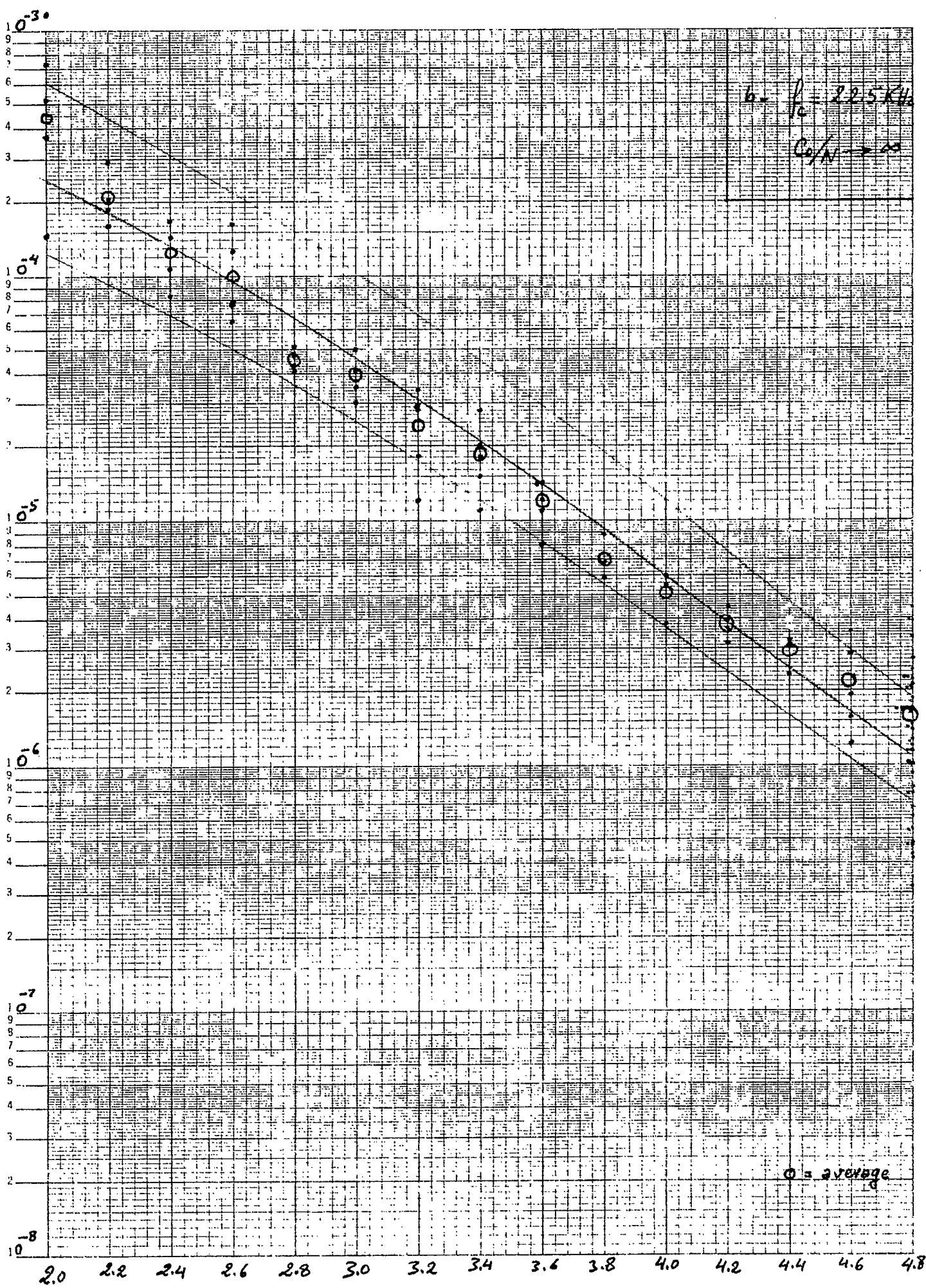
INTERFERING CARRIER $\omega - f_c = 10 \text{ KHz}$

$b - f_c = 22.5 \text{ KHz}$

BER



BER



S3

RADAR LIKE
INTERFERING SIGNAL

PW = 30 μ s

PRF = 3 kHz

SYMBOL	f_c (kHz)
X	0 (MAIN PEAK)
●	40 (1 st NULL)
▲	52 (1 st SIDE PEAK)
■	70 (2 nd NULL)
□	10 kHz

S/N (dB)

40

39

38

37

36

35

34

33

-30

-27

-24

-21

-18

-15

-12

-9

-6

-3

0

3

6

9

12

15

20

24

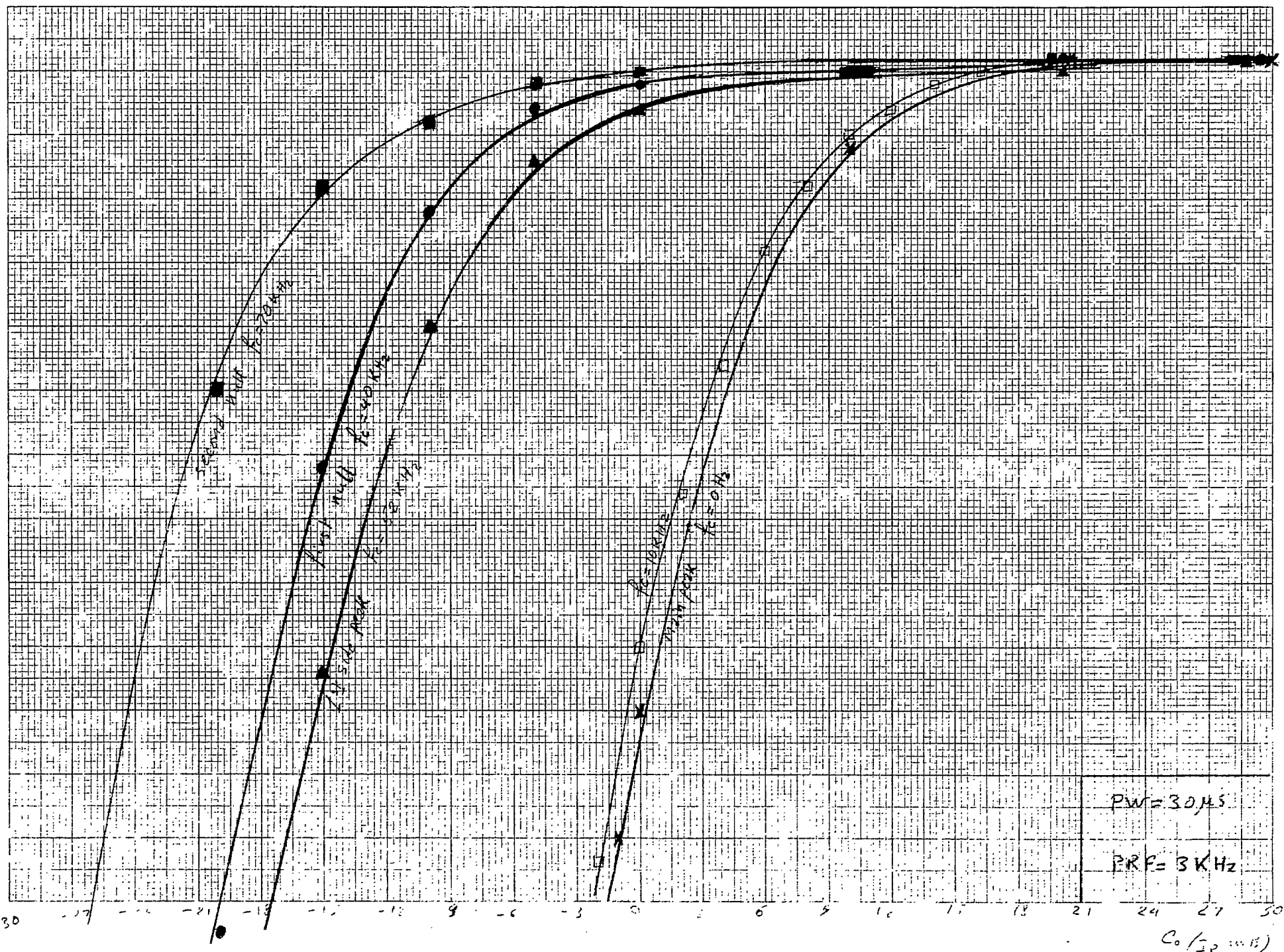
27

30

PWR = 30 μ S

PRF = 3 KHz

C_0 (IP = 15)



S4

S/N_{LOAD}ED vs C₀/I₀/I_V

I₀ : MODULATED BY DISPERSAL
SIGNAL ONLY

I_V : NO MODULATION AT ALL

SYMBOL	LEGEND
O	NO MODULATION AT ALL
Δ	f _L = 30 Hz
X	f _L = 15.75 Hz

? = SPURIOUS READINGS

Δf_{pp} = 1 MHz

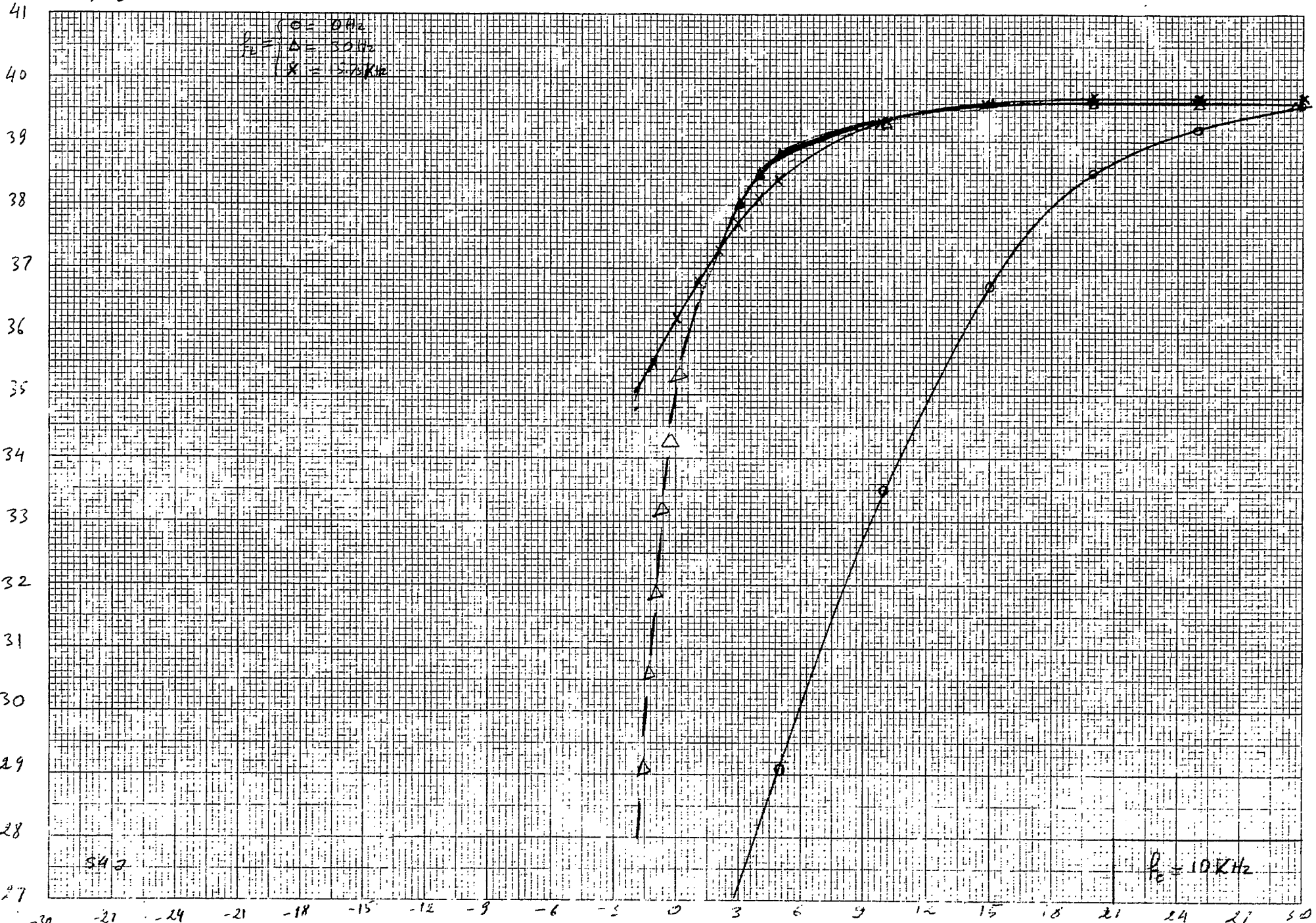
S/N_L (dB)

$f_c = 0 \text{ Hz}$
 $f_m = \Delta = 30 \text{ Hz}$
 $f_x = 50 \text{ kHz}$

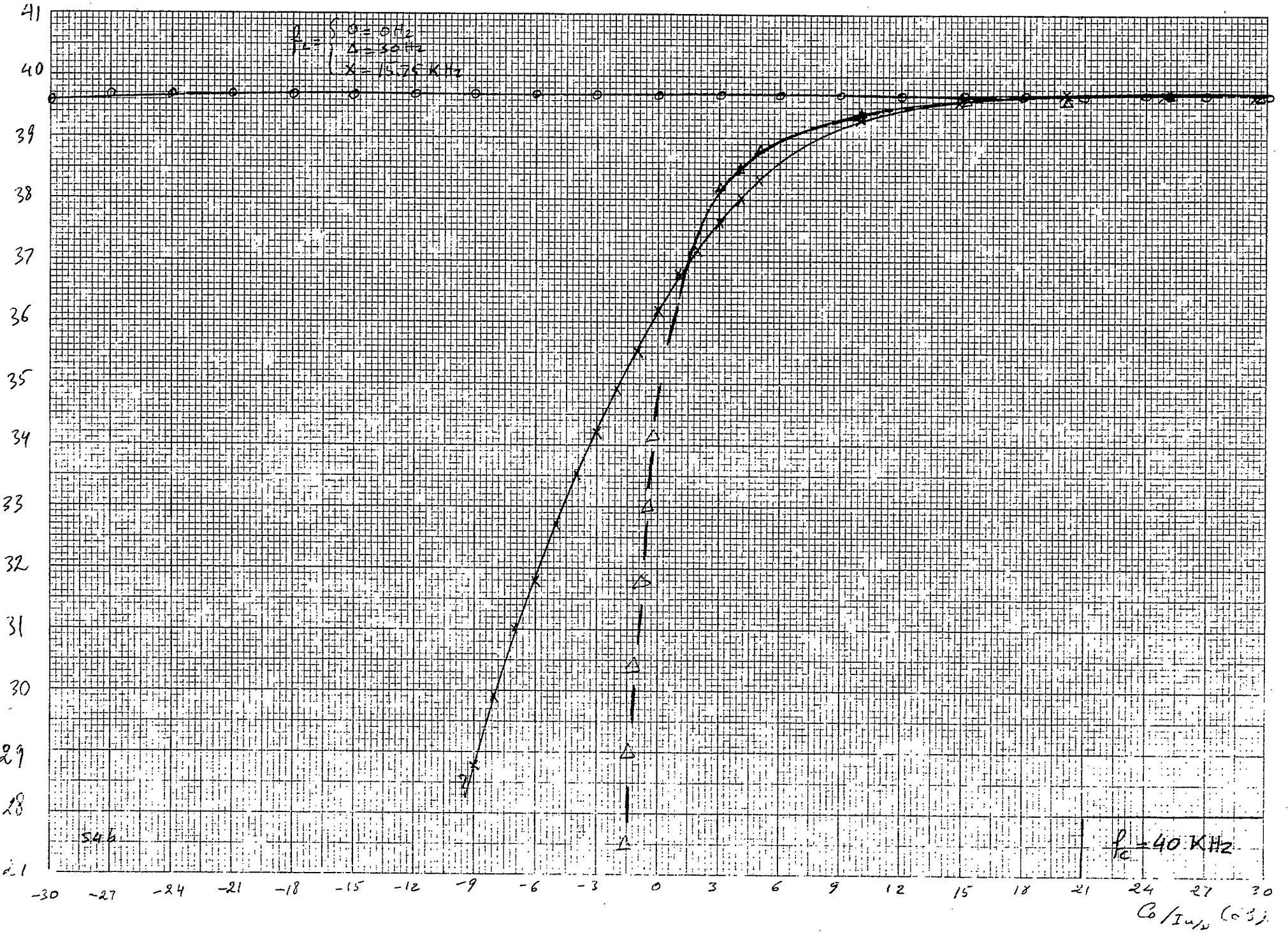
547

$f_c = 10 \text{ kHz}$

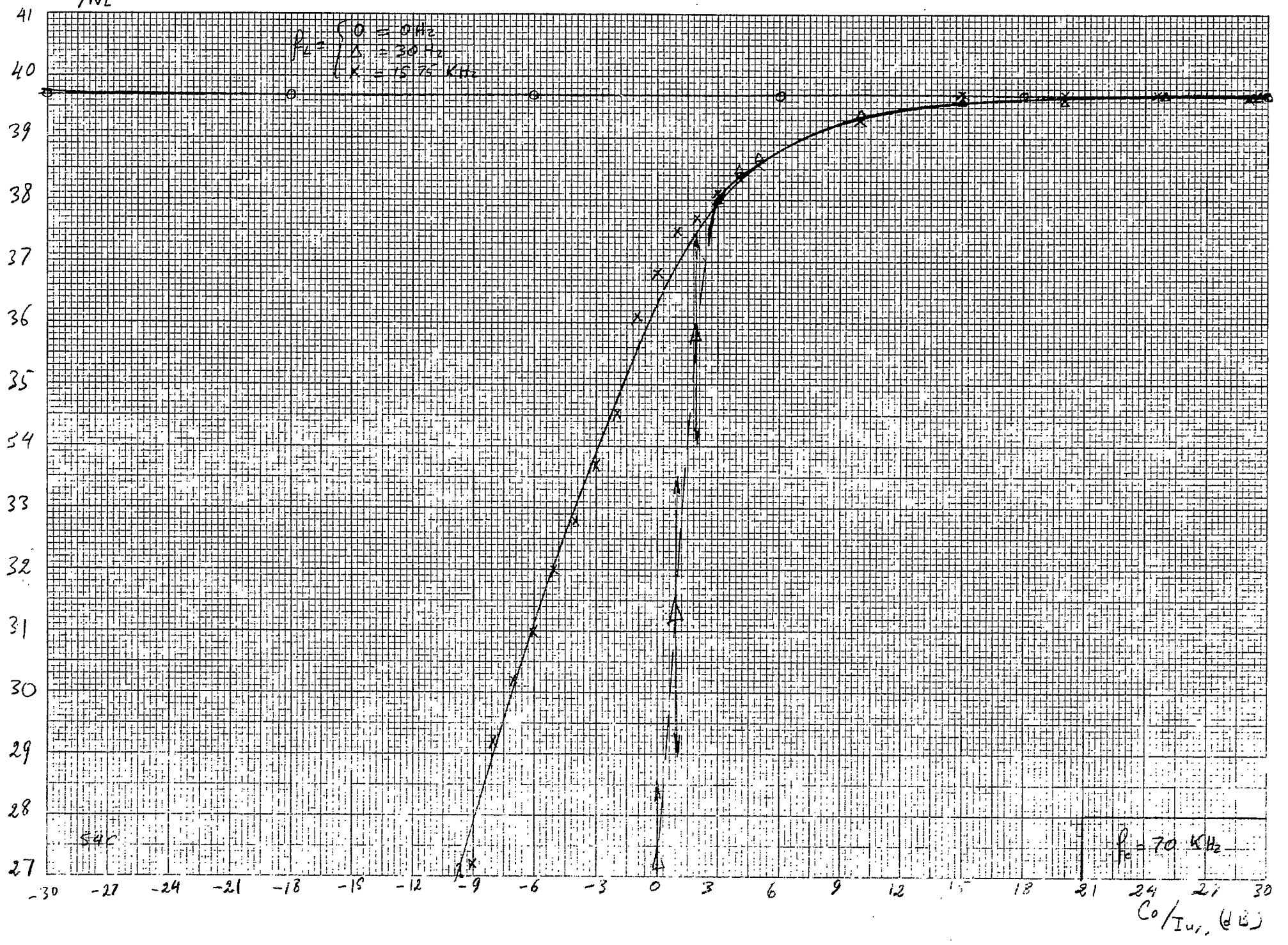
C_0/I_{wp} (dB)



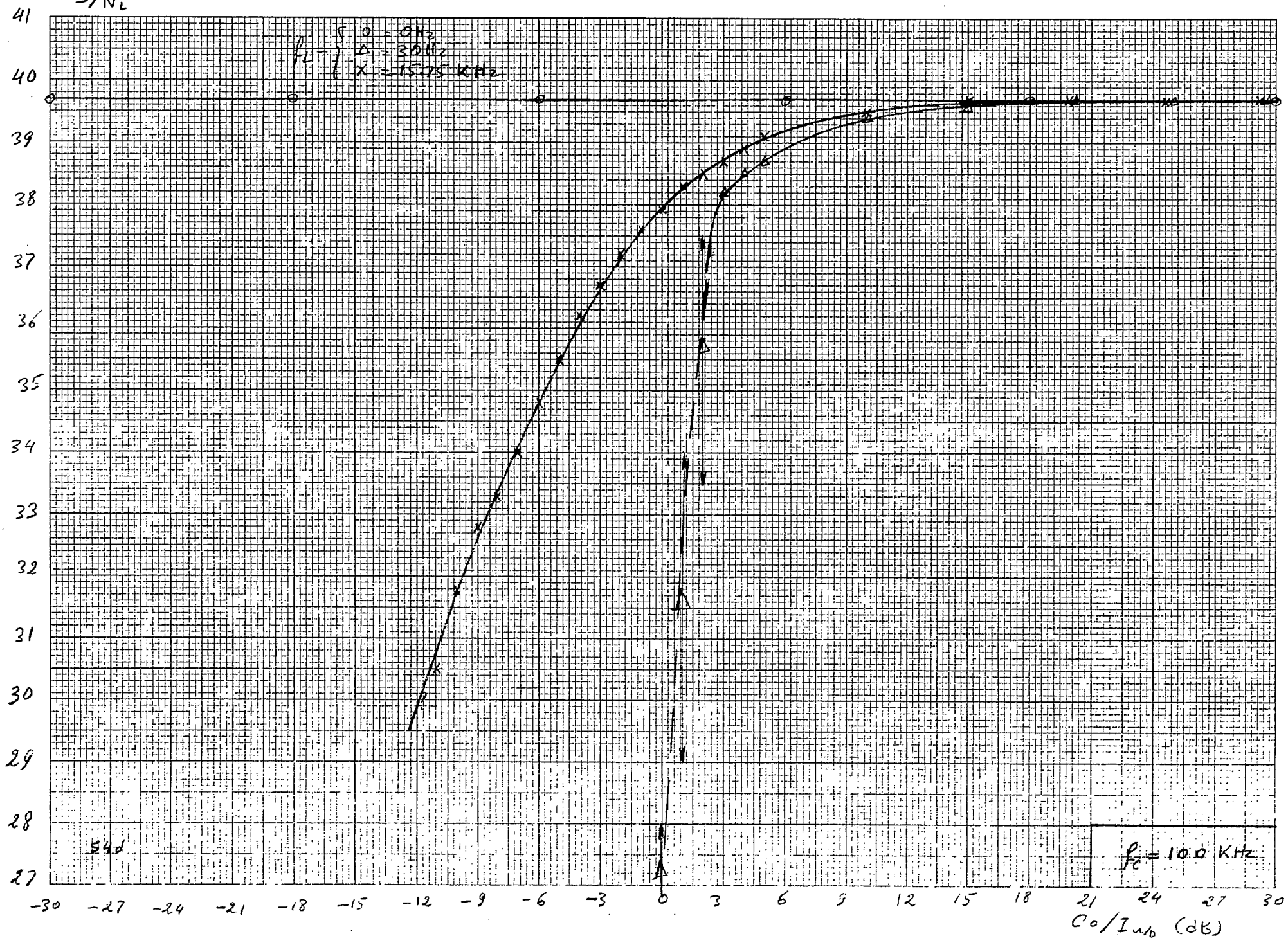
S/N_L (dB)



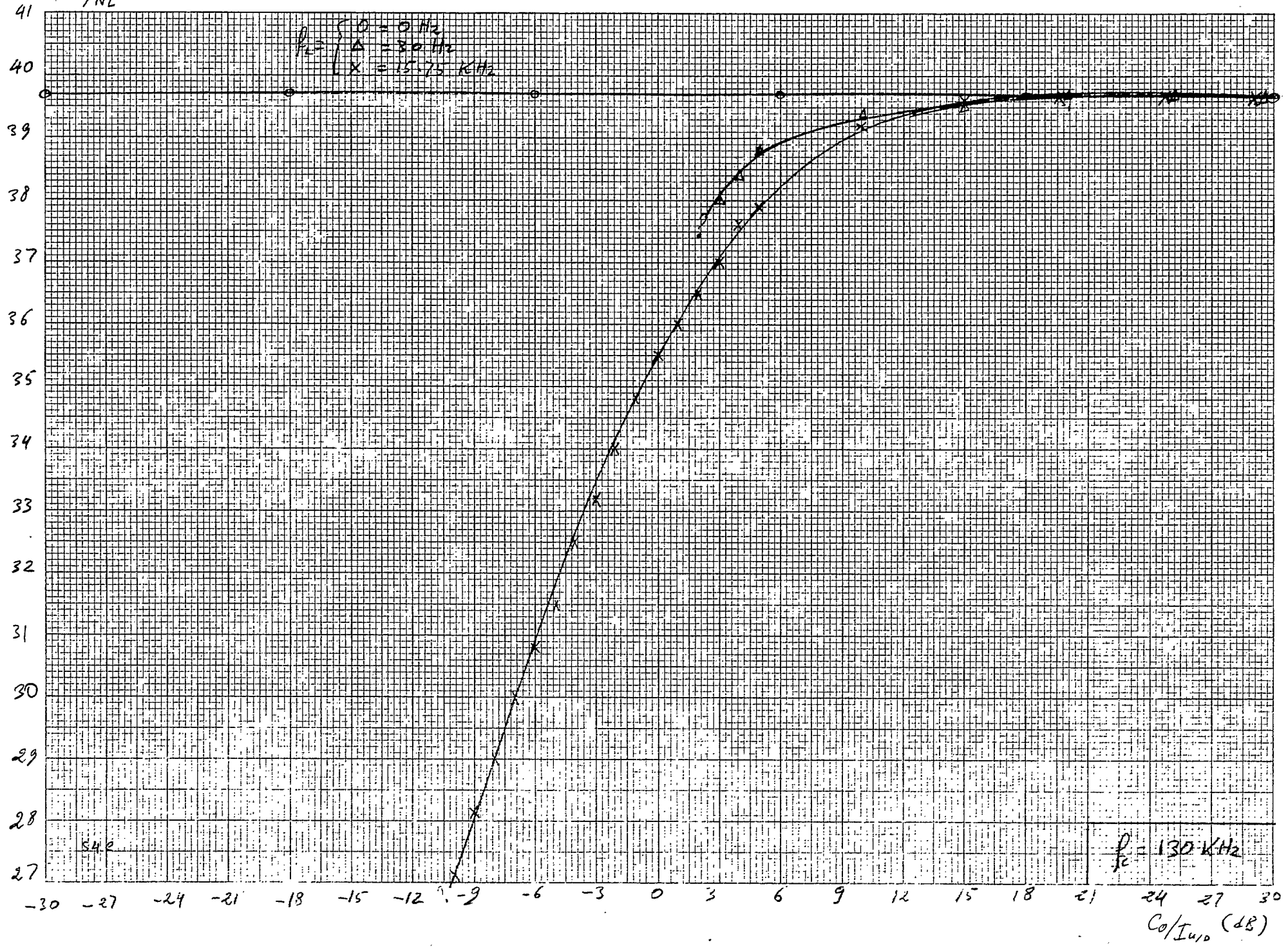
S/N_L (dB)



S/N_L (dB)



S/N_L (dB)

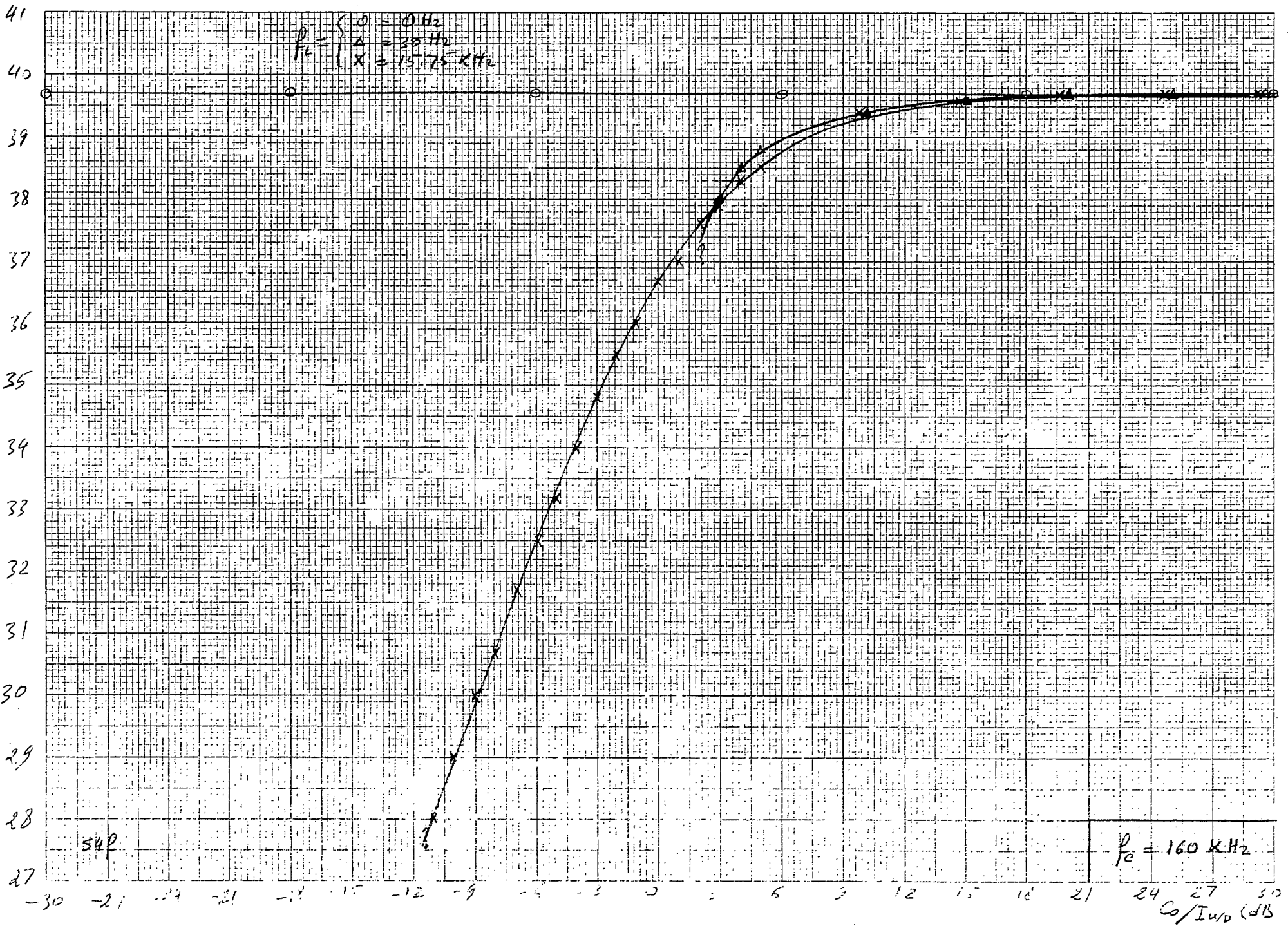


542

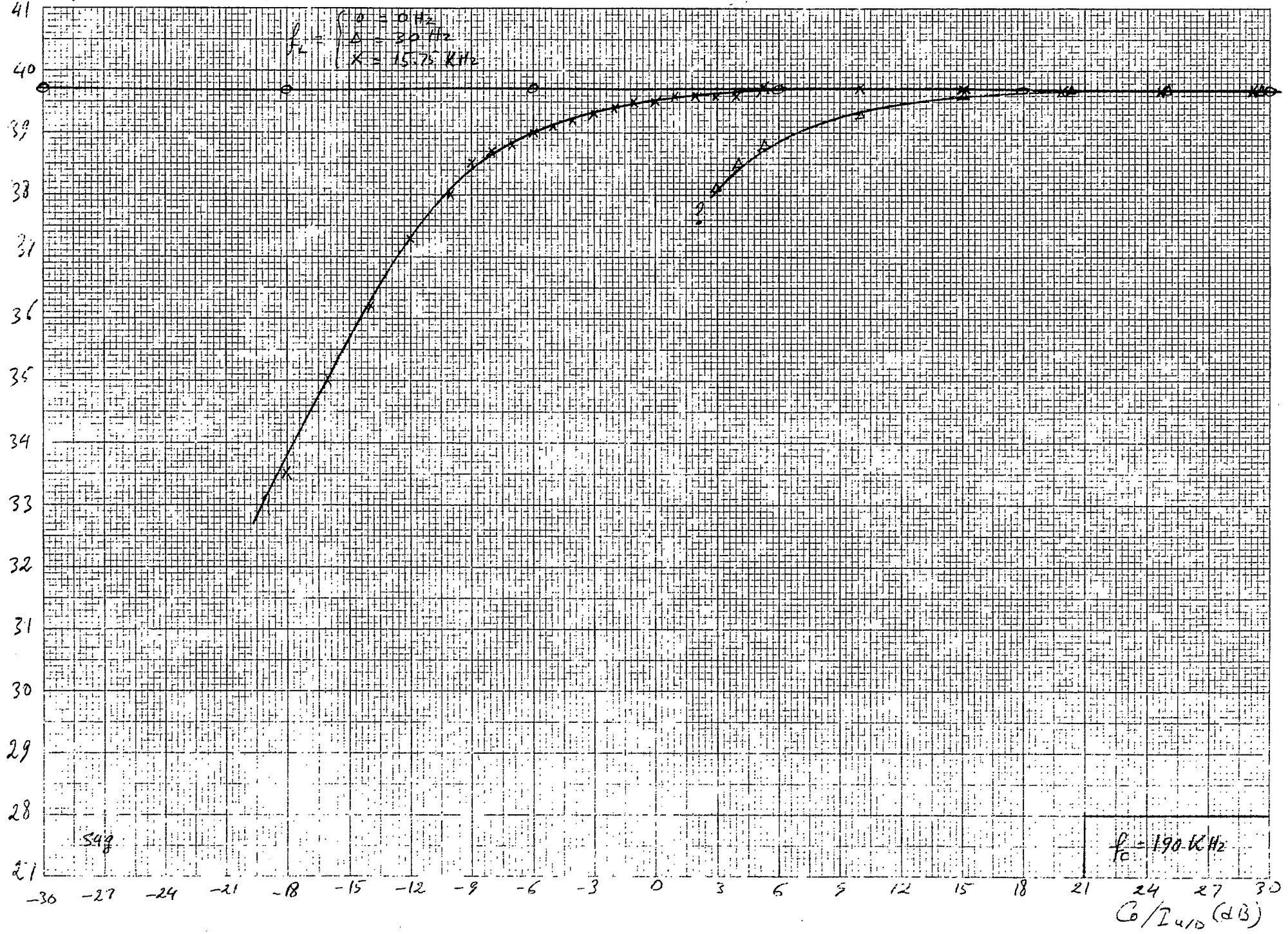
$f_c = 130 \text{ kHz}$

$C_0/I_{u,0}$ (dB)

S/N_L (dB)

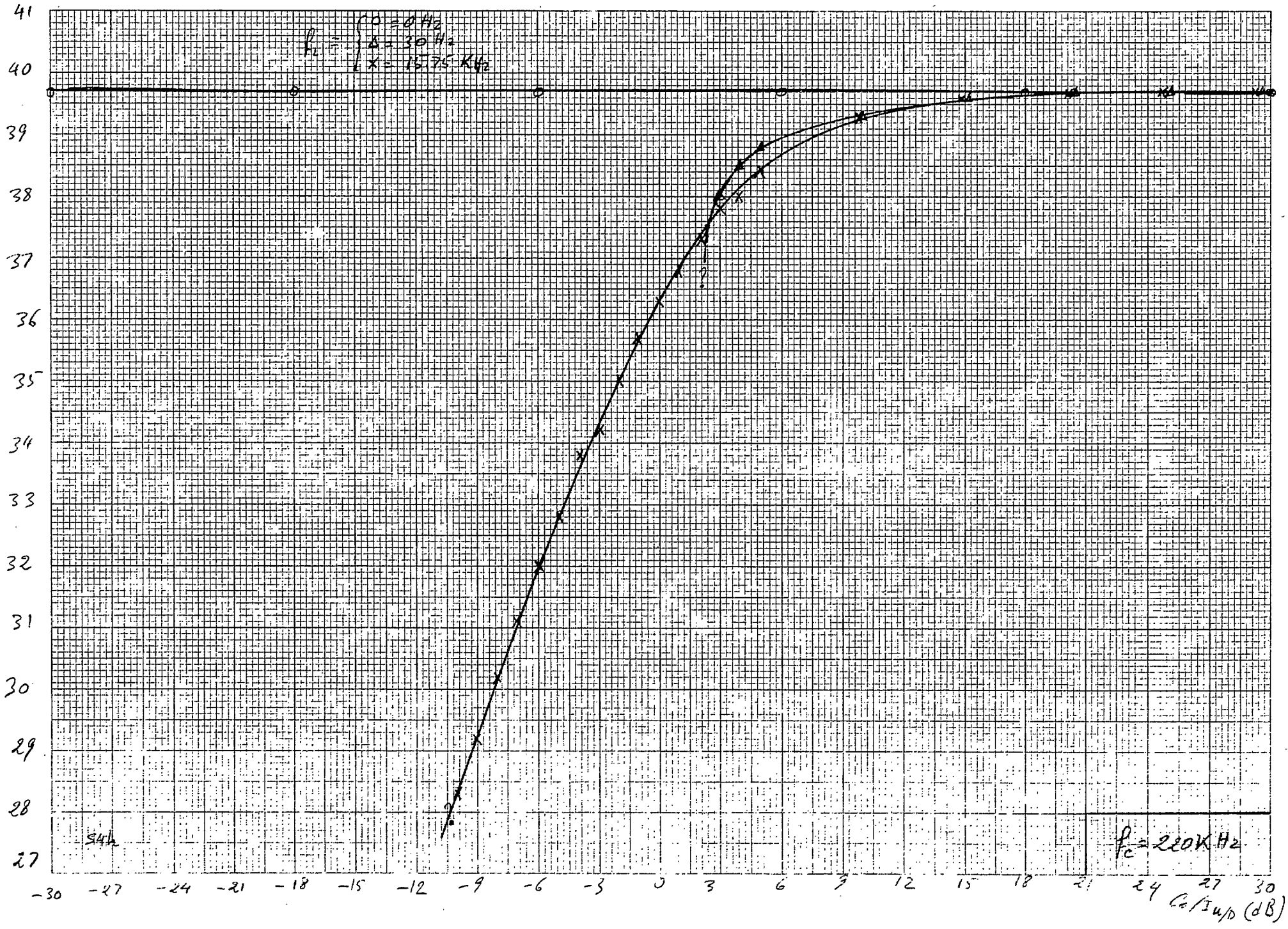


S/N_L (dB)

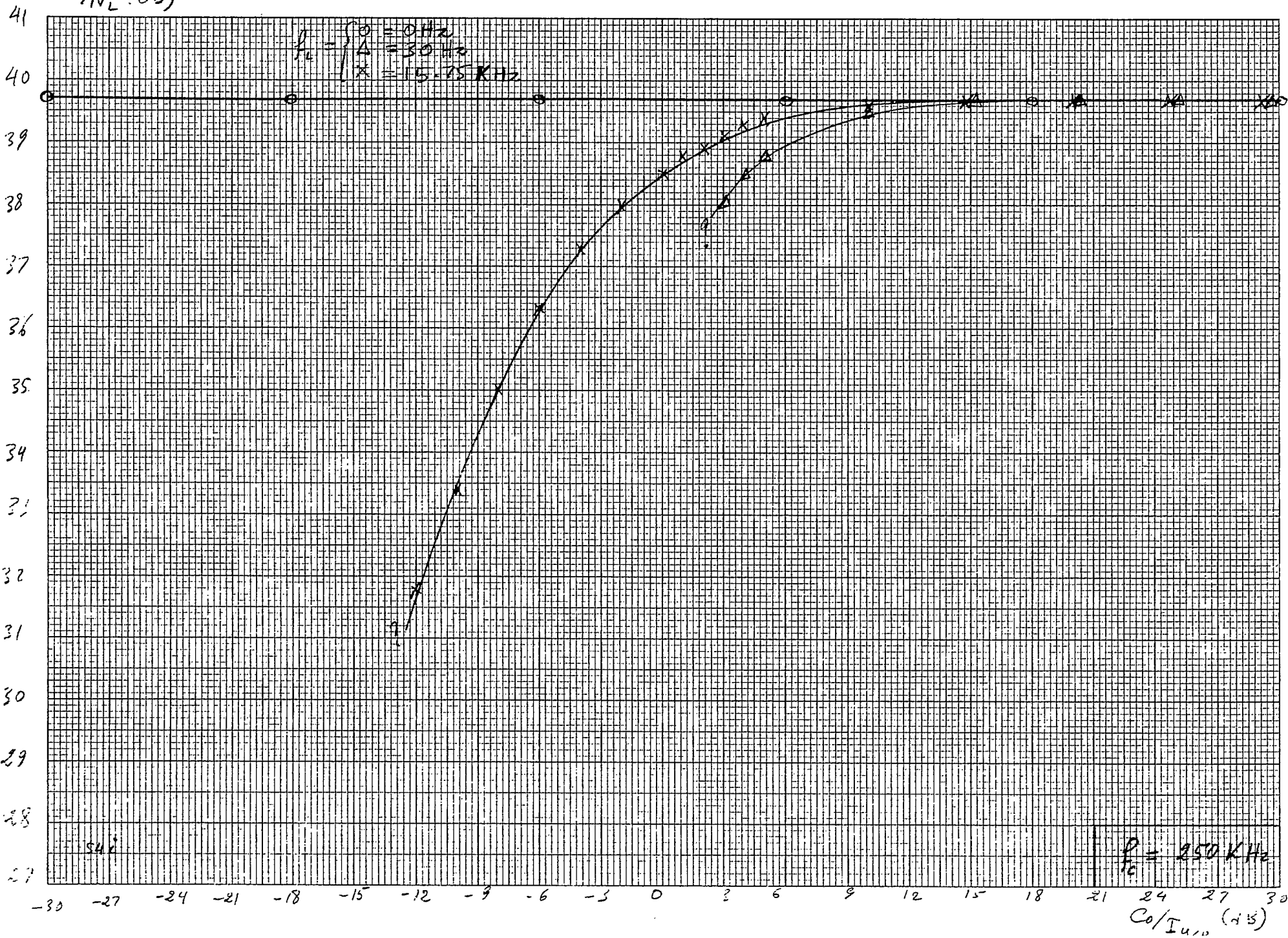


S/N_L (dB)

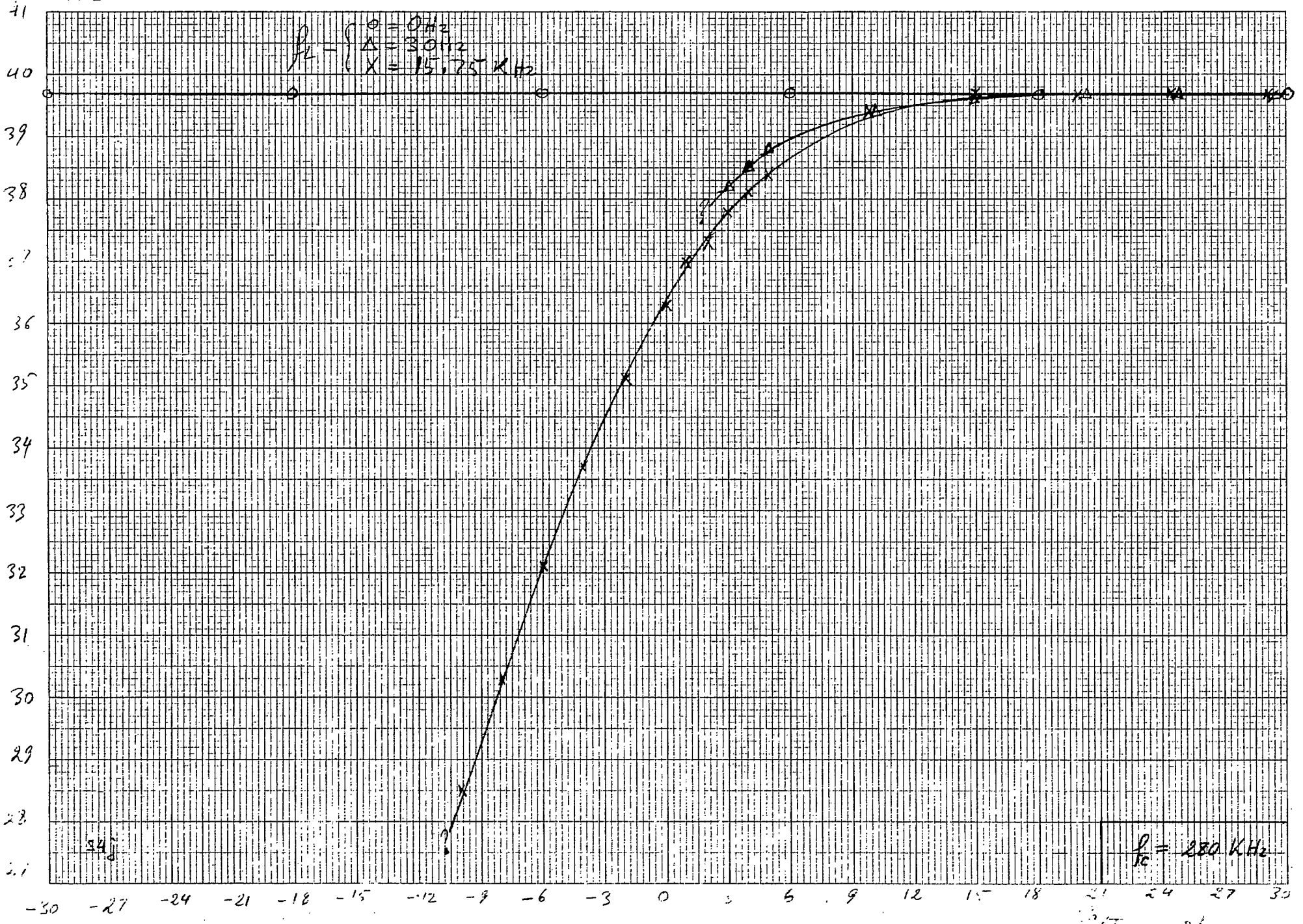
$f_0 = 0 \text{ Hz}$
 $\Delta f = 30 \text{ Hz}$
 $X = 15.75 \text{ KHz}$



S/N_L (dB)

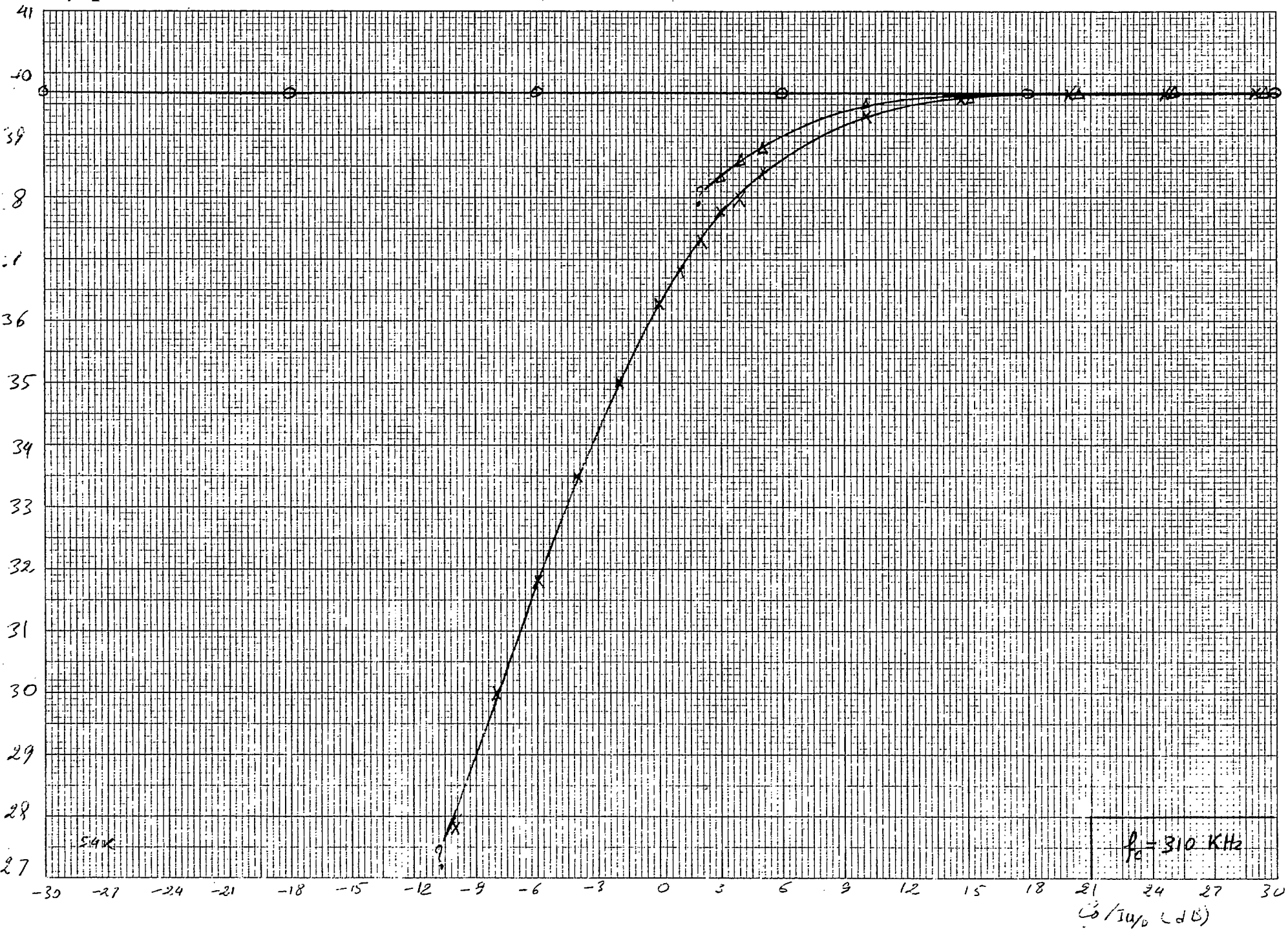


S/N_L (dB)

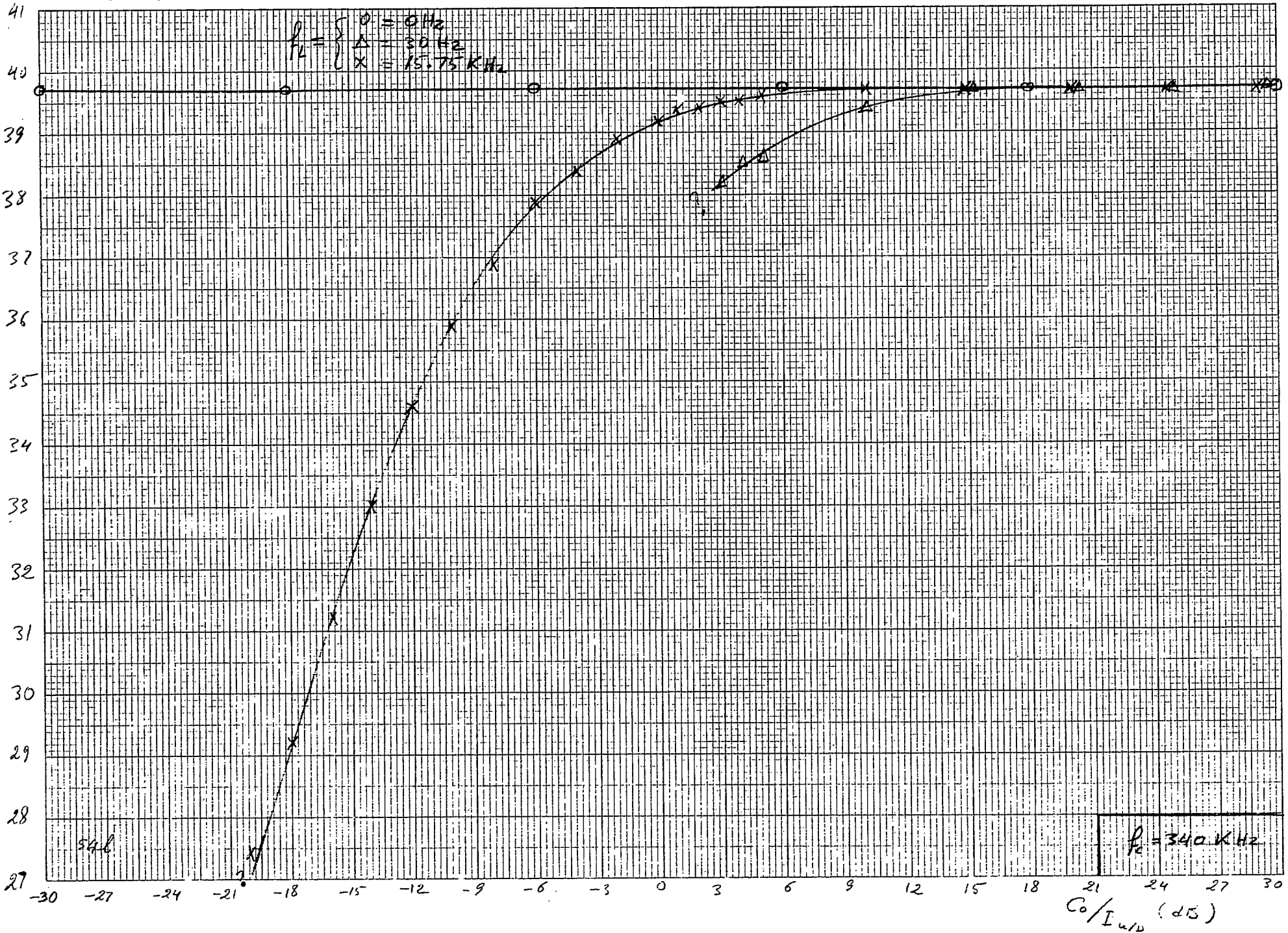


27
28
29
30
31
32
33
34
35
36
37
38
39
40
41

S/N_L (dB)



S/N_L (dB)



546

$f_c = 340 \text{ KHz}$

$C_0/I_{u/d}$ (dB)

C/N_c (dB)

